

# Robust Control

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# Outline

- Nonsingularity of Matrix Families
- LTI Robust Stability Analysis
- Time-Varying Uncertainties
- Robust Stability and Performance with Multipliers
- Controller Synthesis
- A Glimpse at Nonlinear Uncertainties and IQCs

# Nonsingularity of Families of Matrices

The following linear algebra problem is fundamental in robust control:

Let us be given  $M \in \mathbb{C}^{n \times m}$  and  $\mathbf{V} \subset \mathbb{C}^{m \times n}$ . Decide whether

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- If  $\sigma(V) \leq 1$  we can hence infer that  $\sigma_{\max}(M) < 1$  is sufficient for (NS). This **small-gain** condition is easy to check, **but conservative**.
- **First goal:** Develop more refined computationally verifiable sufficient conditions that take the particular structure of  $\mathbf{V}$  into account.

## Key Idea

Observe that

$$\det(I - MV) \neq 0 \iff \text{image} \begin{pmatrix} I \\ M \end{pmatrix} \cap \text{image} \begin{pmatrix} V \\ I \end{pmatrix} = \{0\}.$$

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**Geometrically** we guarantee separation if these graphs are located in the strictly negative/positive cone of some Hermitian  $P \in \mathbb{C}^{(n+m) \times (n+m)}$  respectively:

- Positive cone of  $P$ :  $\{x \in \mathbb{C}^{n+m} : x^* P x \geq 0\}$ .
- Strictly negative cone of  $P$ :  $\{x \in \mathbb{C}^{n+m} : x^* P x < 0\}$ .



# Multipliers

Suppose there exists a Hermitian **multiplier**  $P \in \mathbb{C}^{(n+m) \times (n+m)}$  with

$$\begin{pmatrix} V \\ I \end{pmatrix}^* P \begin{pmatrix} V \\ I \end{pmatrix} \succcurlyeq 0 \text{ for all } V \in \mathbf{V} \quad (\text{POS})$$

and at the same time

$$\begin{pmatrix} I \\ M \end{pmatrix}^* P \begin{pmatrix} I \\ M \end{pmatrix} \prec 0. \quad (\text{NS-LMI})$$

**Then**  $\det(I - MV) \neq 0$  for all  $V \in \mathbf{V}$ .

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Very easy to prove! Extensions? How to use in computations?

## Proof

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For  $z = Vw$  we infer  $z \neq 0$  and

$$\begin{pmatrix} z \\ w \end{pmatrix} = \begin{pmatrix} I \\ M \end{pmatrix} z \text{ as well as } \begin{pmatrix} z \\ w \end{pmatrix} = \begin{pmatrix} V \\ I \end{pmatrix} w.$$

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Since  $z \neq 0$  we obtain

$$0 > z^* \begin{pmatrix} I \\ M \end{pmatrix}^* P \begin{pmatrix} I \\ M \end{pmatrix} z = \begin{pmatrix} z \\ w \end{pmatrix}^* P \begin{pmatrix} z \\ w \end{pmatrix} = w^* \begin{pmatrix} V \\ I \end{pmatrix}^* P \begin{pmatrix} V \\ I \end{pmatrix} w \geq 0$$

which is a **contradiction**.

# Numerical Implementation

Note that the set of  $P$  satisfying (POS) is **convex**. However this does not help much since this constraint involves infinitely many LMIs.

## Key Idea for Computations: Relaxation

For practically relevant  $V$ , try to identify some "nicely described" subclass of all  $P$ 's which satisfy (POS).

Then search in this class of  $P$ 's one which also satisfies (NS-LMI).

Recent years have witnessed a whole variety of possibilities along this line, using techniques from convex analysis and real algebraic geometry (Pólya's theorem, sum-of-squares). We only give three **examples**.



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Let us consider the set  $\mathbf{V}$  of all  $V$  with

$$V = \text{diag}(V_1, \dots, V_p) \text{ with } V_1, \dots, V_p \in \mathbb{C}^{\bullet \times \bullet}, \sigma_{\max}(V) \leq 1.$$

These matrices admit a **diagonal structure with full complex blocks**.

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**Repeated Diagonal Multipliers.** (POS) is satisfied for all matrices

$$P = \begin{pmatrix} -\text{diag}(q_1 I, \dots, q_p I) & 0 \\ 0 & \text{diag}(q_1 I, \dots, q_p I) \end{pmatrix}$$

with arbitrary real numbers  $q_1, \dots, q_p \geq 0$ .

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**Proof.** Just note for  $V \in \mathbf{V}$  and  $P$  as above that

$$\begin{pmatrix} V \\ I \end{pmatrix}^* P \begin{pmatrix} V \\ I \end{pmatrix} = \text{diag}(q_k (I - V_k^* V_k)) \succcurlyeq 0.$$

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As we will see, another interesting set  $\mathbf{V}$  consists of all

$$\mathbf{V} = \text{diag}(v_1 I, \dots, v_p I) \text{ with } v_k \in \mathbb{R}, \quad |v_k| \leq 1, \quad k = 1, \dots, p.$$

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with Hermitian  $Q_1, \dots, Q_p \succcurlyeq 0$  and skew-Hermitian  $S_1, \dots, S_p$ .

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**Proof.** Just note for  $V \in \mathbf{V}$  and  $P$  as above that

$$\begin{pmatrix} V \\ I \end{pmatrix}^* P \begin{pmatrix} V \\ I \end{pmatrix} = \text{diag}(Q_k(1-v_k^2) + v_k(S_k + S_k^*)) = \text{diag}(Q_k(1-v_k^2)) \succcurlyeq 0.$$

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**Full Block Multipliers.** (POS) is satisfied for all  $P$  with

$$\begin{pmatrix} I \\ 0 \end{pmatrix}^T P \begin{pmatrix} I \\ 0 \end{pmatrix} \preceq 0, \quad \begin{pmatrix} V_k \\ I \end{pmatrix}^T P \begin{pmatrix} V_k \\ I \end{pmatrix} \succeq 0, \quad k = 1, \dots, N.$$

Set of multipliers described by **finitely many** LMI constraints.



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Set of multipliers described by **finitely many** LMI constraints.

**Proof.** The first inequality implies that

$$V \rightarrow \begin{pmatrix} V \\ I \end{pmatrix}^T P \begin{pmatrix} V \\ I \end{pmatrix} \text{ is concave.}$$

Hence (POS) is valid iff the inequality holds at the generators  $V_1, \dots, V_N$ .

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# Robust Stability

Consider the system

$$\dot{x} = F(\delta)x \quad \text{with uncertain parameter vector } \delta \in \mathbb{R}^p.$$

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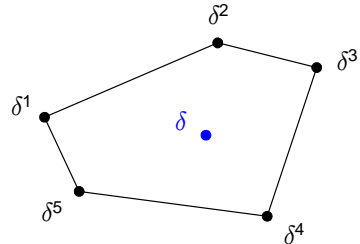
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**Assumptions:** The parameter  $\delta$  is contained in some polytope

$$\delta \in \mathcal{D} = \text{convex hull}\{\delta^1, \dots, \delta^N\} \subset \mathbb{R}^p.$$

Let  $\delta = 0 \in \mathcal{D}$  be the nominal value and assume that  $F(0)$  is Hurwitz.



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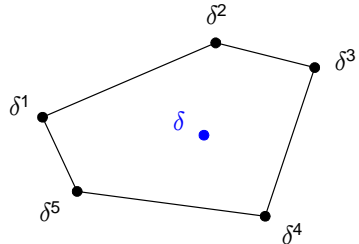
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Can we decide whether all these systems are Hurwitz?

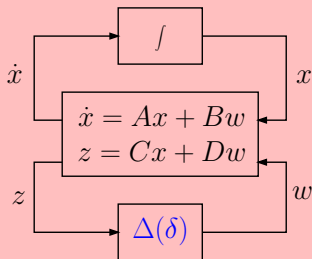
# Linear Fractional Representation

Let  $F(\delta)$  depend rationally on  $\delta$  without having a pole in 0.

One can then write  $\dot{x} = F(\delta)x$  as

$$\left. \begin{aligned} \dot{x} &= Ax + Bw \\ z &= Cx + Dw \end{aligned} \right\} w = \Delta(\delta)z$$

with  $\Delta(\delta)$  being **linear** in  $\delta \in \mathcal{D}$ .



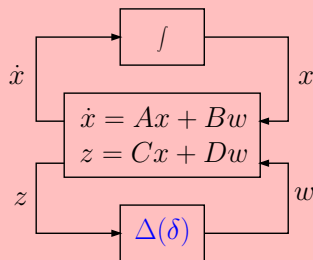
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with  $\Delta(\delta)$  being **linear** in  $\delta \in \delta$ .



It is always possible to take  $\Delta(\delta)$  with the structure

$$\Delta(\delta) = \begin{pmatrix} \delta_1 I_{\nu_1} & & 0 \\ & \ddots & \\ 0 & & \delta_p I_{\nu_p} \end{pmatrix} \text{ for some integers } \nu_1, \dots, \nu_p \geq 0$$

where  $I_\nu$  denotes the identity matrix of size  $\nu$ .

## Example I

$$\dot{x} = \begin{pmatrix} -1 & 2\delta_1 \\ -\frac{1}{2+\delta_1} & -4 + 3\delta_2 \end{pmatrix} x \quad \text{with} \quad |\delta_1| \leq r, \quad |\delta_2| \leq r, \quad r > 0.$$



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Rewrite as

$$\begin{pmatrix} 1 & 0 \\ 0 & 2 + \delta_1 \end{pmatrix} \begin{pmatrix} \dot{x}_1 \\ \dot{x}_2 \end{pmatrix} = \begin{pmatrix} -1 & 2\delta_1 \\ -1 & (-4 + 3\delta_2)(2 + \delta_1) \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}$$

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or as

$$\begin{pmatrix} \dot{x}_1 \\ 2\dot{x}_2 \end{pmatrix} + \begin{pmatrix} 0 \\ 1 \end{pmatrix} w_1 = \begin{pmatrix} -1 & 0 \\ -1 & -4 + 3\delta_2 \end{pmatrix} \begin{pmatrix} x_1 \\ 2x_2 \end{pmatrix} + \begin{pmatrix} 2 \\ -4 + 3\delta_2 \end{pmatrix} w_2$$

$$w_1 = \delta_1 z_1, \quad z_1 = \dot{x}_2, \quad w_2 = \delta_1 z_2, \quad z_2 = x_2$$

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## Example I

Hence  $\dot{x} = F(\delta)x$  can be written as

$$\dot{x} = \begin{pmatrix} -1 & 0 \\ -0.5 & -4 \end{pmatrix} x + \begin{pmatrix} 0 & 2 & 0 \\ -0.5 & -2 & 1.5 \end{pmatrix} w$$
$$z = \begin{pmatrix} -0.5 & -4 \\ 0 & 1 \\ 0 & 2 \end{pmatrix} x + \begin{pmatrix} -0.5 & -2 & 1.5 \\ 0 & 0 & 0 \\ 0 & 1 & 0 \end{pmatrix} w, \quad w = \begin{pmatrix} \delta_1 & 0 & 0 \\ 0 & \delta_1 & 0 \\ 0 & 0 & \delta_2 \end{pmatrix} z.$$

Therefore we can choose

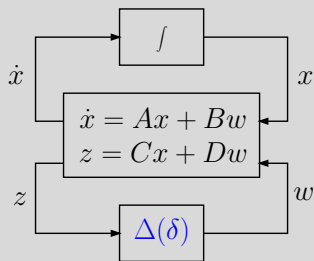
$$\left( \begin{array}{c|c} A & B \\ \hline C & D \end{array} \right) = \left( \begin{array}{cc|ccc} -1 & 0 & 0 & 2 & 0 \\ -0.5 & -4 & -0.5 & -2 & 1.5 \\ \hline -0.5 & -4 & -0.5 & -2 & 1.5 \\ 0 & 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 1 & 0 \end{array} \right), \quad \Delta(\delta) = \begin{pmatrix} \delta_1 & 0 & 0 \\ 0 & \delta_1 & 0 \\ 0 & 0 & \delta_2 \end{pmatrix}.$$

# Linear Fractional Representation

Let  $\dot{x} = F(\delta)x$  be represented as

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with  $\Delta(\delta)$  being linear in  $\delta \in \mathcal{D}$ .



The given derivation shows that this actually means

$$F(\delta) = A + B\Delta(\delta)(I - D\Delta(\delta))^{-1}C =: \begin{pmatrix} A & B \\ C & D \end{pmatrix} \star \Delta(\delta).$$

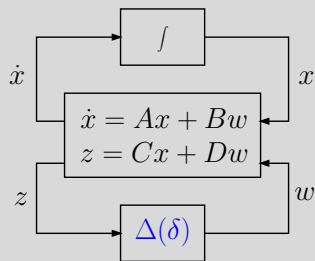
This is called a **linear fractional representation** (LFR) of  $F(\delta)$ .  
It is said to be **well-posed** if  $I - D\Delta(\delta)$  is non-singular on  $\mathcal{D}$ .

# Nonconservative Robust Stability Test

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Define the transfer matrix  $G(s) = D + C(sI - A)^{-1}B$ .

The LFR is well-posed and  $\dot{x} = F(\delta)x$  is Hurwitz for all  $\delta \in \mathcal{D}$  iff

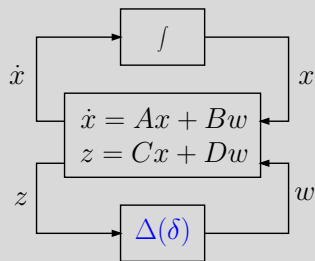
$$\det(I - G(i\omega)\Delta(\delta)) \neq 0 \quad \text{for all } \omega \in \mathbb{R} \cup \{\infty\}, \delta \in \mathcal{D}.$$

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Testing robust stability is reduced to a robust non-singularity condition for the matrices  $G(i\omega)$ ,  $\omega \in \mathbb{R} \cup \{\infty\}$ , and the set  $\mathbf{V} = \Delta(\delta)$ .

## Sketch of Proof

Suppose the LFR is well-posed. Then  $F(\delta)$  is Hurwitz for all  $\delta \in \mathcal{D}$  iff  $\det(sI - A - B\Delta(\delta)(I - D\Delta(\delta))^{-1}C) \neq 0$  for all  $\operatorname{Re}(s) \geq 0, \delta \in \mathcal{D}$



## Sketch of Proof

Suppose the LFR is well-posed. Then  $F(\delta)$  is Hurwitz for all  $\delta \in \mathcal{D}$  iff  $\det (sI - A - B\Delta(\delta)(I - D\Delta(\delta))^{-1}C) \neq 0$  for all  $\operatorname{Re}(s) \geq 0$ ,  $\delta \in \mathcal{D}$

$\iff$  (Schur determinant formula & well-posedness)

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Now just observe that the latter condition is well-posedness for  $\omega = \infty$ .

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- Choose a list frequencies  $\omega_1, \dots, \omega_m \in \mathbb{R} \cup \{\infty\}$ .

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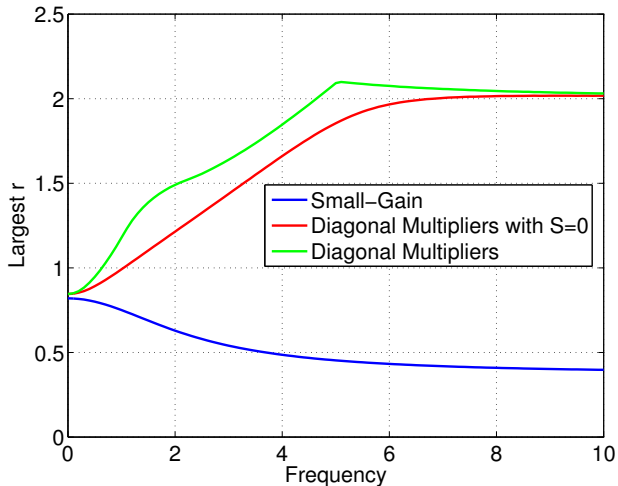
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If all LMIs for  $k = 1, \dots, m$  are feasible, we conclude robust stability.

Note that there is a risk of missing crucial frequencies! Can be handled!

## Example I

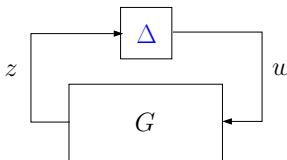
Determine largest  $r$  such that the non-singularity test is successful.



**Observation:** The larger the class of multipliers the better the test!

## Summary and Comments

- Sketched key ideas to obtain linear fractional representation



which forms the basis for advanced robustness analysis.

- Have reduced robust stability to non-singularity test.
- Developed multiplier relaxation schemes to verify robust stability.
- Extends to stable **structured dynamic** uncertainties  $\Delta$  that satisfy

$$\Delta(i\omega) \in \mathbf{V} \text{ for all } \omega \in \mathbb{R} \cup \{\infty\}.$$

This touches the so-called **structured singular value** theory.

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- LTI Robust Stability Analysis
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# Robust Stability: Time-Varying Uncertainties

Consider the system

$$\dot{x}(t) = F(\delta(t))x(t) \text{ affected by time-varying parameter } \delta(t) \in \mathbb{R}^p.$$

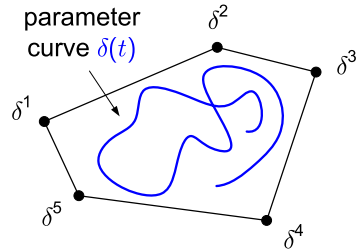
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**Assumption:**  $\delta(\cdot)$  is piece-wise continuous and satisfies  $\delta(t) \in \delta$  for all  $t \in \mathbb{R}_+$  where

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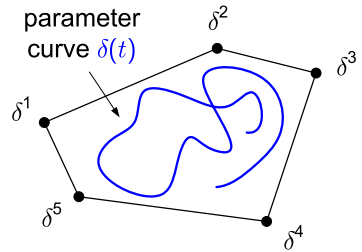
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## Quadratic Stability

All systems are exponentially stable if there exists some  $X$  with

$$X \succ 0, \quad F(\delta)^T X + X F(\delta) \prec 0 \text{ for all } \delta \in \delta.$$



## How to Check?

Consider the **robust LMI**

$$\begin{pmatrix} I \\ F(\delta) \end{pmatrix}^T \begin{pmatrix} 0 & X \\ X & 0 \end{pmatrix} \begin{pmatrix} I \\ F(\delta) \end{pmatrix} = F(\delta)^T X + X F(\delta) \prec 0 \quad \forall \delta \in \delta.$$

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LFR is well-posed and the robust LMI holds **iff** there exists a  $P$  with

$$\begin{pmatrix} \Delta(\delta) \\ I \end{pmatrix}^T P \begin{pmatrix} \Delta(\delta) \\ I \end{pmatrix} \succcurlyeq 0 \quad \text{for all } \delta \in \delta \quad (\text{POS})$$

that also satisfies

$$\begin{pmatrix} A^T X + X A & X B \\ B^T X & 0 \end{pmatrix} + \begin{pmatrix} 0 & I \\ C & D \end{pmatrix}^T P \begin{pmatrix} 0 & I \\ C & D \end{pmatrix} \prec 0. \quad (\text{QS-LMI})$$

Our numerical procedure applies for checking sufficient conditions!

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Abbreviate  $H = (I - D\Delta(\delta))^{-1}C$  to infer from (QS-LMI) that

$$\begin{aligned} 0 \succ \begin{pmatrix} I \\ \Delta(\delta)H \end{pmatrix}^T \text{ lhs of LMI} \begin{pmatrix} I \\ \Delta(\delta)H \end{pmatrix} &= \\ &= F(\delta)^T X + X F(\delta) + \underbrace{H^T \begin{pmatrix} \Delta(\delta) \\ I \end{pmatrix}^T P \begin{pmatrix} \Delta(\delta) \\ I \end{pmatrix} H}_{\succ 0 \text{ due to (POS)}} \end{aligned}$$

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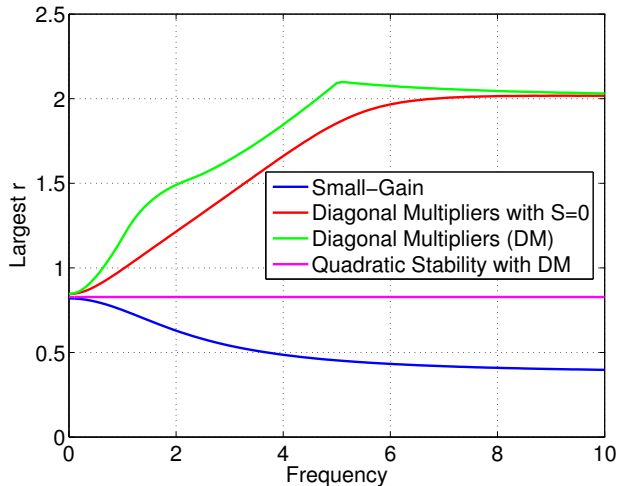
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where  $=$  follows by simple computation. Hence  $F(\delta)^T X + X F(\delta) \prec 0$ .

## Example I

Determine largest  $r$  such that robust quadratic stability can be verified.



**Observation:** Small gap between time-invariant/time-varying case!

## Example II

Test quadratic stability for polynomial parameter dependence:

$$\dot{x} = \begin{pmatrix} -1.25 & 1 - \delta_1 \delta_2^2 \\ 1 - \delta_1 \delta_2 & -1 \end{pmatrix} x, \quad \delta_1 \in [-1, 1], \quad \delta_2 \in [-1, 0].$$



## Example II

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This family can be covered by the following uncertain system with affine parameter dependence:

$$\dot{x} = \begin{pmatrix} -1.25 & 1 - x \\ 1 - y & -1 \end{pmatrix} x, \quad x \in [-r, r], \quad y \in [-r, r], \quad r = 1.$$

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- Polytopic technique from Lecture 6 successful only for  $r \approx 0.11$ .
- The multipliers from slide 9 allow to guarantee quadratic stability for original uncertain system with polynomial parameter dependence!

## Example II: What's going on?

### Blue Region

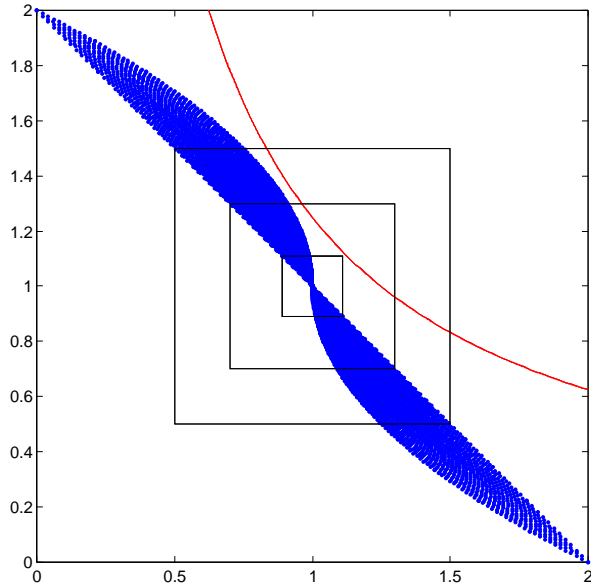
Set of parameter-dependent elements of original system

### Red Line

Boundary of Hurwitz region for original system

### Black Boxes

Set of parameter-dependent elements of affine covering



## Comments

- A trajectory-based proof for robust stability will be given below.
- With affine  $Q_0(v)$ ,  $R_0(v)$ ,  $S_0(v)$  in the decision variable  $v$ , the same technique (proof) applies to finding  $v$  which robustly satisfies

$$\begin{pmatrix} I \\ F(\delta) \end{pmatrix}^T \begin{pmatrix} Q_0(v) & S_0(v) \\ S_0(v)^T & R_0(v) \end{pmatrix} \begin{pmatrix} I \\ F(\delta) \end{pmatrix} \prec 0, \quad R_0(v) \succcurlyeq 0.$$

**Examples:** Discrete-time stability, eigenvalue-location in LMI region.

- The result is a concrete version of the so-called full block S-procedure. It serves to handle general robust LMI problems in which the uncertain parameters enter in a rational fashion.

C.W. Scherer, LMI Relaxations in Robust Control, Eur. J. Cont. 12 (2006) 3-29.

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# Robust Energy-Gain Performance

Consider the uncertain input-output system described as

$$\begin{aligned}\dot{x}(t) &= F(\delta(t))x(t) + G(\delta(t))d(t) \\ e(t) &= H(\delta(t))x(t) + J(\delta(t))d(t)\end{aligned}$$

with continuous parameter-curves  $\delta(\cdot)$  that satisfy

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## Robust Energy-Gain Performance of level $\gamma$

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## Robust Energy-Gain Performance of level $\gamma$

For all parameter-curves,  $\dot{x}(t) = F(\delta(t))x(t)$  is exponentially stable and the system's energy-gain is bounded by  $\gamma$ :

$$\int_0^\infty e(t)^T e(t) dt \leq \gamma^2 \int_0^\infty d(t)^T d(t) dt \text{ for } d \in \mathcal{L}_2(\mathbb{R}_+, \mathbb{R}^{n_d}), x(0) = 0.$$



## Robust $\mathcal{L}_2$ -Gain Performance

Obtain LFR of matrices describing system:

$$\begin{pmatrix} F(\delta) & G(\delta) \\ H(\delta) & J(\delta) \end{pmatrix} = \left( \begin{array}{cc|c} A & B_1 & B_2 \\ C_1 & D_1 & D_{12} \\ \hline C_2 & D_{21} & D_2 \end{array} \right) \star \Delta(\delta).$$

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If LFR well-posed, we have the following alternative system description:

$$\begin{pmatrix} x(t) \\ \dot{x}(t) \\ d(t) \\ e(t) \\ w(t) \\ z(t) \end{pmatrix} = \begin{pmatrix} I & 0 & 0 \\ A & B_1 & B_2 \\ 0 & I & 0 \\ C_1 & D_1 & D_{12} \\ 0 & 0 & I \\ C_2 & D_{21} & D_2 \end{pmatrix} \begin{pmatrix} x(t) \\ d(t) \\ w(t) \end{pmatrix}, \quad \begin{pmatrix} w(t) \\ z(t) \end{pmatrix} = \begin{pmatrix} \Delta(\delta(t)) \\ I \end{pmatrix} z(t).$$

## Robust $\mathcal{L}_2$ -Gain Performance

The LFR is well-posed and the system satisfies robust quadratic performance if there exist  $P = \begin{pmatrix} Q & S \\ S^T & R \end{pmatrix}$  and  $X \succ 0$  with

$$\begin{pmatrix} \Delta(\delta) \\ I \end{pmatrix}^T P \begin{pmatrix} \Delta(\delta) \\ I \end{pmatrix} \succ 0 \text{ for all } \delta \in \mathcal{D} \text{ and} \quad (\text{POS})$$

$$\begin{pmatrix} I & 0 & 0 \\ A & B_1 & B_2 \\ 0 & I & 0 \\ C_1 & D_1 & D_{12} \\ 0 & 0 & I \\ C_2 & D_{21} & D_2 \end{pmatrix}^T \left( \begin{array}{ccc|ccc} 0 & X & 0 & 0 & 0 & 0 \\ X & 0 & 0 & 0 & 0 & 0 \\ \hline 0 & 0 & -\gamma^2 I & 0 & 0 & 0 \\ 0 & 0 & 0 & I & 0 & 0 \\ \hline 0 & 0 & 0 & 0 & Q & S \\ 0 & 0 & 0 & 0 & S^T & R \end{array} \right) \begin{pmatrix} I & 0 & 0 \\ A & B_1 & B_2 \\ 0 & I & 0 \\ C_1 & D_1 & D_{12} \\ 0 & 0 & I \\ C_2 & D_{21} & D_2 \end{pmatrix} \prec 0. \quad (\text{RP})$$

## Proof of Well-Posedness

Considering the right-lower block of (RP) reveals that

$$D_{12}^T D_{12} + \begin{pmatrix} I \\ D_2 \end{pmatrix}^T P \begin{pmatrix} I \\ D_2 \end{pmatrix} =$$

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This implies  $\begin{pmatrix} I \\ D_2 \end{pmatrix}^T P \begin{pmatrix} I \\ D_2 \end{pmatrix} \prec 0$  and with (POS) well-posedness.

## Sketch of Trajectory-Based Proof

Due to (RP) there exists some  $\epsilon > 0$  such that

$$\begin{pmatrix} I & 0 & 0 \\ A & B_1 & B_2 \\ \hline 0 & I & 0 \\ C_1 & D_1 & D_{12} \\ \hline 0 & 0 & I \\ C_2 & D_{21} & D_2 \end{pmatrix}^T \begin{pmatrix} 0 & X & 0 & 0 & 0 & 0 \\ \hline X & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & -\gamma^2 I & 0 & 0 & 0 \\ 0 & 0 & 0 & I & 0 & 0 \\ \hline 0 & 0 & 0 & 0 & Q & S \\ 0 & 0 & 0 & 0 & S^T & R \end{pmatrix} \begin{pmatrix} I & 0 & 0 \\ A & B_1 & B_2 \\ \hline 0 & I & 0 \\ C_1 & D_1 & D_{12} \\ \hline 0 & 0 & I \\ C_2 & D_{21} & D_2 \end{pmatrix} + \begin{pmatrix} \epsilon X & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{pmatrix} \prec 0.$$

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## Sketch of Trajectory-Based Proof

Due to (RP) there exists some  $\epsilon > 0$  such that

$$\begin{pmatrix} I & 0 & 0 \\ A & B_1 & B_2 \\ \hline 0 & I & 0 \\ C_1 & D_1 & D_{12} \\ \hline 0 & 0 & I \\ C_2 & D_{21} & D_2 \end{pmatrix}^T \begin{pmatrix} 0 & X & 0 & 0 & 0 & 0 \\ X & 0 & 0 & 0 & 0 & 0 \\ \hline 0 & 0 & -\gamma^2 I & 0 & 0 & 0 \\ 0 & 0 & 0 & I & 0 & 0 \\ \hline 0 & 0 & 0 & 0 & Q & S \\ 0 & 0 & 0 & 0 & S^T & R \end{pmatrix} \begin{pmatrix} I & 0 & 0 \\ A & B_1 & B_2 \\ \hline 0 & I & 0 \\ C_1 & D_1 & D_{12} \\ \hline 0 & 0 & I \\ C_2 & D_{21} & D_2 \end{pmatrix} + \begin{pmatrix} \epsilon X & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{pmatrix} \prec 0.$$

Choose parameter trajectory  $\delta(t) \in \delta$  and  $d \in \mathcal{L}_2(\mathbb{R}_+, \mathbb{R}^{n_d})$ , and let  $x(\cdot)$  and  $e(\cdot)$  be some corresponding state- and output trajectories for any initial condition  $x(0)$ . Due to well-posedness, these trajectories satisfy the relations on slide 28 for suitable  $w(\cdot)$ ,  $z(\cdot)$ .

Now right-multiply  $\text{col}(x(t), d(t), w(t))$  and left-multiply its transpose.

# Sketch of Trajectory-Based Proof

We obtain

$$\begin{aligned} & \begin{pmatrix} x(t) \\ \dot{x}(t) \end{pmatrix}^T \begin{pmatrix} 0 & X \\ X & 0 \end{pmatrix} \begin{pmatrix} x(t) \\ \dot{x}(t) \end{pmatrix} + \epsilon x(t)^T X x(t) + \\ & + \begin{pmatrix} d(t) \\ e(t) \end{pmatrix}^T \begin{pmatrix} -\gamma^2 I & 0 \\ 0 & I \end{pmatrix} \begin{pmatrix} d(t) \\ e(t) \end{pmatrix} + \begin{pmatrix} w(t) \\ z(t) \end{pmatrix}^T P \begin{pmatrix} w(t) \\ z(t) \end{pmatrix} \leq 0. \end{aligned}$$

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As a key feature observe that  $w(t) = \Delta(\delta(t))z(t)$  and hence with (POS):

$$\begin{pmatrix} w(t) \\ z(t) \end{pmatrix}^T P \begin{pmatrix} w(t) \\ z(t) \end{pmatrix} = z(t)^T \begin{pmatrix} \Delta(\delta(t)) \\ I \end{pmatrix}^T P \begin{pmatrix} \Delta(\delta(t)) \\ I \end{pmatrix} z(t) \succcurlyeq 0.$$

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Even after canceling this term, the above inequality hence stays valid.

# Sketch of Trajectory-Based Proof

With the product-rule we arrive at

$$\frac{d}{dt}x(t)^T X x(t) + \epsilon x(t)^T X x(t) + e(t)^T e(t) - \gamma^2 d(t)^T d(t) \leq 0.$$

# Sketch of Trajectory-Based Proof

With the product-rule we arrive at

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- If  $d(\cdot) = 0$  we infer  $\frac{d}{dt}x(t)^T \mathbf{X}x(t) + \epsilon x(t)^T \mathbf{X}x(t) \leq 0$ . Exploit  $\mathbf{X} \succ 0$  to obtain uniform exponential stability as in Lecture 6.

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- If  $x(0) = 0$  drop the term  $\epsilon x(t)^T \mathbf{X}x(t)$  and observe that the inequality stays true.

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- If  $x(0) = 0$  drop the term  $\epsilon x(t)^T \mathbf{X}x(t)$  and observe that the inequality stays true. We then infer by integration on  $[0, T]$  that

$$x(T)^T \mathbf{X}x(T) + \int_0^T e(t)^T e(t) - \gamma^2 d(t)^T d(t) dt \leq 0.$$



## Sketch of Trajectory-Based Proof

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Since  $\mathbf{X} \succ 0$ , we can drop term  $x(T)^T \mathbf{X}x(T)$  without violating the inequality.

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- If  $x(0) = 0$  drop the term  $\epsilon x(t)^T \mathbf{X}x(t)$  and observe that the inequality stays true. We then infer by integration on  $[0, T]$  that

$$x(T)^T \mathbf{X}x(T) + \int_0^T e(t)^T e(t) - \gamma^2 d(t)^T d(t) dt \leq 0.$$

Since  $\mathbf{X} \succ 0$ , we can drop term  $x(T)^T \mathbf{X}x(T)$  without violating the inequality. Then  $T \rightarrow \infty$  finally leads to  $\|e\|_{\mathcal{L}_2(\mathbb{R}_+, \mathbb{R}^{n_e})} \leq \gamma \|d\|_{\mathcal{L}_2(\mathbb{R}_+, \mathbb{R}^{n_d})}$ .

## How to Apply?

Test feasibility of LMIs

$$\begin{pmatrix} I \\ 0 \end{pmatrix}^T P \begin{pmatrix} I \\ 0 \end{pmatrix} \prec 0, \quad \begin{pmatrix} \Delta(\delta^k) \\ I \end{pmatrix}^T P \begin{pmatrix} \Delta(\delta^k) \\ I \end{pmatrix} \succ 0, \quad k = 1, \dots, N,$$

$$X \succ 0, \quad \left( \begin{array}{ccc|ccc} I & 0 & 0 & 0 & X & 0 & 0 & 0 & 0 & 0 \\ A & B_1 & B_2 & X & 0 & 0 & 0 & 0 & 0 & 0 \\ \hline 0 & I & 0 & 0 & 0 & -\gamma^2 I & 0 & 0 & 0 & 0 \\ C_1 & D_1 & D_{12} & 0 & 0 & 0 & I & 0 & 0 & 0 \\ \hline 0 & 0 & I & 0 & 0 & 0 & 0 & Q & S & 0 \\ C_2 & D_{21} & D_2 & 0 & 0 & 0 & 0 & S^T & R & 0 \end{array} \right) \prec 0.$$

Feasibility guarantees a robust energy-gain level of  $\gamma$ .

Minimize  $\gamma^2$  to determine best possible bound ... with this technique.

## Summary and Comments

- Obtained non-trivial robust stability and robust performance tests which are based on multiplier relaxations.
- Observed trade-off between conservatism and “size” of multiplier set (computational complexity).
- Substantially more instances of the same theme are known.

**Examples:** Uncertainty phase information in  $\mu$ -theory  
Parameter-dependent Lyapunov functions  
Semi-algebraic uncertainty sets

C.W. Scherer, LMI Relaxations in Robust Control, Eur. J. Cont. 12 (2006) 3-29.

- Finally: Hints on synthesis. Larger classes of uncertainties.

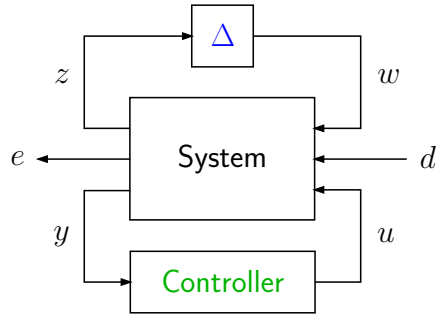
# Outline

- Nonsingularity of Matrix Families
- LTI Robust Stability Analysis
- Time-Varying Uncertainties
- Robust Stability and Performance with Multipliers
- Controller Synthesis
- A Glimpse at Nonlinear Uncertainties and IQCs

# Configuration for Robust Controller Synthesis

Design **controller** guaranteeing

- robust stability
- desired robust performance specification on  $d \rightarrow e$ .



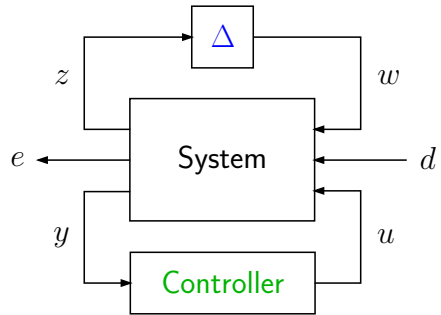
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Consider following approach:

- Use robust performance characterization with multipliers
- Try to satisfy the multiplier characterization with suitable controller



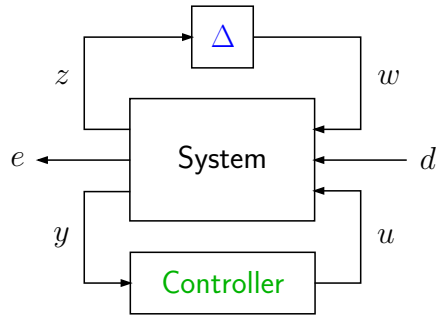
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Design **controller** guaranteeing

- robust stability
- desired robust performance specification on  $d \rightarrow e$ .

Consider following approach:

- Use robust performance characterization with multipliers
- Try to satisfy the multiplier characterization with suitable controller



For notational simplicity: Concentrate on robust quadratic stability with full-block multiplier relaxation.



# System Descriptions

Uncontrolled LTI part:

$$\begin{aligned}\dot{x} &= Ax + B_1w + Bu \\ z &= C_1x + D_1w + Eu \\ y &= Cx + Fw\end{aligned}$$

$w$ : uncertainty input

$z$ : uncertainty output

$u$ : control input

$y$ : measured output

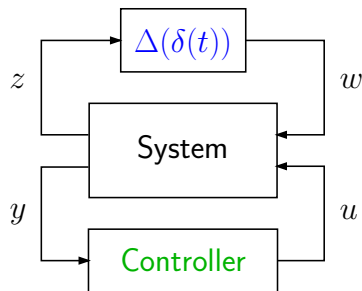
Controller:

$$\begin{aligned}\dot{x}_c &= A_Kx_c + B_Ky \\ u &= C_Kx_c + D_Ky\end{aligned}$$

Controlled LTI part:

$$\begin{aligned}\dot{\xi} &= \mathcal{A}\xi + \mathcal{B}w \\ z &= \mathcal{C}\xi + \mathcal{D}w\end{aligned}$$

Uncertainty:  $w(t) = \Delta(\delta(t))z(t)$ .



## Robust Stability Analysis Inequalities

Assume  $\delta(t) \in \delta = \text{co}\{\delta^1, \dots, \delta^N\}$ .

Robust stability guaranteed if exist  $\mathcal{X}$  and  $Q, R, S$  with

$$\begin{aligned} \begin{pmatrix} I \\ 0 \end{pmatrix}^T P \begin{pmatrix} I \\ 0 \end{pmatrix} \prec 0, \quad \begin{pmatrix} \Delta(\delta^k) \\ I \end{pmatrix}^T P \begin{pmatrix} \Delta(\delta^k) \\ I \end{pmatrix} \succ 0, \quad k = 1, \dots, N, \\ \mathcal{X} \succ 0, \quad \begin{pmatrix} I & 0 \\ \mathcal{X}A & \mathcal{X}B \\ 0 & I \\ \mathcal{C} & \mathcal{D} \end{pmatrix}^T \begin{pmatrix} 0 & I & 0 & 0 \\ I & 0 & 0 & 0 \\ 0 & 0 & Q & S \\ 0 & 0 & S^T & R \end{pmatrix} \begin{pmatrix} I & 0 \\ \mathcal{X}A & \mathcal{X}B \\ 0 & I \\ \mathcal{C} & \mathcal{D} \end{pmatrix} \prec 0. \end{aligned}$$

Apply standard procedure to step from analysis to synthesis.

## Robust Synthesis Inequalities

Exists controller guaranteeing robust stability if exist  $v$ ,  $Q$ ,  $R$ ,  $S$ :

$$\begin{aligned} & \begin{pmatrix} I \\ 0 \end{pmatrix}^T P \begin{pmatrix} I \\ 0 \end{pmatrix} \prec 0, \quad \begin{pmatrix} \Delta(\delta^k) \\ I \end{pmatrix}^T P \begin{pmatrix} \Delta(\delta^k) \\ I \end{pmatrix} \succ 0, \quad k = 1, \dots, N, \\ & X(v) \succ 0, \quad \begin{pmatrix} I & 0 \\ \mathbf{A}(v) & \mathbf{B}(v) \\ 0 & I \\ \mathbf{C}(v) & \mathbf{D}(v) \end{pmatrix}^T \left( \begin{array}{cc|cc} 0 & I & 0 & 0 \\ I & 0 & 0 & 0 \\ \hline 0 & 0 & Q & S \\ 0 & 0 & S^T & R \end{array} \right) \begin{pmatrix} I & 0 \\ \mathbf{A}(v) & \mathbf{B}(v) \\ 0 & I \\ \mathbf{C}(v) & \mathbf{D}(v) \end{pmatrix} \prec 0. \end{aligned}$$

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Unfortunately **not convex** in all variables  $v$  and  $Q$ ,  $R$ ,  $S$ !

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No technique known how to convexify in general!

Usual heuristic remedy: Controller multiplier iteration.

## Dualization Lemma

For real matrices  $P = P^T$  and  $W$  of compatible size, the conditions

$$\begin{pmatrix} 0 \\ I \end{pmatrix}^T P \begin{pmatrix} 0 \\ I \end{pmatrix} \succ 0 \quad \text{and} \quad \begin{pmatrix} I \\ W \end{pmatrix}^T P \begin{pmatrix} I \\ W \end{pmatrix} \prec 0$$

are equivalent to

$$\begin{pmatrix} I \\ 0 \end{pmatrix}^T P^{-1} \begin{pmatrix} I \\ 0 \end{pmatrix} \prec 0 \quad \text{and} \quad \begin{pmatrix} W^T \\ -I \end{pmatrix}^T P^{-1} \begin{pmatrix} W^T \\ -I \end{pmatrix} \succ 0.$$

Note that  $\text{im} \begin{pmatrix} I \\ W \end{pmatrix}^\perp$  equals  $\text{im} \begin{pmatrix} W^T \\ -I \end{pmatrix}$ .

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**In general:** Let  $P = P^T$  be nonsingular with  $k$  negative eigenvalues. If the subspace  $\mathcal{S}$  with dimension  $k$  is  $P$ -negative then  $\mathcal{S}^\perp$  is  $P$ -positive.



## Dual Robust Synthesis Inequalities

Exists controller guaranteeing robust stability if exist  $v$ ,  $\tilde{Q}$ ,  $\tilde{R}$ ,  $\tilde{S}$ :

$$\begin{pmatrix} 0 \\ I \end{pmatrix}^T \tilde{P} \begin{pmatrix} 0 \\ I \end{pmatrix} \succ 0, \quad \begin{pmatrix} -I \\ \Delta(\delta^k) \end{pmatrix}^T \tilde{P} \begin{pmatrix} -I \\ \Delta(\delta^k) \end{pmatrix} \prec 0, \quad k = 1, \dots, N$$

$$\mathbf{X}(v) \succ 0, \quad \begin{pmatrix} \mathbf{A}(v)^T & \mathbf{C}(v)^T \\ -I & 0 \\ \mathbf{B}(v)^T & \mathbf{D}(v)^T \\ 0 & -I \end{pmatrix}^T \begin{pmatrix} 0 & I & 0 & 0 \\ I & 0 & 0 & 0 \\ 0 & 0 & \tilde{Q} & \tilde{S} \\ 0 & 0 & \tilde{S}^T & \tilde{R} \end{pmatrix} \begin{pmatrix} \mathbf{A}(v)^T & \mathbf{C}(v)^T \\ -I & 0 \\ \mathbf{B}(v)^T & \mathbf{D}(v)^T \\ 0 & -I \end{pmatrix} \succ 0.$$

Note that we use the partition  $\tilde{P} = \begin{pmatrix} \tilde{Q} & \tilde{S} \\ \tilde{S}^T & \tilde{R} \end{pmatrix}$ .

No progress in general.

## Dual Robust Synthesis Inequalities

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Note that we use the partition  $\tilde{P} = \begin{pmatrix} \tilde{Q} & \tilde{S} \\ \tilde{S}^T & \tilde{R} \end{pmatrix}$ .

No progress in general. However **it helps** for state-feedback synthesis!

# Lucky Case: Static State-Feedback Synthesis

Recall block substitution:

$$\begin{pmatrix} \mathbf{A}(v) & \mathbf{B}(v) \\ \mathbf{C}(v) & \mathbf{D}(v) \end{pmatrix} = \begin{pmatrix} \mathbf{A}Y + \mathbf{B}M & B_1 \\ \mathbf{C}_1Y + \mathbf{E}M & D_1 \end{pmatrix}, \quad \mathbf{X}(v) = Y.$$

Last column does not depend on  $v$  ...

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Last column does not depend on  $v$  ...

... dual inequalities are **affine** in all variables ...

... robust state-feedback synthesis possible with LMI's!

Extends to robust performance specification in straightforward fashion!

## Dual Robust Synthesis Inequalities: State-feedback

Exists state-feedback controller guaranteeing robust stability if there exist  $Y$ ,  $M$ ,  $\tilde{Q}$ ,  $\tilde{R}$ ,  $\tilde{S}$  satisfying

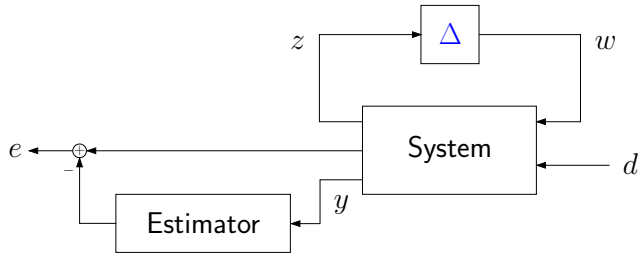
$$\begin{pmatrix} 0 \\ I \end{pmatrix}^T \tilde{P} \begin{pmatrix} 0 \\ I \end{pmatrix} \succ 0, \quad \begin{pmatrix} -I \\ \Delta(\delta^k) \end{pmatrix}^T \tilde{P} \begin{pmatrix} -I \\ \Delta(\delta^k) \end{pmatrix} \prec 0, \quad k = 1, \dots, N$$

$$Y \succ 0, \quad \begin{pmatrix} * \\ * \\ * \\ * \end{pmatrix}^T \left( \begin{array}{cc|cc} 0 & I & 0 & 0 \\ I & 0 & 0 & 0 \\ \hline 0 & 0 & \tilde{Q} & \tilde{S} \\ 0 & 0 & \tilde{S}^T & \tilde{R} \end{array} \right) \begin{pmatrix} (AY + BM)^T & (C_1Y + EM)^T \\ -I & 0 \\ \hline B_1^T & D_1^T \\ 0 & -I \end{pmatrix} \succ 0.$$

Is indeed - obviously - an LMI problem!

# Lucky Case: Robust Estimator Synthesis

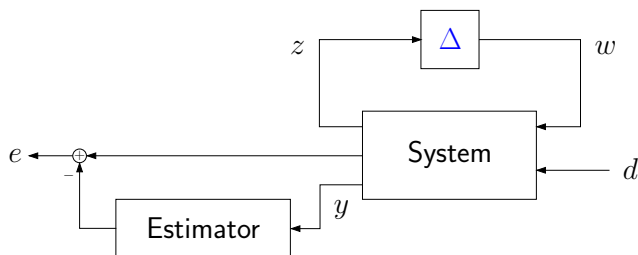
Configuration for robust estimator synthesis:





# Lucky Case: Robust Estimator Synthesis

Configuration for robust estimator synthesis:



The open-loop system with performance channel reads as

$$\begin{pmatrix} \dot{x} \\ z \\ e \\ y \end{pmatrix} = \begin{pmatrix} A & B_1 & B_2 & 0 \\ C_1 & D_1 & D_{12} & 0 \\ C_2 & D_{21} & D_2 & -I \\ C & F_1 & F_2 & 0 \end{pmatrix} \begin{pmatrix} x \\ w \\ d \\ u \end{pmatrix}.$$

# Lucky Case: Robust Estimator Synthesis

General variable substitution simplifies to

$$\mathbf{X}(v) = \begin{pmatrix} Y & I \\ I & X \end{pmatrix}$$

$$\begin{pmatrix} A(v) & B_1(v) & B_2(v) \\ C_1(v) & D_1(v) & D_{12}(v) \\ C_2(v) & D_{21}(v) & D_2(v) \end{pmatrix} =$$
$$= \left( \begin{array}{cc|cc} AY & A & B_1 & B_2 \\ K & XA + LC & XB_1 + LF_1 & XB_2 + LF_2 \\ \hline C_1Y & C_1 & D_1 & D_{12} \\ C_2Y - M & C_2 - NC & D_{21} - NF_1 & D_2 - NF_2 \end{array} \right)$$

## Lucky Case: Robust Estimator Synthesis

Robust  $\mathcal{L}_2$ -gain estimator synthesis: Multiplier constraints and LMIs

$$\begin{pmatrix} Y & I \\ I & X \end{pmatrix} \succ 0$$

$$* \begin{pmatrix} 00I0 & 0 & 0 & 0 & 0 \\ 000I & 0 & 0 & 0 & 0 \\ I000 & 0 & 0 & 0 & 0 \\ 0I00 & 0 & 0 & 0 & 0 \\ \hline 0000 & Q & S & 0 & 0 \\ 0000 & S^T & R & 0 & 0 \\ \hline 0000 & 0 & 0 & -\gamma^2 I & 0 \\ 0000 & 0 & 0 & 0 & I \end{pmatrix} \begin{pmatrix} I & 0 & 0 & 0 \\ 0 & I & 0 & 0 \\ AY & A & B_1 & B_2 \\ K & XA + LC & XB_1 + LF_1 & XB_2 + LF_2 \\ \hline 0 & 0 & I & 0 \\ C_1Y & C_1 & D_1 & D_{12} \\ \hline 0 & 0 & 0 & I \\ C_2Y - M & C_2 - NC & D_{21} - NF_1 & D_2 - NF_2 \end{pmatrix} \succ 0$$

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**Non-convex.**

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**Non-convex.** Congruence trafos with  $\text{diag}(Y^{-1}, I)$ ,  $\text{diag}(Y^{-1}, I, I, I) \dots$

# Lucky Case: Robust Estimator Synthesis

... leads to

$$\begin{pmatrix} Y^{-1} & Y^{-1} \\ Y^{-1} & X \end{pmatrix} \prec 0$$

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## Lucky Case: Robust Estimator Synthesis

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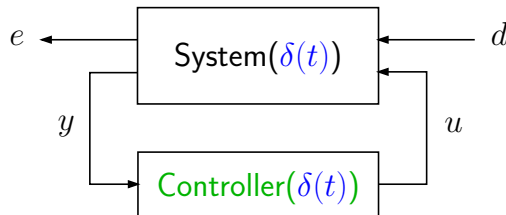
$$\begin{pmatrix} Y^{-1} & Y^{-1} \\ Y^{-1} & X \end{pmatrix} \succ 0$$

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which is **convex** in new variables  $\hat{Y} = Y^{-1}$ ,  $\hat{K} = KY^{-1}$ ,  $\hat{M} = MY^{-1}$ !

# Summary and Comments

- Identified trouble in output-feedback synthesis.
- Discussed lucky cases for robust synthesis by LMIs.
- **Gain-scheduling synthesis:** The controller is allowed to adapt itself according to on-line measurement parameters:



Output-feedback synthesis **can be transformed into LMIs.**



# Outline

- Nonsingularity of Matrix Families
- LTI Robust Stability Analysis
- Time-Varying Uncertainties
- Robust Stability and Performance with Multipliers
- Controller Synthesis
- A Glimpse at Nonlinear Uncertainties and IQCs

# Nonlinear Uncertainties

Consider the system

$$\begin{pmatrix} \dot{x}(t) \\ e(t) \\ z(t) \end{pmatrix} = \begin{pmatrix} A & B_1 & B_2 \\ C_1 & D_1 & D_{12} \\ C_2 & D_{21} & \mathbf{0} \end{pmatrix} \begin{pmatrix} x(t) \\ d(t) \\ w(t) \end{pmatrix}, \quad w(t) = \Delta(z(t)).$$

which involves the (smooth) **nonlinear uncertainty**  $\Delta : \mathbb{R}^{n_z} \rightarrow \mathbb{R}^{n_w}$ .

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The statement on slide 29 persists to hold if replacing (POS) by

$$\begin{pmatrix} \Delta(z) \\ z \end{pmatrix}^T P \begin{pmatrix} \Delta(z) \\ z \end{pmatrix} \geq 0 \quad \text{for all vectors } z \in \mathbb{R}^{n_z}.$$

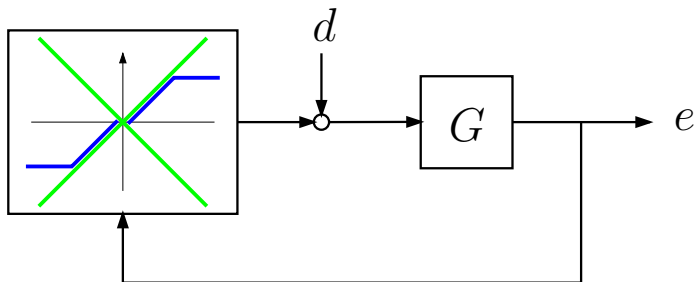
**Proof.** Literally as before!

## IQCs: Example

For transfer function

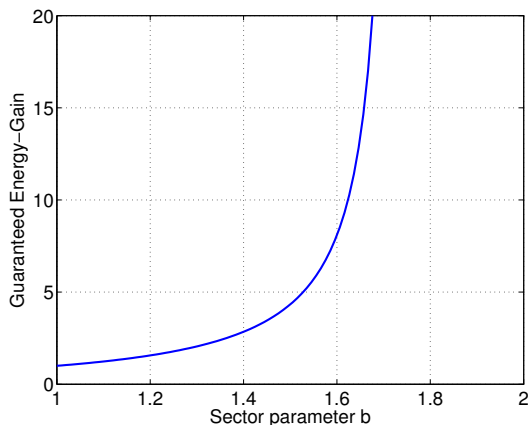
$$G(s) = \frac{-1}{2(s+1)(\frac{1}{2}s+1)(\frac{1}{3}s+1)}$$

consider the following interconnection with **saturation nonlinearity**:



Compute a good bound on the energy-gain of  $d \rightarrow e$ .

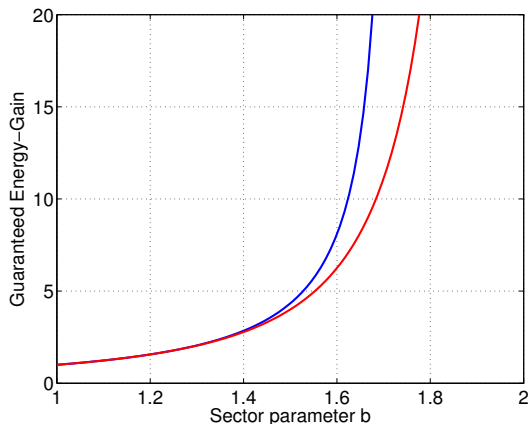
## IQCs: Example



Saturation nonlinearity with gain  $b$  satisfies

$$|\Delta(z)| \leq b|z| \quad \text{or} \quad \begin{pmatrix} \Delta(z) \\ z \end{pmatrix}' \begin{pmatrix} -1 & 0 \\ 0 & b^2 \end{pmatrix} \begin{pmatrix} \Delta(z) \\ z \end{pmatrix} \geq 0$$

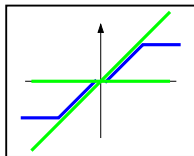
## IQCs: Example



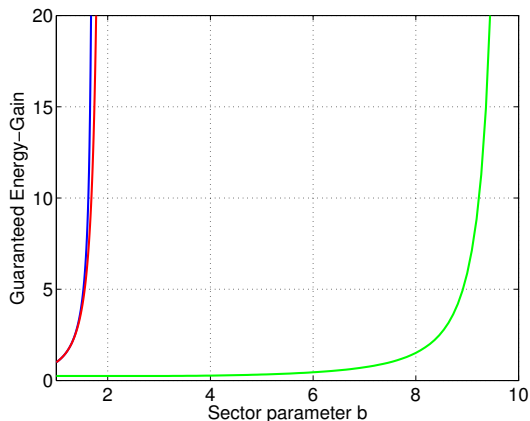
Introduce **multiplier** to reduce conservatism:

$$\begin{pmatrix} \Delta(z) \\ z \end{pmatrix}' \begin{pmatrix} -\tau & 0 \\ 0 & \tau b^2 \end{pmatrix} \begin{pmatrix} \Delta(z) \\ z \end{pmatrix} \geq 0 \quad \text{for all } \tau \geq 0$$

## IQCs: Example



$$\Delta(z)^2 \leq b z \Delta(z)$$

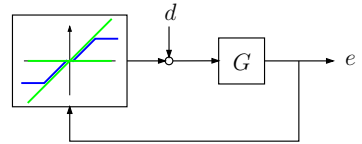


Refined information about saturation:

$$\begin{pmatrix} \Delta(z) \\ z \end{pmatrix}' \begin{pmatrix} -2\tau & b\tau \\ b\tau & 0 \end{pmatrix} \begin{pmatrix} \Delta(z) \\ z \end{pmatrix} \geq 0 \quad \text{for all } \tau \geq 0$$



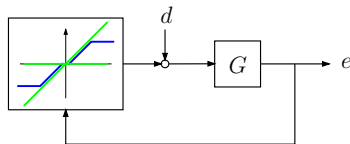
# Integral Quadratic Constraints



For any  $\tau = (\tau_1, \tau_2)$  (elementwise) define the **dynamic** multiplier

$$\Pi_{\tau}(s) = \tau_1 \begin{pmatrix} -2 & b \\ b & 0 \end{pmatrix} + \tau_2 \begin{pmatrix} 0 & \frac{s}{s+100} \\ \frac{-s}{-s+100} & 0 \end{pmatrix}.$$

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Saturation satisfies **Integral Quadratic Constraint (IQC)**

$$\int_{-\infty}^{\infty} \begin{pmatrix} \lambda \widehat{\Delta}(z)(i\omega) \\ \hat{z}(i\omega) \end{pmatrix}^* \Pi_{\tau}(i\omega) \begin{pmatrix} \lambda \widehat{\Delta}(z)(i\omega) \\ \hat{z}(i\omega) \end{pmatrix} d\omega \geq 0$$

for all  $z \in \mathcal{L}_2(\mathbb{R}_+, \mathbb{R})$ ,  $\lambda \in [0, 1]$ ,  $\tau \geq 0$  (elementwise).

Dynamic (frequency-dependent) multipliers!

# Integral Quadratic Constraints

Suppose that

$$\begin{pmatrix} T_{11} & T_{12} \\ T_{21} & T_{22} \end{pmatrix} \text{ is the transfer matrix of } \begin{pmatrix} d \\ w \end{pmatrix} \rightarrow \begin{pmatrix} e \\ z \end{pmatrix}.$$

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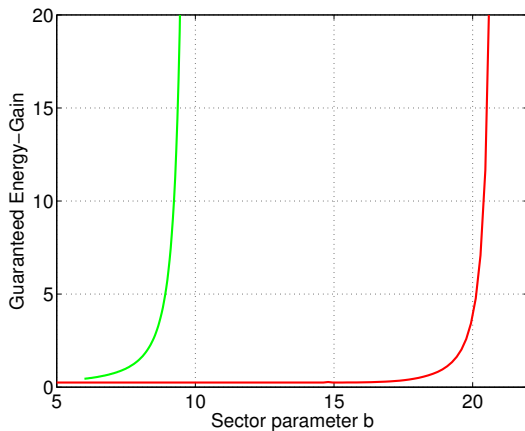
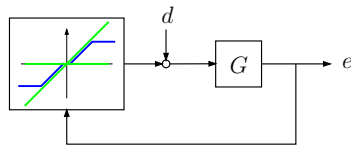
Robust stability and energy-gain performance of level  $\gamma$  is guaranteed if there exists  $\tau \geq 0$  for which the following FDI holds:

$$\begin{pmatrix} \bullet \\ \bullet \\ \bullet \\ \bullet \end{pmatrix}^* \left( \left( \begin{array}{cc|c} -\gamma^2 I & 0 & 0 \\ 0 & I & 0 \\ \hline 0 & 0 & \Pi_\tau(i\omega) \end{array} \right) \begin{pmatrix} I & 0 \\ T_{11}(i\omega) & T_{12}(i\omega) \\ 0 & I \\ T_{21}(i\omega) & T_{22}(i\omega) \end{pmatrix} \right) \prec 0 \quad \forall \omega \in \mathbb{R} \cup \{\infty\}.$$

**Computation:** Application of KYP lemma leads to LMI feasibility test.

A. Megretski, A. Rantzer, System analysis via Integral Quadratic Constraints, IEEE Trans. Autom. Contr. 42 (1997) 819-830.

## IQCs: Example



Dynamics are highly beneficial!

# Main Points

Here is a summary of the **main issues** we addressed:

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