



## INTERCONNECTED SYSTEMS

# TEARING and ZOOMING CONTROL

Chaire Francqui, Lecture V, May 19, 2004

**UCL** Université catholique de Louvain

# Road Map

**What is a dynamical system ?**

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## What is a dynamical system ?

Historically:

● 'Closed' systems  $\rightsquigarrow$   $\frac{d}{dt}x = f(x)$  **very limited**

● Input/output map?

● Input/state/output structure ?

**Assumes I/O partition. Possible? Obtainable? How? Needed?**

● **Other possibilities? CS? Graph theory? Object oriented modeling?**

# Road Map

What is a dynamical system ?

What is a mathematical model, really?  $\rightsquigarrow$

Dynamical system :=  $(\mathbb{T}, \mathbb{W}, \mathfrak{B})$  with  $\mathfrak{B} \subseteq (\mathbb{W})^{\mathbb{T}}$  the 'behavior'.

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Behavioral eq'ns contain **latent variables**

$\rightsquigarrow$  elimination thms, algorithms.

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### Basic notions

- Controllability  $\rightsquigarrow$  image representation
- Observability
- (Dissipative systems)
- (Stability)
- State  $\rightsquigarrow$  state representation algorithms

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### Basic problems

● Modeling from data (system ID)

● Modeling by interconnecting components

● Control (= interconnection)  $\rightsquigarrow$  LQ,  $\mathcal{H}_2$ ,  $\mathcal{H}_\infty$

● ...

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### Projects

- 🔴 N-d systems and PDE's (Rocha, Shankar, Pillai, Zerz, Oberst)
- 🔴 Software
- 🔴 Stochastic systems
- 🔴 ...



# THEME

## 1. **Modeling by tearing and zooming**

- **General ideas**
- **Terminals**
- **Modules**
- **Interconnection architecture**
- **Examples**
- **RTCT circuits**

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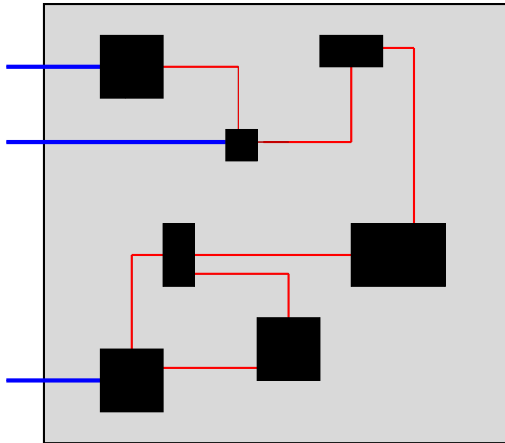
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## 3. Conclusions

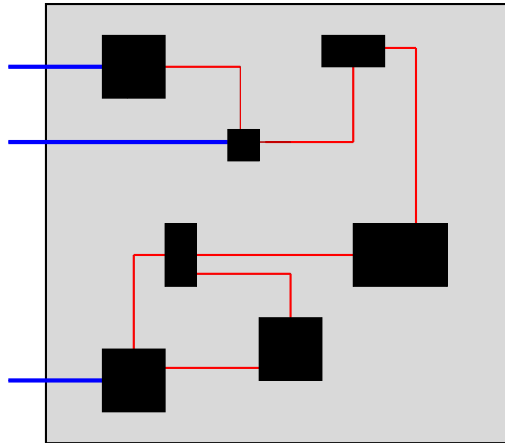
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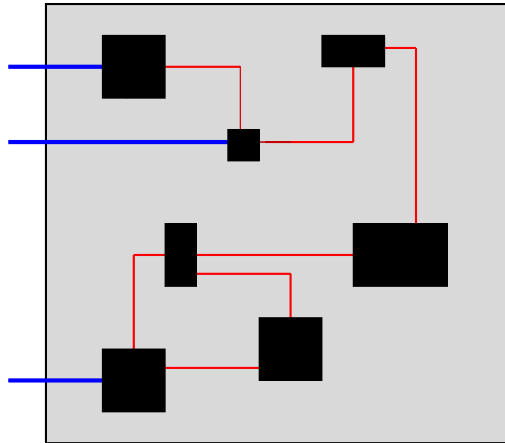
When systems are interconnected, what really happens?

How do we obtain a model from

the **components** and the **interconnections**?

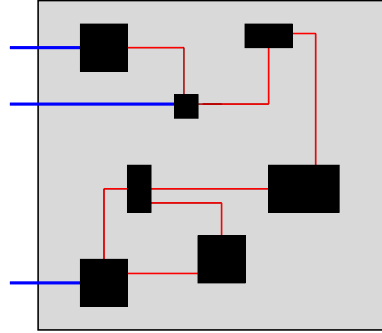
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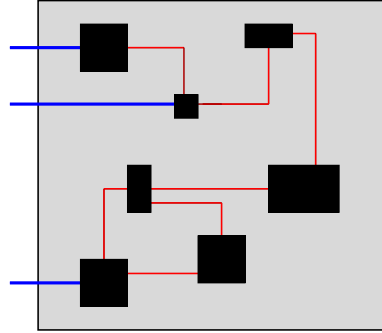
**'Tearing'** the system into subsystems,  
and, in order to model, **'zooming'** on the individual subsystems

# TEARING and ZOOMING



**The ingredients of the language and methodology that we propose:**

# TEARING and ZOOMING

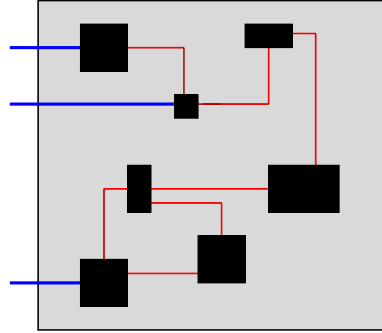


The ingredients of the language and methodology that we propose:

1. **Modules** : the subsystems



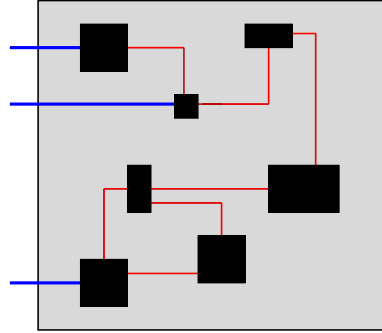
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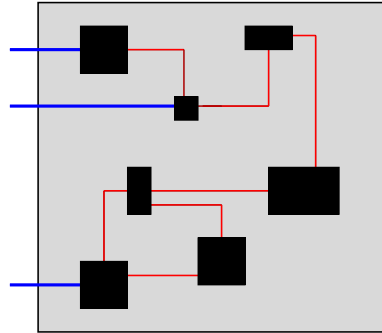
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the layout of the modules and their interconnection

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The ingredients of the language and methodology that we propose:

1. **Modules** : the subsystems
2. **Terminals** : the physical links between subsystems
3. The **interconnection architecture** :  
the layout of the modules and their interconnection
4. The **manifest variable assignment** :  
which variables does the model aim at?

# TEARING and ZOOMING

## Features:

- **Reality** — ‘physics’ — **based**
- **Uses behavioral systems concepts**  
more akin to bond-graphs and across/through variables,  
than to input/output thinking.
- **Hierarchical:** allows new systems to be build from old
- Models are **reusable, generalizable & extendable**
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System theory with its inputs and outputs and signal flow graphs,  
as implemented e.g. in **SIMULINK**® is hopelessly inadequate.  
**MODELICA**® is much better.

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Example:

An **electrical terminal (type)**  
implies a **(voltage, current) pair** of real **terminal** variables.



# Examples

Type of terminal	Variables	Signal space
electrical	(voltage, current)	$\mathbb{R}^2$
mechanical (1-D)	(force, position)	$\mathbb{R}^2$
mechanical (2-D)	((position, attitude), (force, torque))	$(\mathbb{R}^2 \times S^1)$ $\times (\mathbb{R}^2 \times T^* S^1)$
mechanical (3-D)	((position, attitude), (force, torque))	$(\mathbb{R}^2 \times S^2)$ $\times (\mathbb{R}^2 \times T^* S^2)$
thermal	(temp., heat flow)	$\mathbb{R}^2$
fluidic	(pressure, flow)	$\mathbb{R}^2$
thermal - fluidic	(pressure, temp., mass flow, heat flow)	$\mathbb{R}^4$

# Examples

Type of terminal	Variables	Signal space
chemical		
input	$u$	$U \subseteq \mathbb{R}$
output	$y$	$Y \subseteq \mathbb{R}$
m-dim input	$(u_1, u_2, \dots, u_m)$	$U \subseteq \mathbb{R}^m$
p-dim output	$(y_1, y_2, \dots, y_p)$	$Y \subseteq \mathbb{R}^p$
etc.	etc.	etc.

# MODULES

A **module** is specified by

- its *type*,
- its *parametrization*,
- and its *parameter values*.

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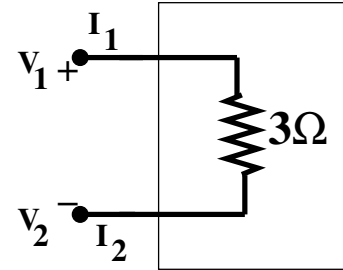
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The idea is the following.

By specifying the **module type**, we give the variables living on its terminals. We want a fully automated way of specifying the **behavior** of these variables. This typically happens by specifying some **parameters**, and a map, **the parametrization**, which maps these parameters into the correct behavior.

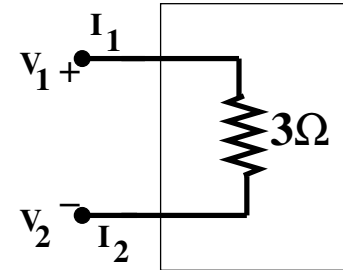
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Example: The module is a 3 Ohm resistor:



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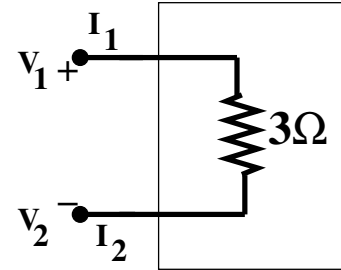
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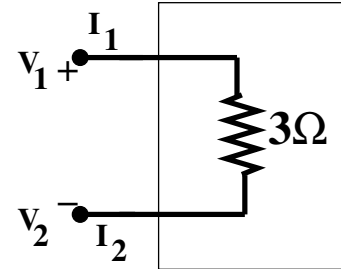
This means that it has **two electrical terminals**

$\rightsquigarrow$  **terminal variables**  $((V_1, I_1), (V_2, I_2))$ .

The possible behaviors form a family of two-dimensional linear subspaces of  $\mathbb{R}^4$ .

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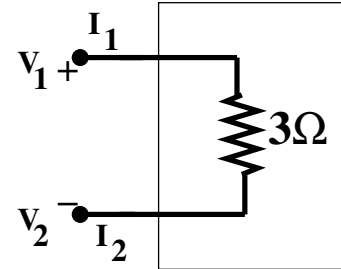
The **resistance parametrization** is the map from  $R \in [0, \infty)$  into the behavioral eq'ns

$$V_1 - V_2 = R I_1, \quad I_1 + I_2 = 0.$$



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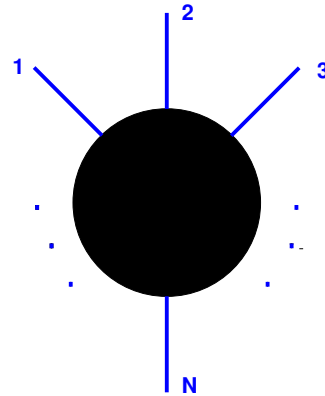
The **parameter value** equals **3**, via the parametrization  $\rightsquigarrow$

$$V_1 - V_2 = 3I_1, \quad I_1 + I_2 = 0.$$

# Module type

The **module type** specifies an **ordered set of terminals**

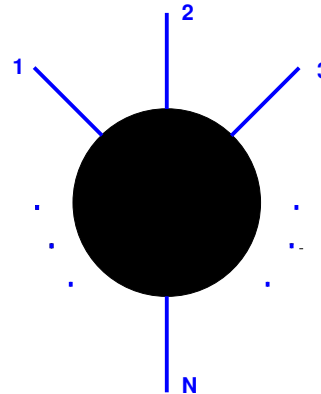
$(t_1, t_2, \dots, t_N)$ .



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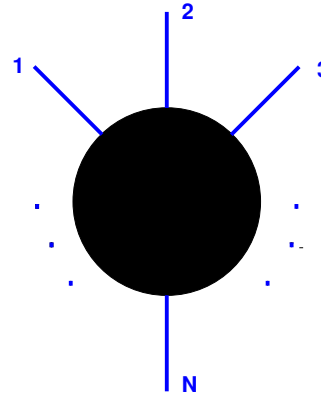
$((w_{t_1,1}, w_{t_1,2}, \dots), \dots, (w_{t_N,1}, w_{t_N,2}, \dots))$

taking values in the product space of the terminal signal spaces.

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taking values in the product space of the terminal signal spaces.

The **module type** also specifies a set  $\mathbb{B}$

of **possible behaviors** of the terminal variables of the module.

# Parametrization

We assume that the **module** is further specified by

a **parametrization** of  $\mathbb{B}$ ,

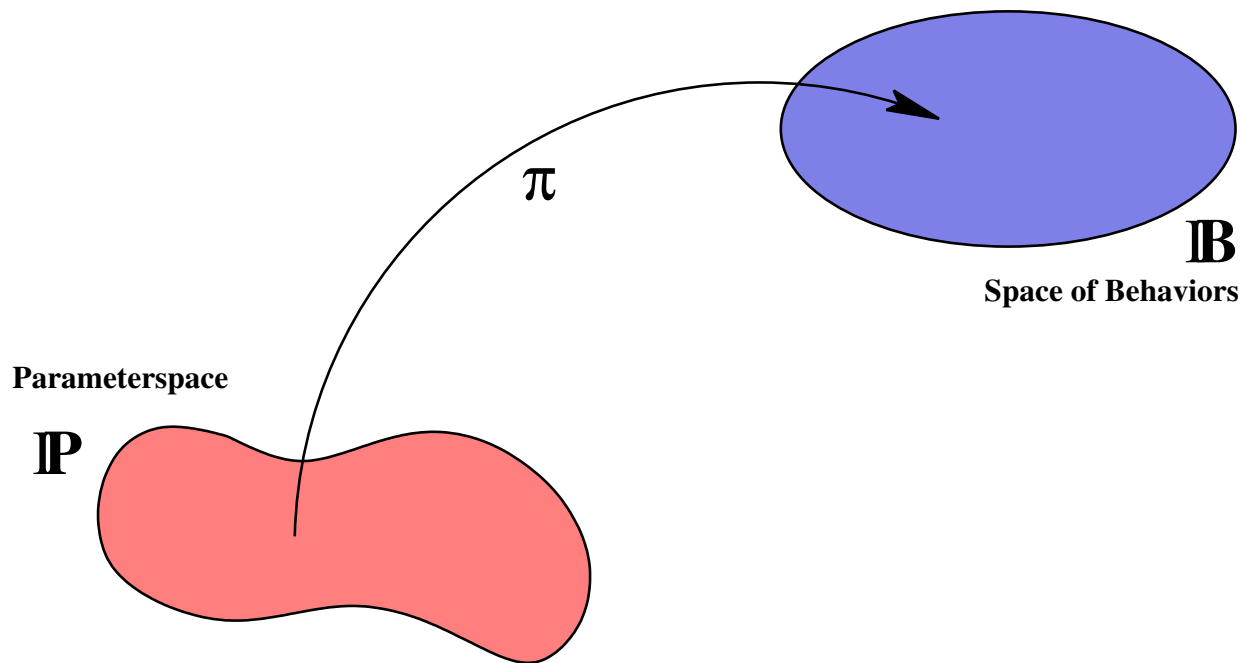
that is, a surjective map  $\pi$  from a **parameterspace**  $\mathbb{P}$  into the **space of behaviors**  $\mathbb{B}$ .

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$\mathbb{P}$  is typically a combination of a set of integers and real numbers.

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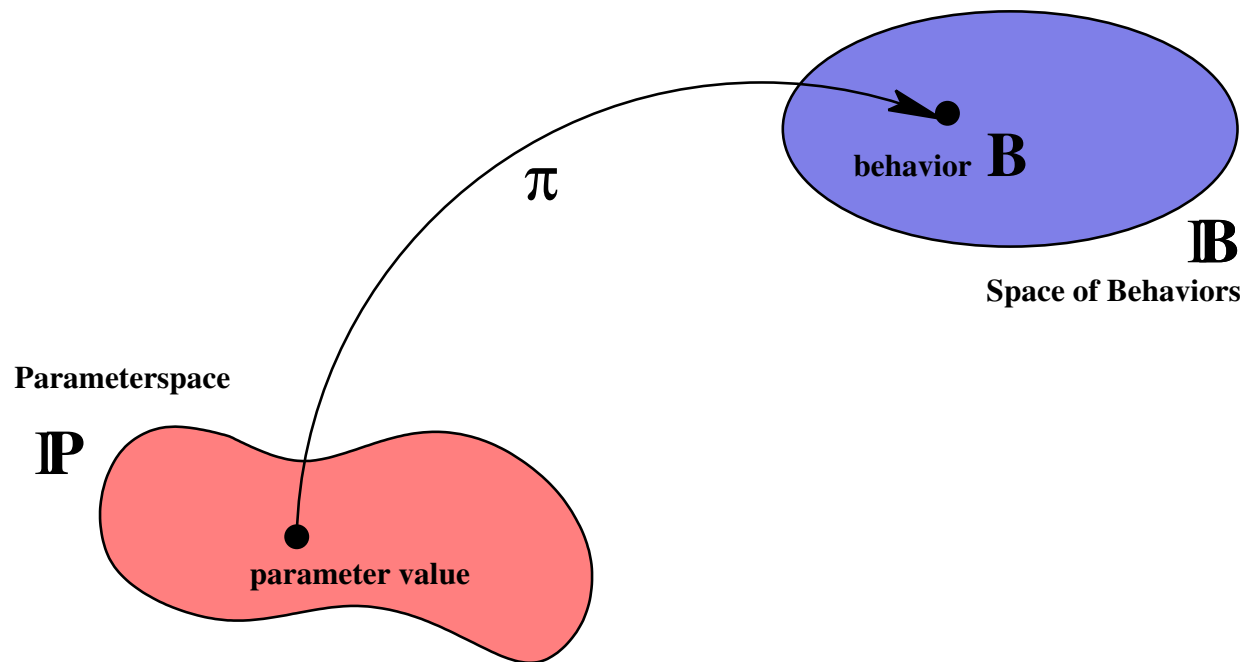
A module is further specified by giving **the value of the parameters**.

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# MODULES

By specifying a module, we thus obtain the *behavior* of the variables

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on the *terminals of the module.*

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By specifying a module, we thus obtain the *behavior* of the variables

$$(w_1, w_2, \dots, w_n)$$

on the *terminals of the module*.

This way we obtain a dynamic model of the interaction of the module with its environment.

# Examples

## ELECTRICAL MODULES

Module type	Parametrization	Parameter value
<b>2-terminal Ohmic resistor</b>	<b>resistance</b> $\pi : \mathbb{R}_+ \rightarrow \dots$	$R$ in ohms
<b>2-terminal Ohmic resistor</b>	<b>conductance</b> $\pi : \mathbb{R}_+ \rightarrow \dots$	$G$ in mhos
<b>2-terminal current driven resistor</b>	<b>all maps:</b> $\mathbb{R} \rightarrow \mathbb{R}$	$\rho : \mathbb{R} \rightarrow \mathbb{R}$
<b>capacitor</b>	<b>capacitance</b> $\pi : \mathbb{R}_+ \rightarrow \dots$	$C$ in farads
<b>inductor</b>	<b>inductance</b> $\pi : \mathbb{R}_+ \rightarrow \dots$	$L$ in henrys

# Examples

Module type	Parametrization domain	Parameter value
linear impedances	$\mathbb{N}$ (number of ports) $\times \mathbb{R}^{n \times n}(\xi)$	$Z \in \mathbb{R}^{n \times n}[\xi]$
resistive $\triangle$	$\mathbb{R}$	$R$ in ohms
Y with linear diff. systems	$(\mathbb{R}^2[\xi])^3$	$(R_1, R_2, R_3)$ $\in \mathbb{R}^{1 \times 2}[\xi]$
transformer	$\mathbb{R}$	$n \in \mathbb{R}$
transmission line	$(\mathbb{R}_+)^5$	$L, \ell, c, r_s, r_p$
transistor		
etc.	etc.	etc.

# Examples

## MECHANICAL MODULES

<b>Module type</b>	<b>Parametrization</b>	<b>Parameters</b>
mass	$\pi : \mathbb{R}_+ \rightarrow \dots$	$m$ in kg
solid bar	length, mass/unit length $\pi : \mathbb{R}_+ \times \mathbb{R}_+ \rightarrow \dots$	$L, m$
spring		
damper		
multi-terminal mass		geometry
flexible bar		
etc.	etc.	etc.

# Examples

## OTHER DOMAINS

<b>Module type</b>	<b>Parametrization</b>	<b>Parameters</b>
servo joint		$m_r, m_s, J_r, J_s,$ $L, R, K$
2 inlet tank		geometry
etc.	etc.	etc.

# Examples

## LINEAR SYSTEMS

Module type	Parametrization	Parameters
$\Sigma \in \mathcal{L}^\bullet$	$\mathbb{N} \times \{\text{ker, im, etc.}\} \times \mathbb{R}^{\bullet \times \bullet}[\xi], \text{ or } \dots$	$(w, \text{ker}, R \in \mathbb{R}^{\bullet \times w}[\xi])$ ...
$\Sigma \in \mathcal{L}_{\text{cont}}^\bullet$	$\mathbb{N} \times \{\text{im}, \dots\}$	$(w, M \in \mathbb{R}^{w \times \bullet}[\xi]),$ ...
$\Sigma \in \mathcal{L}_{\text{cont}}^{\text{i/o}}$	$\mathbb{N} \times \mathbb{N} \times \{\text{tf. f'n.,} \dots\} \times \mathbb{R}^{\bullet \times \bullet}(\xi), \dots$	$m, p, G \in \mathbb{R}^{p \times m}[\xi]$ ...
$\Sigma \in \mathcal{L}^{\text{i/s/o}}$	$\mathbb{N}^3, \dots$	$m, n, p, (A, B, C, D)$
etc.	etc.	etc.

# INTERONNECTION ARCHITECTURE

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The **interconnection architecture** is a set of *terminal pairs* (unordered, disjoint, and with distinct elements), denoted by  $\mathbb{I}$ .

If  $\{t_i, t_j\} \in \mathbb{I}$ , then we say that these terminals are **connected**.



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We impose that connected terminals must be **adapted**.

In the case of **physical terminals**, this means that they must be of the **same type** (both electrical, 2-D mechanical, thermal, etc.).

In the case of **logical terminals** (input or output terminals), this means that if one of the connected terminals is an  $m$ -dimensional input terminal, the other must be an  $m$ -dimensional output terminal.

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Pair of terminals	Terminal 1	Terminal 2	Interconnection law
electrical	$(V_1, I_1)$	$(V_2, I_2)$	$V_1 = V_2, I_1 + I_2 = 0$
1-D mech.	$(F_1, q_1)$	$(F_2, q_2)$	$F_1 + F_2 = 0, q_1 = q_2$
2-D mech.			
thermal	$(Q_1, T_1)$	$(Q_2, T_2)$	$Q_1 + Q_2 = 0, T_1 = T_2$
fluidic	$(p_1, f_1)$	$(p_2, f_2)$	$p_1 = p_2, f_1 + f_2 = 0$
info processing	m-input $u$	m-output $y$	$u = y$
etc.	etc.	etc.	etc.

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The model unavoidably contains many other variables. These **latent variables** could be

either

**interconnection variables,**

or

**latent variables used to describe the behavior of the modules.**



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- A set of modules  $M_1, M_2, \dots, M_m$

so, for each module,

the **type**, the **parametrization**, and **parameter value**.

This yields a list of terminals  $T = \{t_1, t_2, \dots, t_{|T|}\}$

and the behavior  $\mathcal{B}_i, i = 1, \dots, m$ , for the terminal variables.

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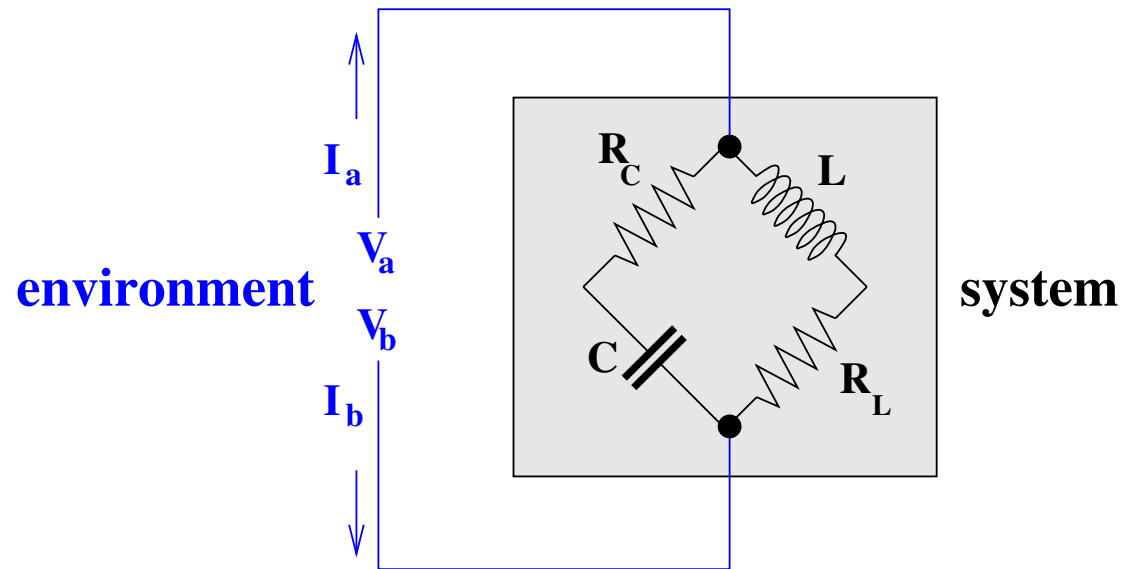
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- The manifest variable assignment.
- The yields  $\mathcal{B}' \cap \mathcal{B}'' =$  the **full behavior**  
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- Elimination of latent variables  $\rightarrow$  the manifest behavior  $\mathcal{B}$ .

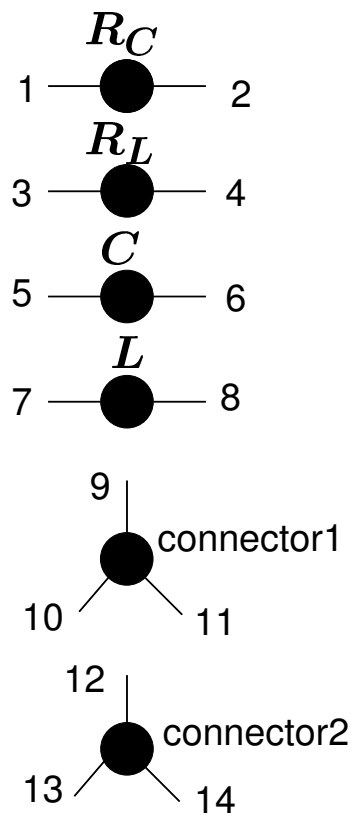
# Examples

## RLC circuit



# RLC circuit

## TEARING





# RLC circuit

## ZOOMING

The list of the modules & the associated terminals:

<b>Module</b>	<b>Type</b>	<b>Terminals</b>	<b>Parameter</b>
$R_C$	resistor	(1, 2)	$R$ in ohms
$R_L$	resistor	(3, 4)	$R$ in ohms
$C$	capacitor	(5, 6)	$C$ in farad
$L$	inductor	(7, 8)	$L$ in henry
connector1	3-terminal connector	(9, 10, 11)	
connector2	3-terminal connector	(12, 13, 14)	

## The interconnection architecture:

Pairing
{10, 1}
{11, 7}
{2, 5}
{8, 3}
{6, 13}
{4, 14}

## RLC circuit

### Manifest variable assignment:

the variables

$$V_9, I_9, V_{12}, I_{12}$$

on the external terminals **{9, 12}**, i.e,

$$V_a = V_9, I_a = I_9, V_b = V_{12}, I_b = I_{12}.$$

## RLC circuit

### Manifest variable assignment:

the variables

$$V_9, I_9, V_{12}, I_{12}$$

on the external terminals **{9, 12}**, i.e,

$$V_a = V_9, I_a = I_9, V_b = V_{12}, I_b = I_{12}.$$

The internal terminals are

$$\{1, 2, 3, 4, 5, 6, 7, 8, 10, 11, 13, 14\}$$

The variables on these terminals are **latent variables**.

# RLC circuit

Equations for the full behavior:

Modules	Constitutive equations	
$R_C$	$I_1 + I_2 = 0$	$V_1 - V_2 = R_C I_1$
$R_L$	$I_7 + I_8 = 0$	$V_7 - V_8 = R_L I_7$
$C$	$I_5 + I_6 = 0$	$C \frac{d}{dt} (V_5 - V_6) = I_5$
$L$	$I_7 + I_8 = 0$	$V_7 - V_8 = L \frac{d}{dt} I_7$
connector1	$I_9 + I_{10} + I_{11} = 0$	$V_9 = V_{10} = V_{11}$
connector2	$I_{12} + I_{13} + I_{14} = 0$	$V_{12} = V_{13} = V_{14}$

## RLC circuit

<b>Interconnection pair</b>	<b>Interconnection equations</b>	
{10, 1}	$V_{10} = V_1$	$I_{10} + I_1 = 0$
{11, 7}	$V_{11} = V_7$	$I_{11} + I_7 = 0$
{2, 5}	$V_2 = V_5$	$I_2 + I_5 = 0$
{8, 3}	$V_8 = V_3$	$I_8 + I_3 = 0$
{6, 13}	$V_6 = V_{13}$	$I_6 + I_{13} = 0$
{4, 14}	$V_4 = V_{14}$	$I_4 + I_{14} = 0$

## RLC circuit

All these eq'ns combined define a latent variable system in the manifest variables

$$w = (V_a, I_a, V_b, I_b)$$

with latent variables

$$\ell = (V_1, I_1, V_2, I_2, V_3, I_3, V_4, I_4, V_5, I_5, V_6, I_6, V_7, I_7, V_8, I_8, V_{10}, I_{10}, V_{11}, I_{11}, V_{13}, I_{13}, V_{14}, I_{14}).$$

## RLC circuit

All these eq'ns combined define a latent variable system in the manifest variables

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with latent variables

$$\ell = (V_1, I_1, V_2, I_2, V_3, I_3, V_4, I_4, V_5, I_5, V_6, I_6, V_7, I_7, V_8, I_8, V_{10}, I_{10}, V_{11}, I_{11}, V_{13}, I_{13}, V_{14}, I_{14}).$$

The manifest behavior  $\mathfrak{B}$  is given by

$$\mathfrak{B} = \{(V_a, I_a, V_b, I_b) : \mathbb{R} \rightarrow \mathbb{R}^4 \mid \exists \ell : \mathbb{R} \rightarrow \mathbb{R}^{24} \dots\}$$



## RLC circuit

**Elimination:** for example, using Gröbner bases.

## RLC circuit

**Elimination:** for example, using Gröbner bases.

Case 1:  $CR_C \neq \frac{L}{R_L}$ .

$$\begin{aligned} & \left( \frac{R_C}{R_L} + \left(1 + \frac{R_C}{R_L}\right) CR_C \frac{d}{dt} + CR_C \frac{L}{R_L} \frac{d^2}{dt^2} \right) (V_a - V_b) \\ & = \left(1 + CR_C \frac{d}{dt}\right) \left(1 + \frac{L}{R_L} \frac{d}{dt}\right) R_C I_a. \end{aligned}$$

$$I_a + I_b = 0$$

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$$I_a + I_b = 0$$

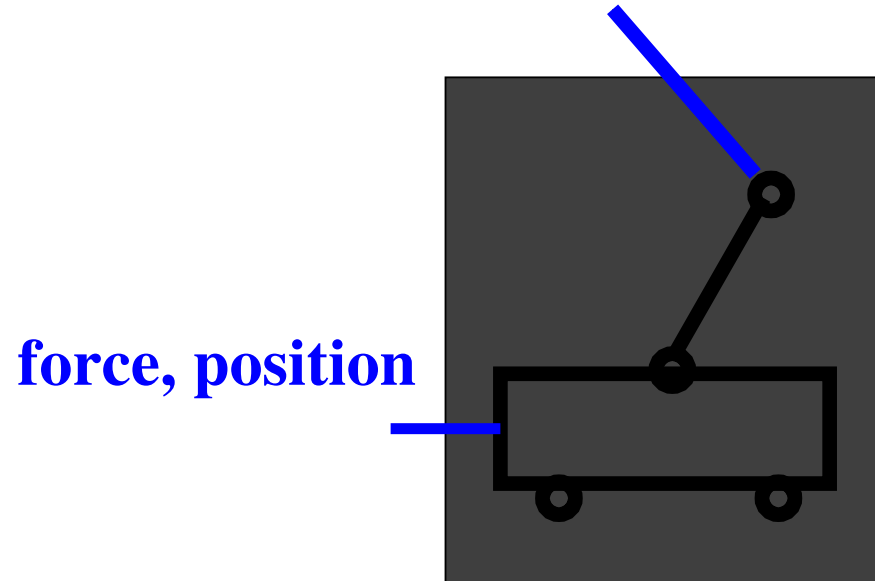
**Case 2:**  $CR_C = \frac{L}{R_L}$ .

$$\left( \frac{R_C}{R_L} + CR_C \frac{d}{dt} \right) (V_a - V_b) = \left(1 + CR_C \frac{d}{dt}\right) R_C I_a$$

$$I_a + I_b = 0$$

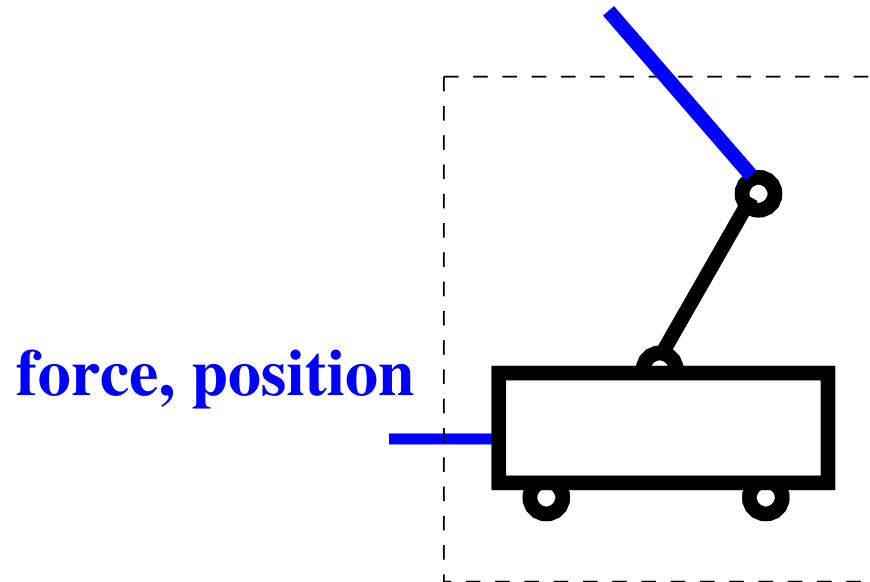
# CART

force, position, torque, angle



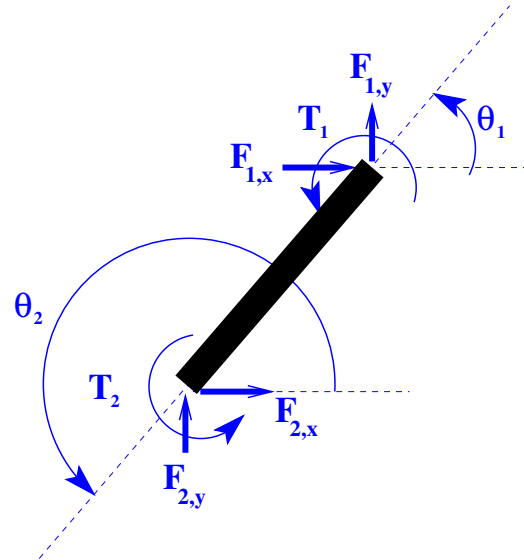
# CART

force, position, torque, angle



Required modules: Solid bars, cart, servo's.

# CART



Solid bar

Terminals: 2 mechanical 2-D terminals.

Parameters:

$L \in \mathbb{R}_+$  (length),  $m \in \mathbb{R}_+$  (mass per unit length).

## Behavioral equations:

$$mL \frac{d^2}{dt^2} x_c = F_{x_1} + F_{x_2},$$

$$mL \frac{d^2}{dt^2} y_c = F_{y_1} + F_{y_2} - mLg,$$

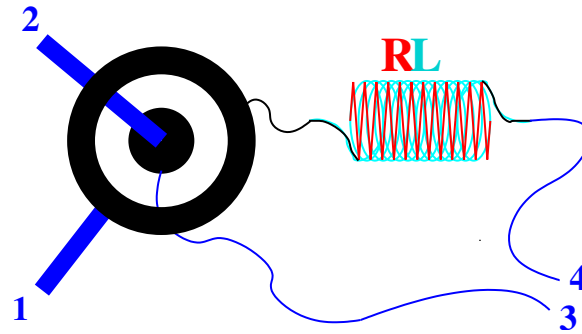
$$m \frac{L^3}{12} \frac{d^2}{dt^2} \theta_c = T_1 + T_2 - \frac{L}{2} F_{x_1} \sin(\theta_1) \\ + \frac{L}{2} F_{y_1} \cos(\theta_1) - \frac{L}{2} F_{x_2} \sin(\theta_2) + \frac{L}{2} F_{y_2} \cos(\theta_2),$$

$$\theta_1 = \theta_c, \theta_2 = \theta_1 + \pi,$$

$$x_1 = x_c + \frac{L}{2} \cos(\theta_c), x_2 = x_c - \frac{L}{2} \cos(\theta_c),$$

$$y_1 = y_c + \frac{L}{2} \sin(\theta_c), y_2 = y_c - \frac{L}{2} \sin(\theta_c).$$

Note: Contains latent variables  $x_c, \theta_c$ .



Hinge with servo

Terminals: 2 mechanical 2-D terminals, 2 electrical.

Parameters:

rotor mass  $m_r$ , the stator mass  $m_s$ , the rotor inertia  $J_r$ , the stator inertia  $J_s$ , the inductance  $L$ , the resistance  $R$  of the motor circuit, the motor torque constant  $K$ .



# CART

## Behavioral equations:

$$(m_r + m_s) \frac{d^2}{dt^2} x_1 = F_{x_1} + F_{x_2}$$

$$(m_r + m_s) \frac{d^2}{dt^2} y_1 = F_{x_1} + F_{x_2}$$

$$J_r \frac{d^2}{dt^2} \theta_1 = T_1 + T_m$$

$$J_s \frac{d^2}{dt^2} \theta_2 = T_2 - T_m$$

$$V_3 - V_4 = L \frac{d}{dt} I_3 + R I_3 + K \frac{d}{dt} (\theta_1 - \theta_2)$$

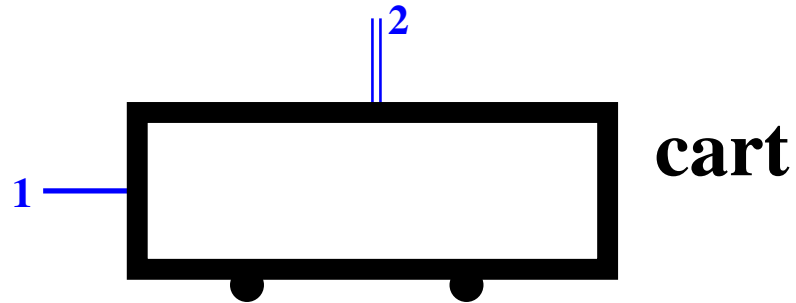
$$K I_3 = T_m, I_3 = -I_4$$

$$x_1 = x_2, y_1 = y_2$$

Terminal variables:  $(x_1, y_1, \theta_1, F_{x_1}, F_{y_1}, T_1,$   
 $x_2, y_2, \theta_2, F_{x_2}, F_{y_2}, T_2, V_3, I_4, V_4, I_4).$

The motor torque  $T_m$  is a latent variable.

# CART



**Terminals:** 1 mechanical 1-D terminal, 1 mechanical 2-D terminal.

**Parameters:** mass  $M$ .

# CART

## Behavioral equations:

$$M \frac{d^2}{dt^2} x_1 = F_1 + F_{x_2}$$

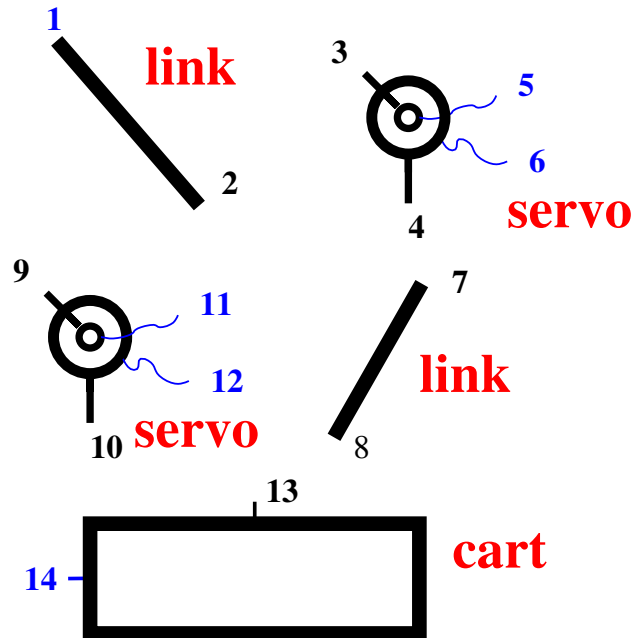
$$x_2 = x,$$

$$y_2 = 0,$$

$$\theta_2 = \pi/2$$

# CART

## TEARING



# CART

## ZOOMING

The list of the modules & the associated terminals:

Module	Type	Terminals	Parameter
Link 1	bar	(7,8)	$L_1, m_1$
Link 2	bar	(1,2)	$L_2, m_2$
Cart	cart	(13,14)	$M$
Hinge 1	servo	(9,10,11,12)	$m_{r_1}, m_{s_1}, J_{r_1}, J_{r_1}, L_1, R_1, K_1$
Hinge 2	servo	(3,4,5,6 )	$m_{r_2}, m_{s_2}, J_{r_2}, J_{r_2}, L_2, R_2, K_2$

# CART

The interconnection architecture:

Pairing
{2, 3}
{4, 7}
{8, 9}
{10, 13}

# CART

The interconnection architecture:

Pairing
{2, 3}
{4, 7}
{8, 9}
{10, 13}

Manifest variable assignment:

the variables on the external terminals {1, 5, 6, 11, 12, 14}.

All other variables are **latent variables**.

## Equations for the full behavior:

$$m_1 L_1 \frac{d^2}{dt^2} x_{c_1} = F_{x_1} + F_{x_2},$$

$$m_1 L_1 \frac{d^2}{dt^2} y_{c_1} = F_{y_1} + F_{y_2} - m_1 L_1 g,$$

$$m_1 \frac{L_1^3}{12} \frac{d^2}{dt^2} \theta_{c_1} = T_1 + T_2 -$$

$$\frac{L_1}{2} F_{x_1} \sin(\theta_1) + \frac{L_1}{2} F_{y_1} \cos(\theta_1) - \frac{L_1}{2} F_{x_2} \sin(\theta_2) + \frac{L_1}{2} F_{y_2} \cos(\theta_2),$$

$$\theta_1 = \theta_{c_1},$$

$$\theta_2 = \theta_1 + \pi,$$

$$x_1 = x_{c_1} + \frac{L_1}{2} \cos(\theta_{c_1}),$$

$$x_2 = x_{c_1} - \frac{L_1}{2} \cos(\theta_{c_1}),$$

$$y_1 = y_{c_1} + \frac{L_1}{2} \sin(\theta_{c_1}),$$

$$y_2 = y_{c_1} - \frac{L_1}{2} \sin(\theta_{c_1}),$$



# CART

$$m_2 L_2 \frac{d^2}{dt^2} x_{c_2} = F_{x_7} + F_{x_8},$$

$$m_2 L_2 \frac{d^2}{dt^2} y_{c_2} = F_{y_7} + F_{y_8} - m_2 L_2 g,$$

$$m_2 \frac{L_2^3}{12} \frac{d^2}{dt^2} \theta_{c_2} = T_7 + T_8 - \frac{L_2}{2} F_{x_7} \sin(\theta_7) + \frac{L_2}{2} F_{y_7} \cos(\theta_7),$$
$$- \frac{L_2}{2} F_{x_8} \sin(\theta_8) + \frac{L_2}{2} F_{y_8} \cos(\theta_8),$$

$$\theta_7 = \theta_{c_2},$$

$$\theta_8 = \theta_7 + \pi,$$

$$x_7 = x_{c_2} + \frac{L_1}{2} \cos(\theta_{c_2}),$$

$$x_8 = x_{c_2} - \frac{L_1}{2} \cos(\theta_{c_2}),$$

$$y_7 = y_{c_2} + \frac{L_1}{2} \sin(\theta_{c_2}),$$

$$y_8 = y_{c_2} - \frac{L_1}{2} \sin(\theta_{c_2}),$$

# CART

$$M \frac{d^2}{dt^2} x_{14} = F_{14} + F_{x_{14}}$$

$$x_{14} = x_{13},$$

$$y_{13} = 0,$$

$$\theta_{13} = \pi/2,$$

# CART

$$(m_{r_1} + m_{s_1}) \frac{d^2}{dt^2} x_3 = F_{x_3} + F_{x_4},$$

$$(m_{r_1} + m_{s_1}) \frac{d^2}{dt^2} y_3 = F_{y_3} + F_{y_4},$$

$$J_{r_1} \frac{d^2}{dt^2} \theta_3 = T_3 + T_m,$$

$$J_{s_1} \frac{d^2}{dt^2} \theta_4 = T_4 - T_m,$$

$$V_5 - V_6 = L_1 \frac{d}{dt} I_5 + R_1 I_5 + K \frac{d}{dt} (\theta_3 - \theta_4),$$

$$K_1 I_5 = T_{m_1},$$

$$x_3 = x_4, y_3 = y_4,$$

$$I_5 = -I_6,$$

# CART

$$(m_{r_2} + m_{s_2}) \frac{d^2}{dt^2} x_9 = F_{x_9} + F_{x_{10}},$$

$$(m_{r_2} + m_{s_2}) \frac{d^2}{dt^2} y_9 = F_{y_9} + F_{y_{10}},$$

$$J_{r_2} \frac{d^2}{dt^2} \theta_9 = T_9 + T_m,$$

$$J_{s_2} \frac{d^2}{dt^2} \theta_{10} = T_{10} - T_m,$$

$$V_{11} - V_{12} = L_2 \frac{d}{dt} I_{11} + R_2 I_{11} + K \frac{d}{dt} (\theta_9 - \theta_{10}),$$

$$K_2 I_{11} = T_{m_2},$$

$$x_{10} = x_{11}, y_{10} = y_{11},$$

$$I_{11} = -I_{12},$$

# CART

$$F_{x_2} + F_{x_3} = 0, F_{y_2} + F_{y_3} = 0, x_2 = x_3, y_2 = y_3,$$

$$\theta_2 = \theta_3 + \pi, T_2 + T_3 = 0,$$

$$F_{x_4} + F_{x_7} = 0, F_{y_4} + F_{y_7} = 0, x_4 = x_7, y_4 = y_7,$$

$$\theta_4 = \theta_7 + \pi, T_4 + T_7 = 0,$$

$$F_{x_8} + F_{x_9} = 0, F_{y_8} + F_{y_9} = 0, x_8 = x_9, y_8 = y_9,$$

$$\theta_8 = \theta_9 + \pi, T_8 + T_9 = 0,$$

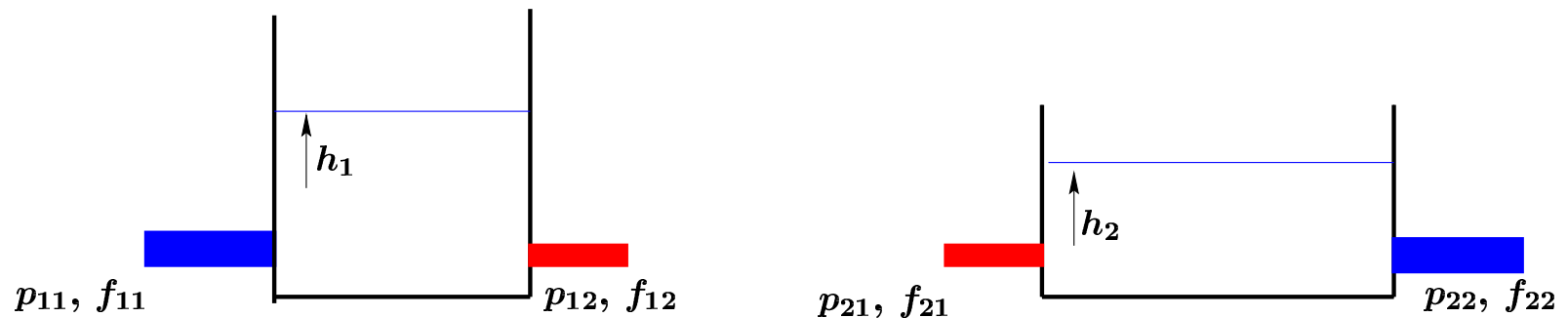
$$F_{x_{10}} + F_{x_{13}} = 0, F_{x_{10}} + F_{x_{13}} = 0,$$

$$x_{10} = x_{13}, y_{10} = y_{13}.$$

$$\theta_{10} = \theta_{13} + \pi, T_{10} + T_{13} = 0.$$

# INPUT - to - OUTPUT CONNECTIONS

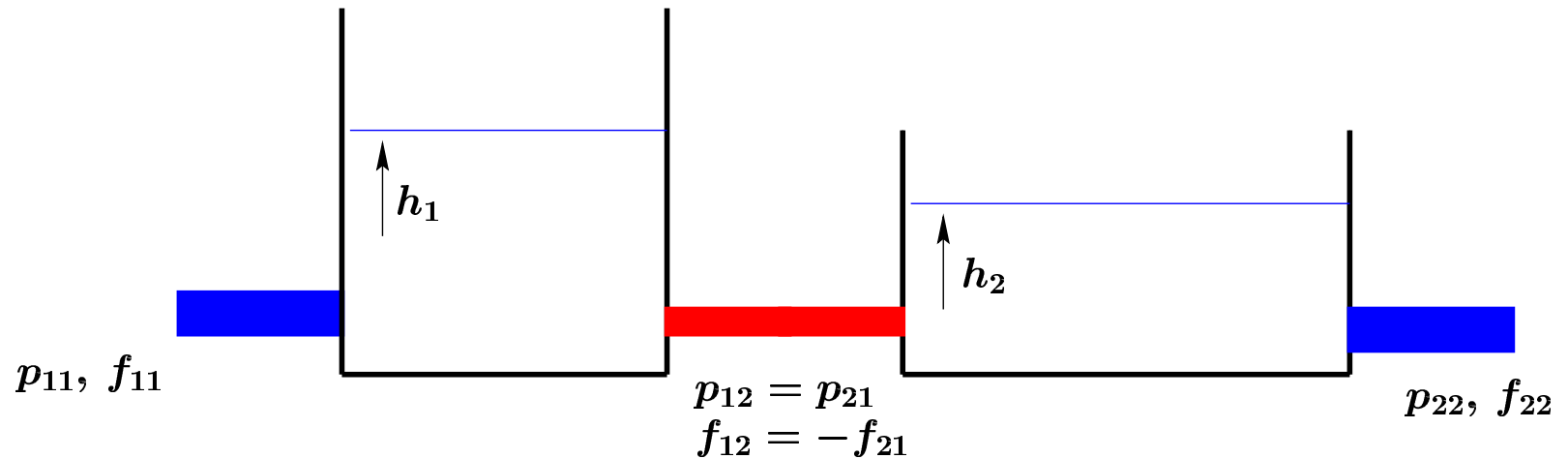
The **inappropriateness** of input - to - output connections is best illustrated by the following simple physical example:



Logical choice of inputs: the pressures  $p_{11}, p_{12}, p_{21}, p_{22}$ , and of the outputs  $f_{11}, f_{12}, f_{21}, f_{22}$ .

In any case, the choice should be **‘symmetric’**.

# INPUT - to - OUTPUT CONNECTIONS



## Interconnection constraints:

$$p_{12} = p_{21}, \quad f_{12} = f_{21}.$$

Equates two 'inputs' and two 'outputs'.

# LINEAR RLCT CIRCUITS

## BUILDING BLOCKS

### Module Types:

Resistors, Capacitors, Inductors, Transformers, Connectors.

All terminals are of the same type: **electrical**

There are 2 variables associated with each terminal,  $(V, I)$ ,

$V$  the *potential*,

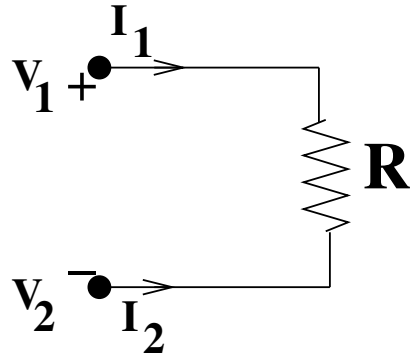
$I$  the *current* (counted  $> 0$  when it flows *into* the module).

~> terminal signal space  $\mathbb{R}^2$ .



# LINEAR RLCT CIRCUITS

## SPECIFICATION of the BEHAVIOR of the MODULES



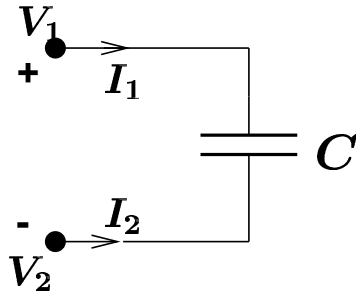
**Resistor:** 2-terminal module.

Parameter:  $R$  (resistance in ohms, say).

Device laws:

$$V_1 - V_2 = R I_1 ; \quad I_1 + I_2 = 0.$$

# LINEAR RLCT CIRCUITS



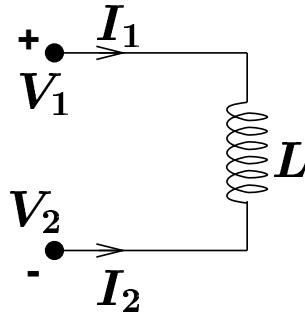
**Capacitor:** 2-terminal module.

Parameter:  $C$  (capacitance in farads, say).

Device laws:

$$C \frac{d}{dt}(V_1 - V_2) = I_1; \quad I_1 + I_2 = 0.$$

# LINEAR RLCT CIRCUITS



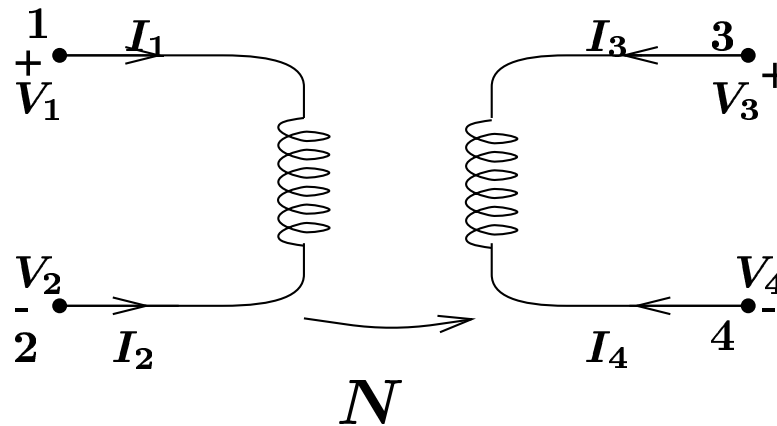
**Inductor:** 2-terminal module.

**Parameter:**  $L$  (inductance in henrys, say).

**Device laws:**

$$L \frac{d}{dt} I_1 = V_1 - V_2 ; \quad I_1 + I_2 = 0.$$

# LINEAR RLCT CIRCUITS



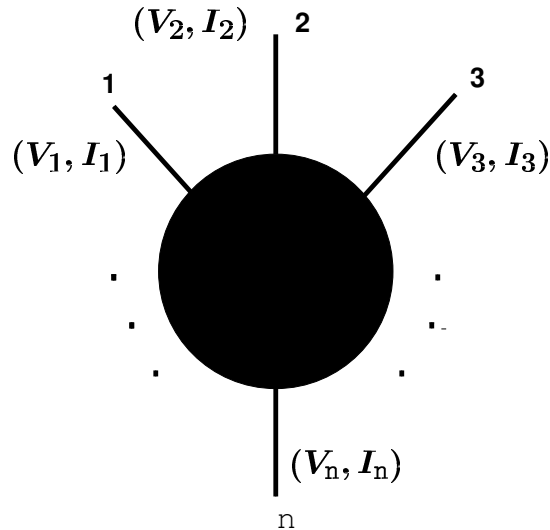
**Transformer:** 4-terminal module; terminals (1,2): primary;  
terminals (3,4): secondary.

Parameter:  $N$  (the turns ratio,  $\in (0, \infty)$ ).

Device laws:

$$\begin{aligned} V_3 - V_4 &= N(V_1 - V_2); & I_1 &= -NI_3; \\ I_1 + I_2 &= 0; & I_3 + I_4 &= 0. \end{aligned}$$

# LINEAR RLCT CIRCUITS



**Connector:** many-terminal module.

**Parameter:**  $n$  (number of terminals, an integer).

**Device laws:**

$$V_1 = V_2 = \dots = V_n; \quad I_1 + I_2 + \dots + I_n = 0.$$

# LINEAR RLCT CIRCUITS

## MODULES and TERMINAL ASSIGNMENT

### Modules

<b>Resistors</b>	$r_1, r_2, \dots, r_{n_r},$	parameters $R_1, R_2, \dots, R_{n_r};$
<b>Capacitors</b>	$c_1, c_2, \dots, c_{n_c},$	parameters $C_1, C_2, \dots, C_{n_c};$
<b>Inductors</b>	$\ell_1, \ell_2, \dots, \ell_{n_\ell},$	parameters $L_1, L_2, \dots, L_{n_\ell};$
<b>Transformers</b>	$T_1, T_2, \dots, T_{n_T},$	parameters $N_1, N_2, \dots, N_{n_T};$
<b>Connectors</b>	$k_1, k_2, \dots, k_{n_k},$	parameters $n_1, n_2, \dots, n_{n_k}.$

This yields the set of terminals

$$\mathbb{T} = \{1, 2, \dots, 2(n_r + n_c + n_\ell) + 4n_T + n_1 + n_2 + \dots + n_{n_k}\}.$$

# LINEAR RLCT CIRCUITS

## INTERCONNECTION ARCHITECTURE

Interconnection architecture :

$\mathbb{I}$  = a set of disjoint (unordered) pairs of different elements (i.e., doubletons) from  $\mathbb{T}$ .

# LINEAR RLCT CIRCUITS

## MANIFEST VARIABLE ASSIGNMENT

External terminals =  $\mathbb{E} := \mathbb{T} - \cup_{\mathbb{I}} \{a, b\}$ .

Manifest variables = external terminal voltages and currents  
=  $\prod_{k \in \mathbb{I}} (V_k, I_k)$ . Denote the manifest variables by  
 $\prod_{k \in \mathbb{I}} (V_k, I_k)$  as  $(V, I) \in \mathbb{R}^{2\mathbb{E}}$ .

Manifest behavior:  $\subseteq (\mathbb{R}^{2\mathbb{E}})^{\mathbb{R}}$ .

Denote further the **full behavior** (the behavior of all the terminal voltages and currents) by  $\mathfrak{B}_{\mathbb{T}} \subseteq (\mathbb{R}^{2\mathbb{T}})^{\mathbb{T}}$ .



# LINEAR RLCT CIRCUITS

## FULL BEHAVIORAL EQUATIONS

### 1. Module Laws:

1.1 Resistors: for each resistor  $r_n$ , terminals  $(t_1^{r_n}, t_2^{r_n})$ ,

$$V_{t_1^{r_n}} - V_{t_2^{r_n}} = R_n I_{t_1^{r_n}}; \quad I_{t_1^{r_n}} + I_{t_2^{r_n}} = 0.$$

1.2 Capacitors: for each capacitor  $c_n$ , terminals  $(t_1^{c_n}, t_2^{c_n})$ ,

$$\frac{d}{dt} C_n (V_{t_1^{c_n}} - V_{t_2^{c_n}}) = I_{t_1^{c_n}}; \quad I_{t_1^{c_n}} + I_{t_2^{c_n}} = 0.$$

1.3 Inductors: for each inductor  $\ell_n$ , terminals  $(t_1^{\ell_n}, t_2^{\ell_n})$ ,

$$\frac{d}{dt} L_n I_{t_1^{\ell_n}} - V_{t_2^{\ell_n}}; \quad I_{t_1^{\ell_n}} + I_{t_2^{\ell_n}} = 0.$$

1.4 Transformers: for each transformer  $T_n$ , terminals  $(t_1^{T_n}, t_2^{T_n}, t_3^{T_n}, t_4^{T_n})$ ,

$$\begin{aligned} V_{t_1^{T_n}} - V_{t_2^{T_n}} &= N_n (V_{t_3^{T_n}} - V_{t_4^{T_n}}); & I_{t_3^{T_n}} &= -N_n I_{t_1^{T_n}} \\ I_{t_1^{T_n}} + I_{t_2^{T_n}} &= 0; & I_{t_3^{T_n}} + I_{t_4^{T_n}} &= 0. \end{aligned}$$

1.5 Connectors: for each connector  $k_n$ , terminals  $(t_1^{k_n}, \dots, t_{n_{k_n}}^{k_n})$ ,

$$V_{t_1^{k_n}} = \dots = V_{t_{n_{k_n}}^{k_n}}; \quad I_{t_1^{k_n}} + \dots + I_{t_{n_{k_n}}^{k_n}} = 0.$$

# LINEAR RLCT CIRCUITS

## 2. Interconnection Laws:

For each 'connected' terminal pair  $\{a, b\} \in \mathbb{I}$  :

$$V_a = V_b; \quad I_a + I_b = 0.$$

Solution of behavioral equations  $\rightsquigarrow \mathcal{B}_{\mathbb{T}}$ .

After elimination of internal variables  $\rightsquigarrow \mathcal{B}_{\mathbb{E}}$ .

# LINEAR RLCT CIRCUITS

## PROPERTIES of $\mathcal{B}_E$

When is  $\mathcal{B}_E \subseteq (\mathbb{R}^{2E})^{\mathbb{R}}$   
the external terminal behavior of a circuit  
containing a finite number of positive  
 $R$ 's,  $L$ 's,  $C$ 's,  $T$ 's, and connectors?

It is possible to derive **necessary & sufficient conditions!**

# LINEAR RLCT CIRCUITS

1.  $\mathfrak{B}_{\mathbb{E}} \in \mathcal{L}^{2\mathbb{E}}.$

2. KVL:

$$((V, I) \in \mathfrak{B}_{\mathbb{E}}) \text{ and } (\alpha \in \mathcal{C}^{\infty}(\mathbb{R}, \mathbb{R})) \Rightarrow ((V + \alpha e) \in \mathfrak{B}_{\mathbb{E}})$$

with

$$e = \begin{bmatrix} 1 \\ 1 \\ \vdots \\ 1 \end{bmatrix}$$

3. KCL:

$$((V, I) \in \mathfrak{B}_{\mathbb{E}}) \Rightarrow (e^{\top} I = 0)$$

# LINEAR RLCT CIRCUITS

4. Input cardinality:  $m(\mathcal{B}_{\mathbb{E}}) = \mathbb{E}$

5. Hybridicity:

There exists an input/output choice such that the input variables  $(u_1, u_2, \dots, u_{\mathbb{E}})$  and output variables  $(y_1, y_2, \dots, y_{\mathbb{E}})$  pair as follows:

$$\{u_i, y_i\} = \{V_i, I_i\}$$

Each terminal is either **current controlled** or **voltage controlled**.

# LINEAR RLCT CIRCUITS

## 6. Passivity:

Assume for simplicity  $\mathfrak{B}_{\mathbb{E}} \in \mathbb{L}_{\text{controllable}}^{2\mathbb{E}}$ . There holds

$$\int_0^{+\infty} V^{\top}(t)I(t) dt \geq 0$$

for all  $(V, I) \in \mathfrak{B}_{\mathbb{E}}$  of compact support.

This states that the net electrical energy flows into the circuit.

# LINEAR RLCT CIRCUITS

## 7. Reciprocity:

Assume again for simplicity  $\mathfrak{B}_{\mathbb{E}} \in \mathbb{L}_{\text{controllable}}^{2\mathbb{E}}$ . There holds

$$\int_{-\infty}^{+\infty} V_1^\top(t) I_2(-t) dt = \int_{-\infty}^{+\infty} I_1^\top(t) V_2(-t) dt$$

for all  $(V_1, I_1), (V_2, I_2) \in \mathfrak{B}_{\mathbb{E}}$  of compact support.

Equivalently:  $\mathfrak{B}_{\mathbb{E}} = \text{rev}(\mathfrak{B}_{\mathbb{E}}^{\perp \Sigma})$ ,

where **rev** denotes **time-reversal**, and  $\Sigma = \begin{bmatrix} O & I \\ -I & O \end{bmatrix}$ .

This curious properties may be translated into:

**The influence of terminal  $i$  on terminal  $j$  is equal to the influence of terminal  $j$  on terminal  $i$ .**

# LINEAR RLCT CIRCUITS

## Proof of necessity:

Show that the modules satisfy properties (1) to (7).

Show that these properties remain valid after one interconnection.

The difficult part here is (4).

## Proof of necessity:

‘Synthesis’.



## TERMINALS or PORTS?

Note that (for instance for electrical circuits) we have used the **terminal description**. It is simply more appropriate and more general than the **port description** (even when using only ‘port’ devices).

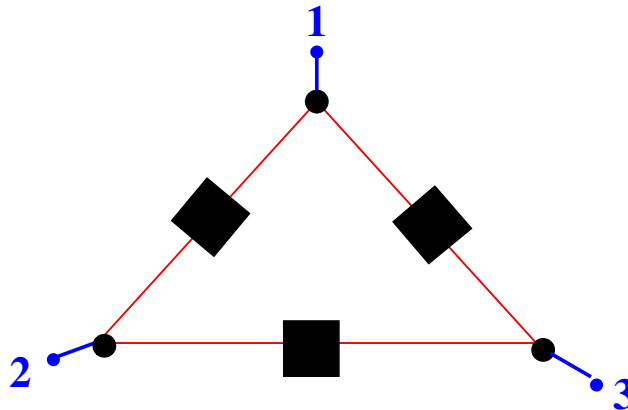
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Example:



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The port description is not **‘closed under interconnection’**.

However, port descriptions are more **parsimonious** in the choice of variables (it halves their number). It is important to incorporate this parsimony.

# RECAPITULATION

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- **Adapted to computer assisted modelling**
- **Hierarchical, reusable, extendable**
- **Many latent variables, many equations (many static relations, i.e., algebraic equations). Far distance from i/o, i/s/o, tf. f'ns.**
- **Importance of elimination algorithms**

# CONCLUSION

**\* for physical systems ( $\Rightarrow \Leftarrow$  signal processors) \***



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\* for physical systems ( $\Rightarrow \Leftarrow$  signal processors) \*

- External variables are basic, but **what 'drives' what**, is not.
- **Interconnection, variable sharing**, rather than **input selection**, is the basic mechanism by which a system interacts with its environment.

# BONDGRAPHS

**Views interconnected systems indeed in terms of  
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their (inner) product must be **power.**

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an **effort** variable and a **flow** variable

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Examples:

- Electrical ports: **effort**: voltage, **flow**: current
- Mechanical ports: **effort**: force, **flow**: velocity
- Thermal ports: **effort**:  $T$ , **flow**:  $Q/T$
- etc. etc.

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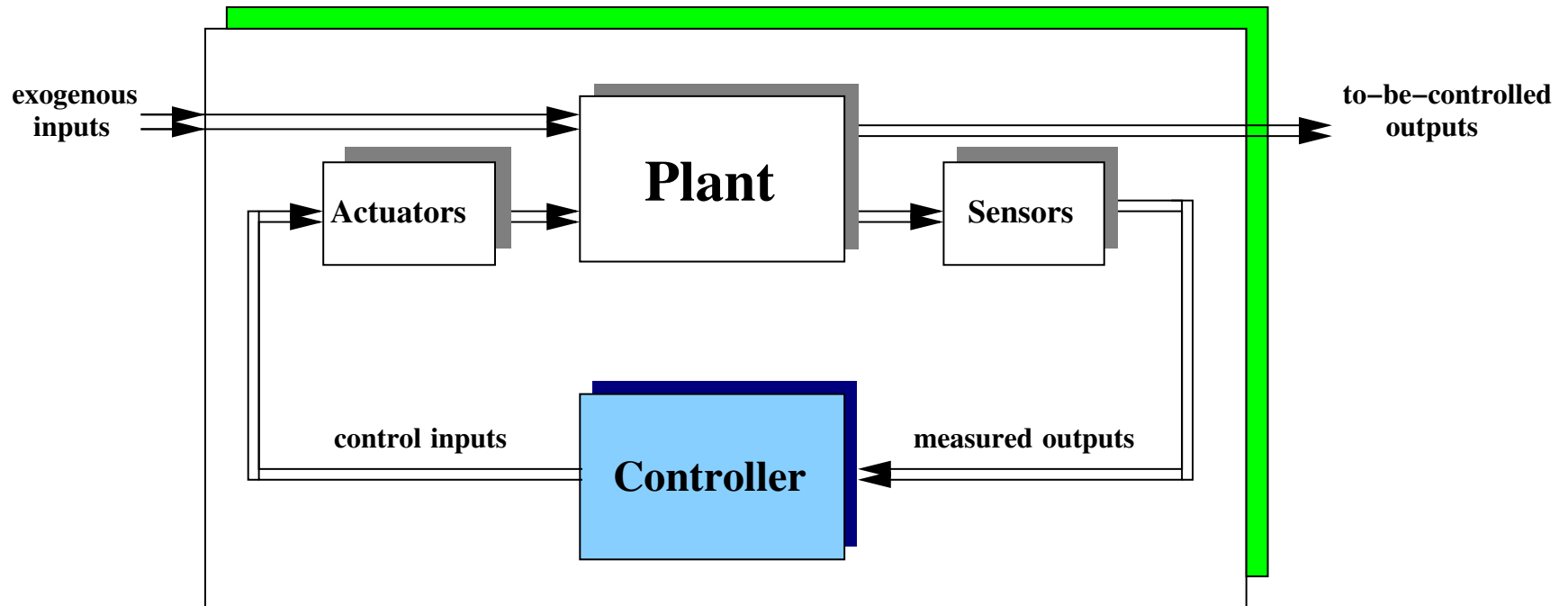
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mechanical terminals equate **positions**, **NOT velocities**
- effort/flow, while apparently deep, remains unexplored
- interconnections happen via terminals, not ports.
- there is more structure to interconnection variables than effort/flow.



# **CONTROL in a BEHAVIORAL SETTING**

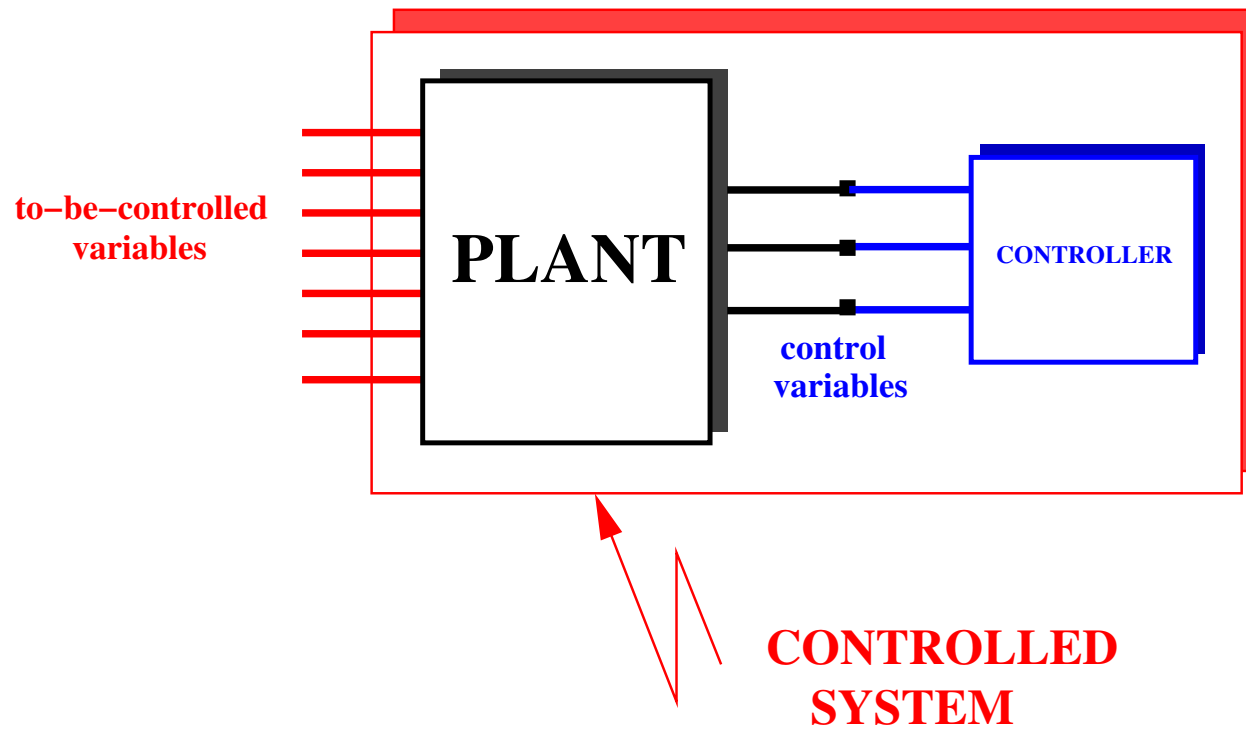
# FEEDBACK CONTROL

The usual paradigm for control:



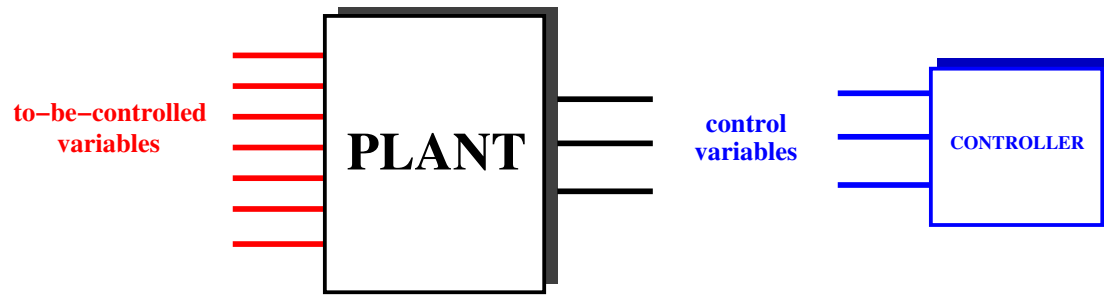
'Intelligent' Control

# BEHAVIORAL CONTROL



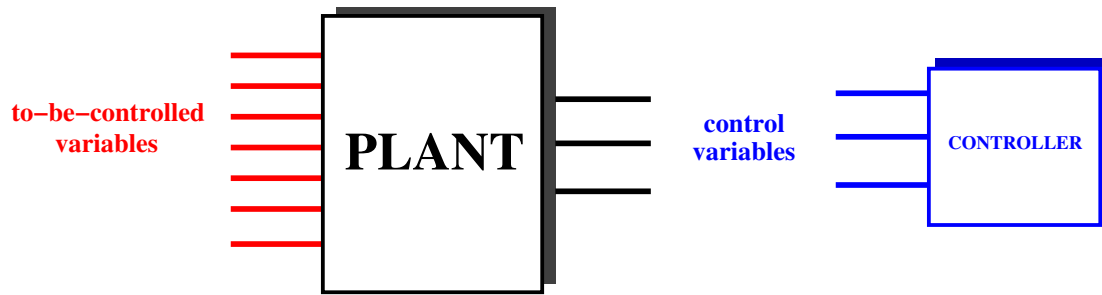
**Control as interconnection**

# BEHAVIORAL CONTROL

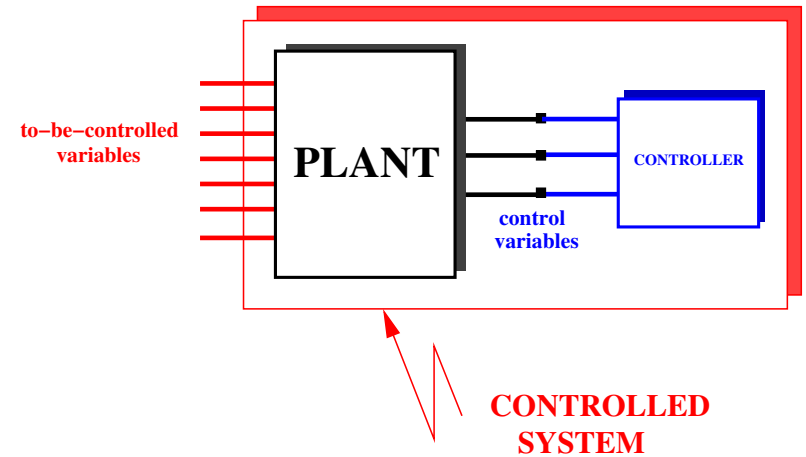


**Before interconnection**

# BEHAVIORAL CONTROL



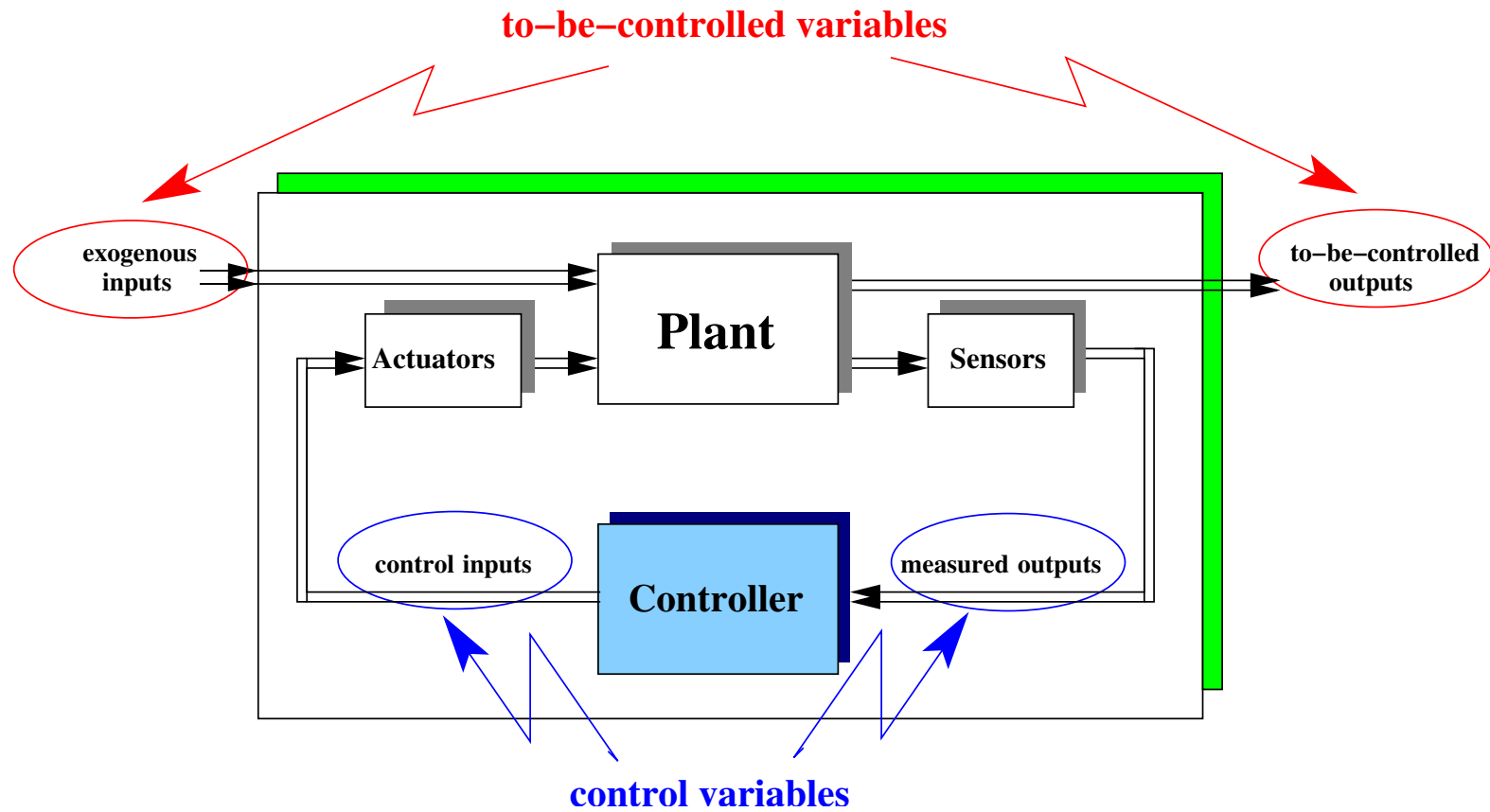
Before interconnection



After interconnection

**Control = designing a subsystem**

# Feedback control as an example

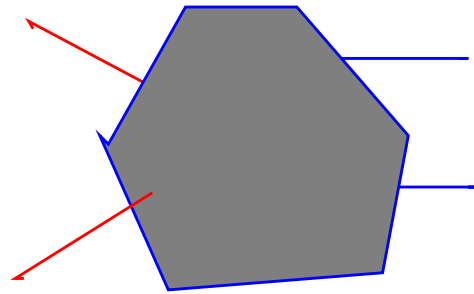


## **‘Example’**

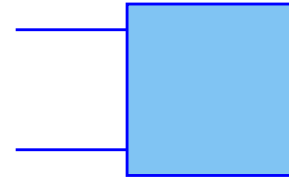
**Many practical control devices do not function as feedback controllers! Dampers, heat fins, pressure valves, overflows, turbulence control strips, characteristic impedances, etc. etc.**

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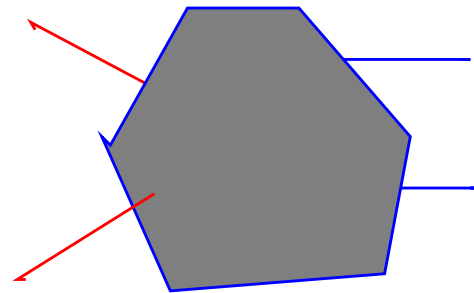


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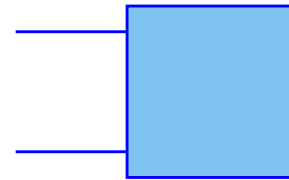


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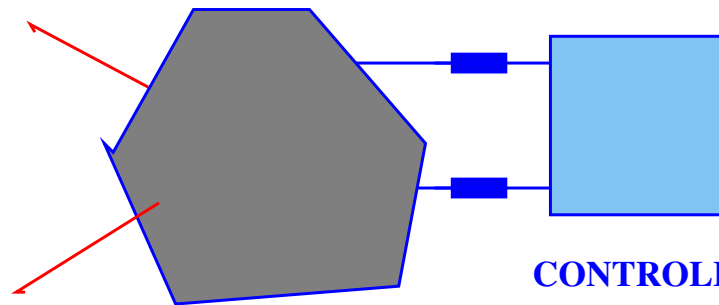
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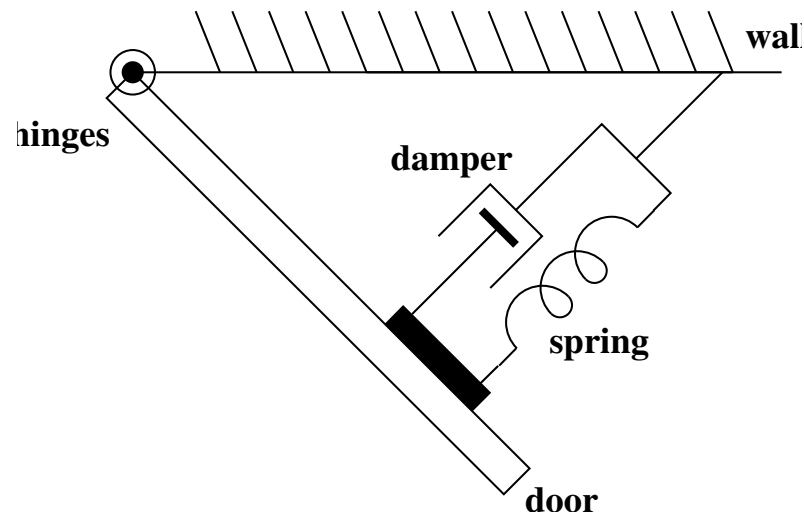


**PLANT**

**CONTROLLER**

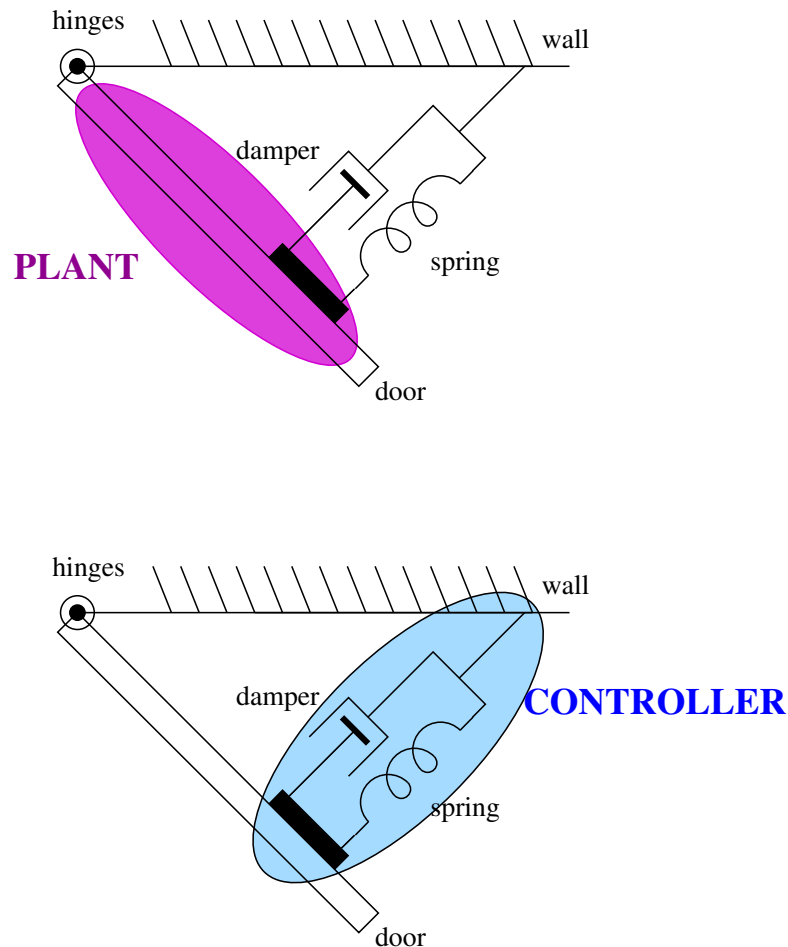
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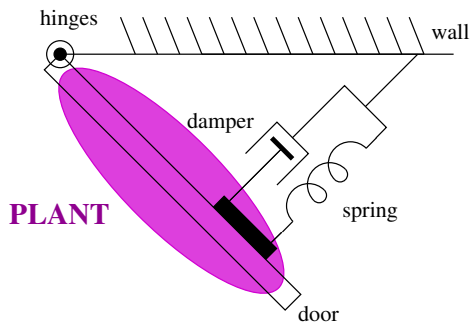
# 'Example'

Equation of motion of the door (the plant):

$$M' \frac{d^2 \theta}{dt^2} = F_c + F_e$$

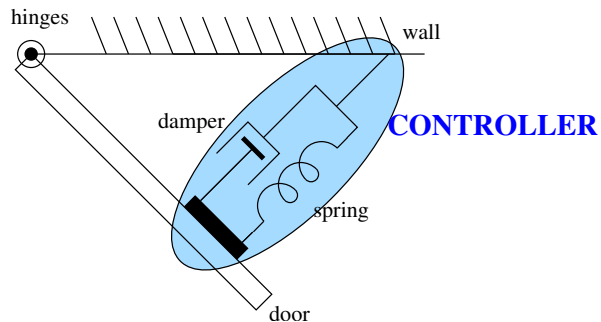
$\theta$ : opening angle,

$F_c$  force device,  $F_e$  exogenous force.



Door closing mechanism (the controller):

$$M'' \frac{d^2 \theta}{dt^2} + D \frac{d\theta}{dt} + K\theta = -F_c.$$



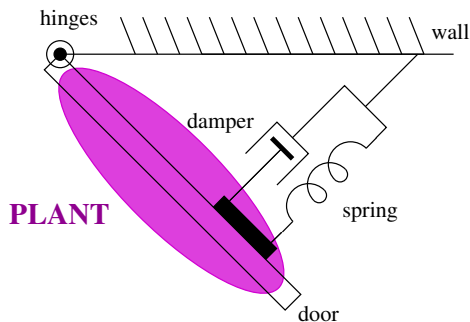
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Controlled behavior:

$$(M' + M'') \frac{d^2 \theta}{dt^2} + D \frac{d\theta}{dt} + K\theta = F_e$$

Specs: small overshoot, fast settling, not-to-high gain from  $F_e \mapsto \theta$ . Controller  $\leadsto M', K$  and  $D$ .

Note: Plant: **second** order; Controller: **second** order; Controlled plant: **second (not fourth)** order.

Note: **PDD controller**, but no noise problems

# MATHEMATIZATION

**Domain of the to-be-controlled variables:  $\mathbb{W}$**

**Domain of the control variables:  $\mathbb{C}$**

**Typically: families of time-signals**

# MATHEMATIZATION

**Full plant behavior:**

$$\mathcal{P}_{\text{full}} = \{(w, c) \in \mathbb{W} \times \mathbb{C} \mid \text{allowed by plant laws}\}$$

**Controller:**

$$\mathcal{C} = \{c \in \mathbb{C} \mid \text{allowed by controller laws}\}$$

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$$\mathcal{K} := \{w \in \mathbb{W} \mid \exists c \in \mathbb{C}$$

such that  $(w, c) \in \mathcal{P}_{\text{full}}$  and  $c \in \mathcal{C}\}$ .



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We say that  $\mathcal{C}$  *implements*  $\mathcal{K}$ , and that  $\mathcal{K}$  is *implementable*

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## Questions:

- Which  $\mathcal{C}$  implements the *desired controlled behavior*  $\mathcal{D}$ ?
- Given  $\mathcal{P}_{\text{full}}$ , which  $\mathcal{K} \subseteq \mathbb{W}$  are implementable?



We henceforth restrict attention to  
**linear time-invariant differential systems.**



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**linear time-invariant differential systems.**

The *behavior*  $\mathfrak{B}$  belongs to  $\mathfrak{L}^w$

$:\Leftrightarrow$

$\exists$  a polynomial matrix  $R \in \mathbb{R}^{\bullet \times w}[\xi]$  such that

$$\mathfrak{B} = \left\{ w \in \mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w) \mid R\left(\frac{d}{dt}\right)w = 0 \right\}.$$



**Plant:**

$$\mathcal{P}_{\text{full}} \in \mathcal{L}^{w+c}.$$

**Controller:**

$$\mathcal{C} \in \mathcal{L}^c.$$

**Controlled system:**

$$\mathcal{K} = \{w \in \mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w) \mid \exists c \in \mathcal{C} : (w, c) \in \mathcal{P}_{\text{full}}\}.$$



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**By the ‘elimination theorem’**

$$\mathcal{K} \in \mathcal{L}^w$$

# IMPLEMENTABILITY

*Which behaviors  $\mathcal{K} \in \mathcal{L}^w$  can be implemented by attaching a controller  $\mathcal{C} \in \mathcal{L}^c$  to a given plant*

*$\mathcal{P}_{\text{full}} \in \mathcal{L}^{w+c}$ ?*

# IMPLEMENTABILITY

*Which behaviors  $\mathcal{K} \in \mathcal{L}^w$  can be implemented by attaching a controller  $\mathcal{C} \in \mathcal{L}^c$  to a given plant  $\mathcal{P}_{\text{full}} \in \mathcal{L}^{w+c}$ ?*

This question has a very concrete and intuitive answer.

Theorem: Let  $\mathcal{P}_{\text{full}} \in \mathcal{L}^{w+c}$  be given.

The behavior  $\mathcal{K} \in \mathcal{L}^w$  is implementable if and only if

$$\mathcal{N} \subseteq \mathcal{K} \subseteq \mathcal{P}$$



# IMPLEMENTABILITY

The behavior  $\mathcal{K} \in \mathcal{L}^w$  is implementable if and only if

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where  $\mathcal{N} \in \mathcal{L}^w$  is the **hidden behavior** defined by

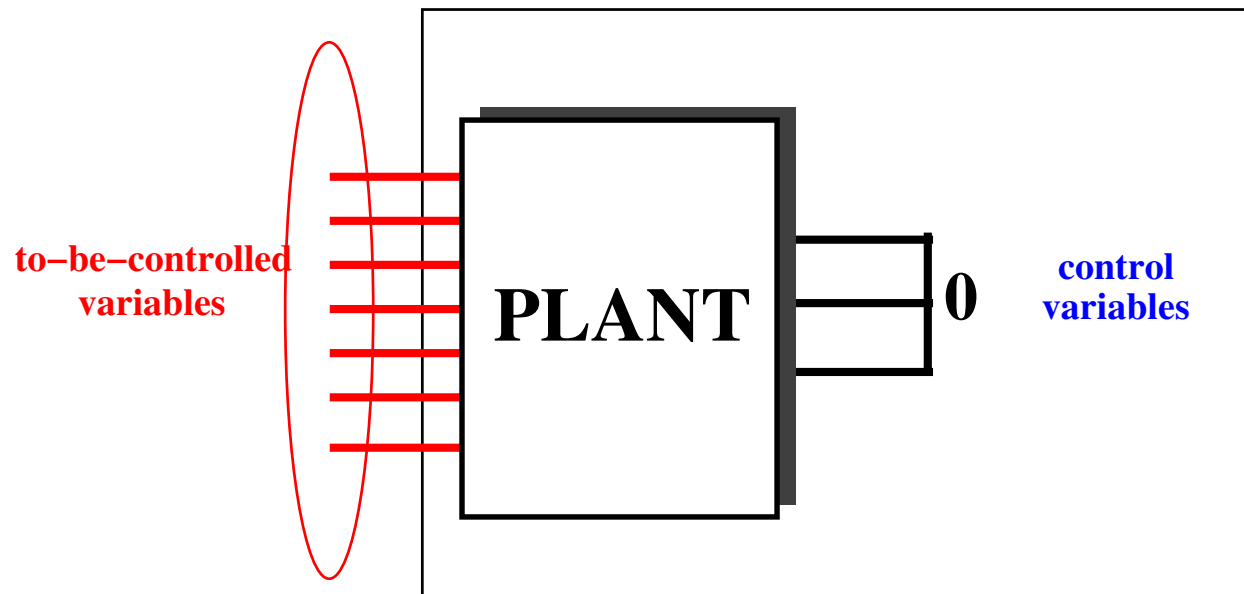
$$\mathcal{N} := \{w \in \mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w) \mid (w, \mathbf{0}) \in \mathcal{P}_{\text{full}}\},$$

and  $\mathcal{P}$  is the **manifest plant behavior** defined by

$$\mathcal{P} := \{w \in \mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w) \mid \exists c : (w, c) \in \mathcal{P}_{\text{full}}\}.$$

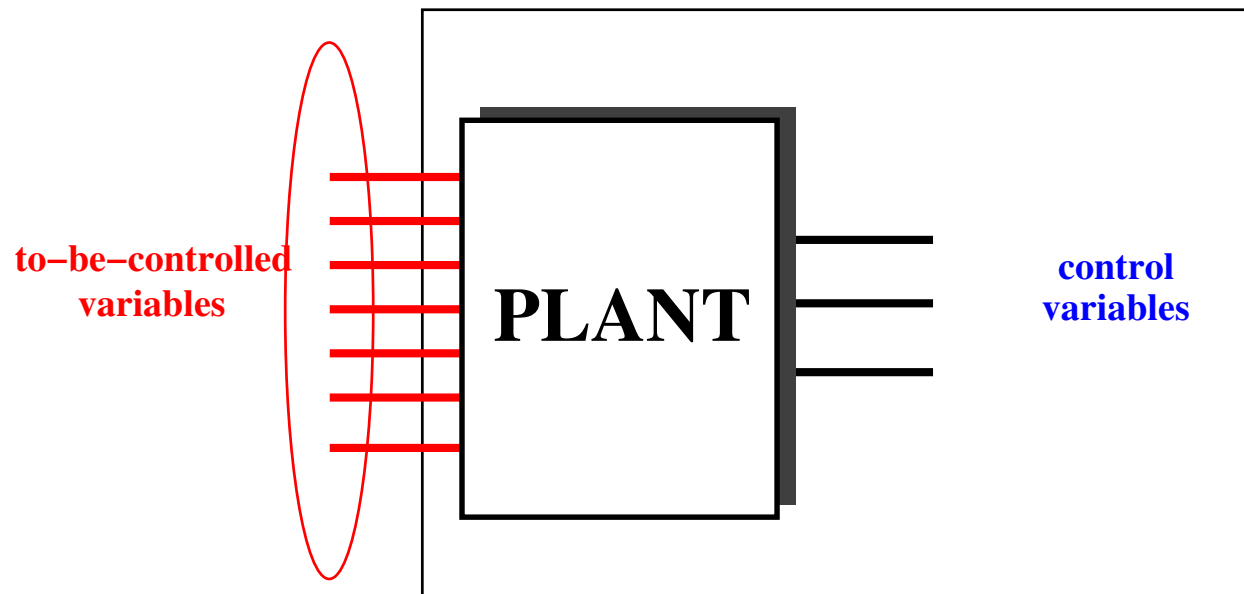
# IMPLEMENTABILITY

$\mathcal{N} \in \mathcal{L}^w$ , the **hidden behavior**



# IMPLEMENTABILITY

$\mathcal{P} \in \mathcal{L}^w$ , the **manifest plant behavior**



# IMPLEMENTABILITY

The behavior  $\mathcal{K} \in \mathcal{L}^w$  is implementable if and only if

$$\mathcal{N} \subseteq \mathcal{K} \subseteq \mathcal{P}$$

This theorem reduces control to linear algebra / functional analysis:  
finding suitable subspaces wedged between given subspaces.

## Example:

Assume **observability** of the to-be-controlled variables  $w$  from the control variables  $c \Leftrightarrow \mathcal{N} = \{0\}$ . Assume  $\mathcal{P} \neq \{0\}$ , **controllable**.

$\Rightarrow$  pole assignability  $\Rightarrow$  stabilizability

e.g.,  $\frac{d}{dt}x = Ax + Bu, y = Cx + Du, c = (u, y), w = x.$

# IMPLEMENTABILITY

The behavior  $\mathcal{K} \in \mathcal{L}^w$  is implementable if and only if

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This theorem reduces control to linear algebra / functional analysis:  
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LQ-control and  $\mathcal{H}_\infty$  control are very neatly worked out from this point of view/

# Regularity

The *full controlled behavior*  $\mathcal{K}_{\text{full}} \subseteq \mathcal{P}_{\text{full}}$  is defined by

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Consider the maps  $m, p : \mathcal{L}^w \rightarrow \{0, 1, \dots, w\}$   
with  $m(\mathcal{B})$  the **number of input variables**,  
and  $p(\mathcal{B})$  the **number of output variables** in  $\mathcal{B}$ .

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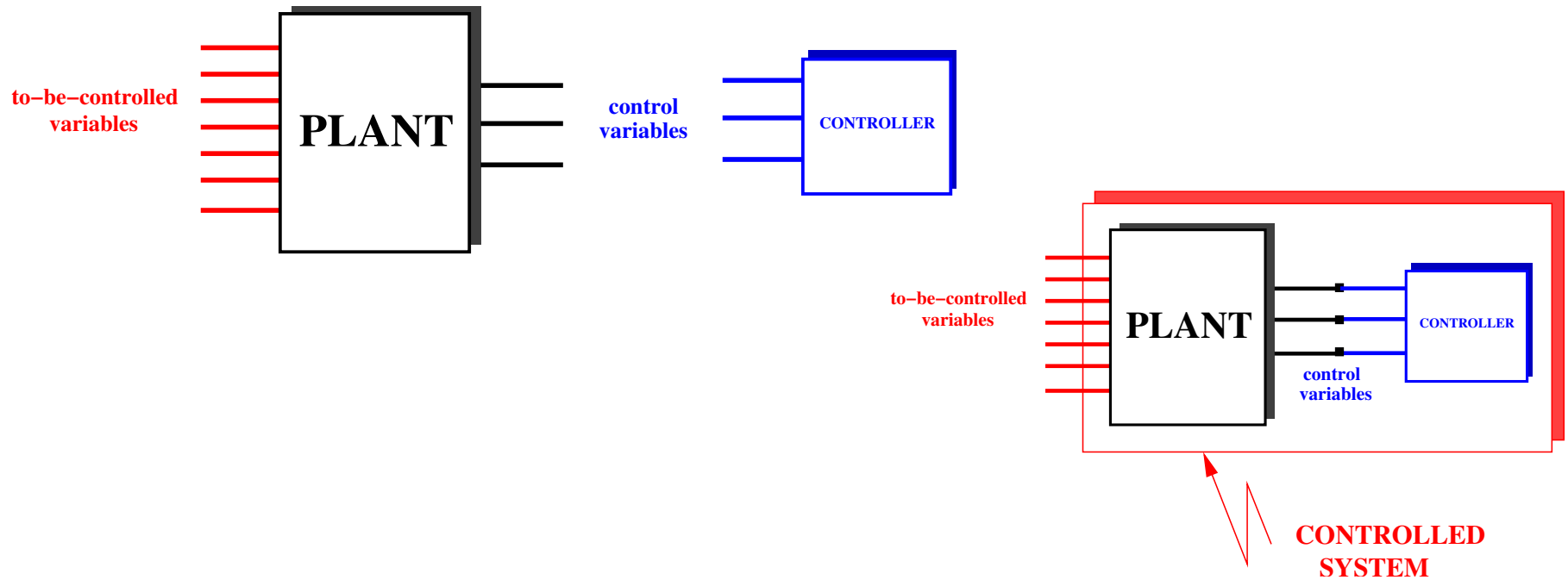
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with  $m(\mathcal{B})$  the **number of input variables**,  
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The controller  $\mathcal{C} \in \mathcal{L}^c$  is said to be **regular** if

$$p(\mathcal{K}_{\text{full}}) = p(\mathcal{P}_{\text{full}}) + p(\mathcal{C}).$$



# Regularity

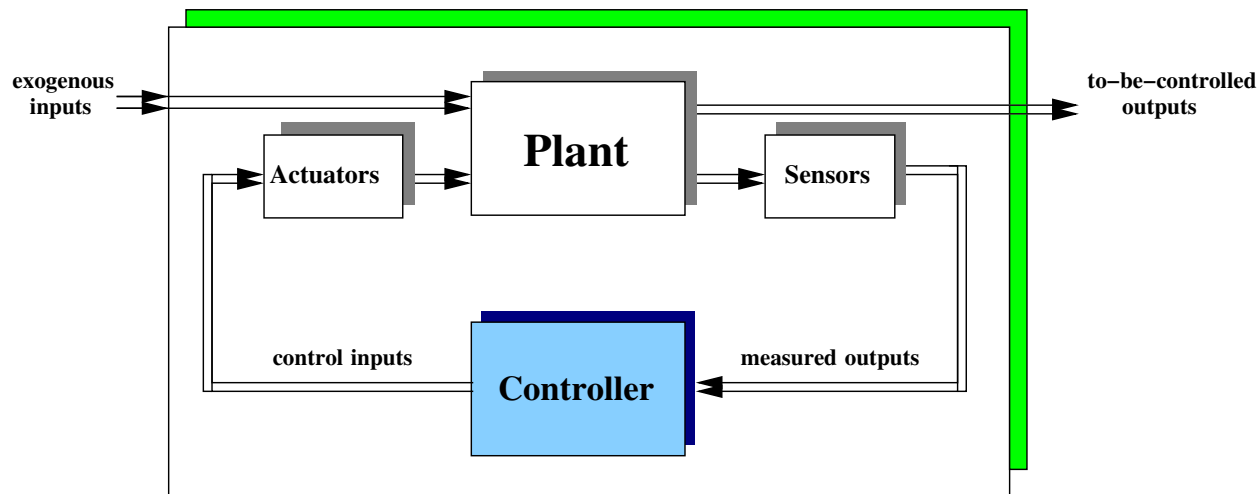


**Regularity :=**

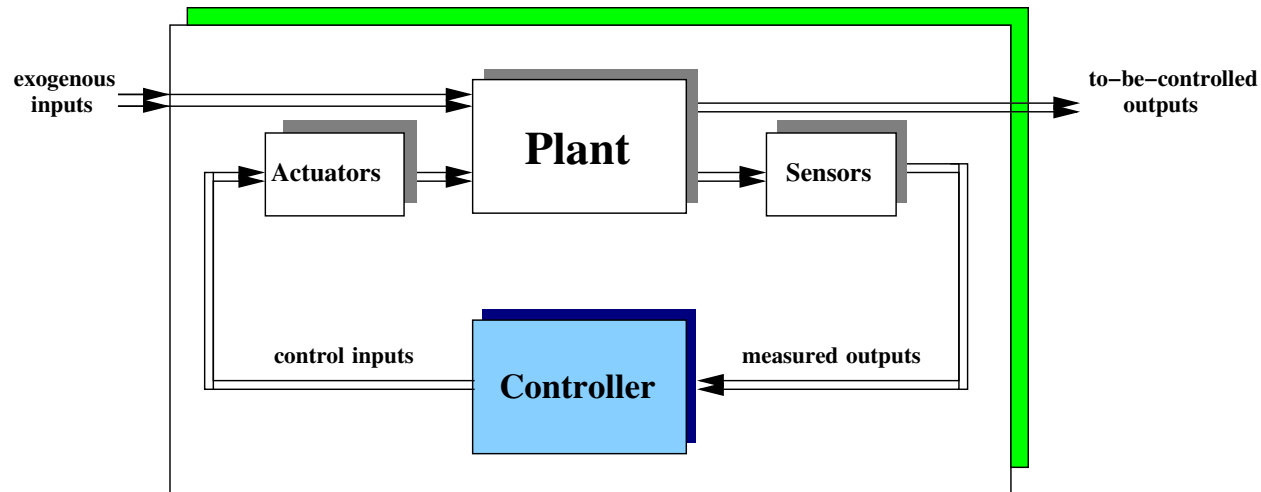
if the controller has  $p$  bound (i.e. output) variables, then the plant loses  $p$  free variables **after interconnection**.

# Regularity

A controller is regular if and only if it can be realized as a **feedback controller** with a **(possibly non-proper)** transfer function from an output to an input in  $\mathcal{P}_{\text{full}}$  for an input/output partition of  $c$ .



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⇒ A controller is regular if and only if it can be viewed as an **'intelligent controller'** that processes sensor inputs outputs into actuator inputs.

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In feedback control, we have the additional property that the controller can be (de)coupled at any time. No state perparatiuon is required in attaching the controller.

## A LOOK BACK

*What have we been trying to do, really ?*

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Set up a 'correct' mathematical framework for discussing **dynamical** systems.

Usable in control, signal processing, econometrics, and, especially, incorporating in an honest way the classical models of physical systems.

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## Basic observations:

- Any reasonable theory takes **open** systems as the basic paradigm.
- Most dynamical models will be described by (differential or difference) equations, but we need a basic notion of **equivalence** of models.
- First principles models invariably contain auxiliary variables
- $\exists$  a complete theory for linear time-invariant systems.



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 $\rightsquigarrow$  **polynomial matrix** based models, with as highlights the **elimination thm.**, **controllability**, and **image repr.**

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- **An input/output model is simply not a 'map'.**
- **The state is a construct, and so are the input and output.**
- **Many technologically very relevant controllers are not sensor-output-to-actuator-input signal processors.**

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    ~> The classical I/O framework fails in the **first and most elementary** examples.
- An input/output model is simply not a 'map'.
- The state is a construct, and so are the input and output.
- Many technologically very relevant controllers are not sensor-output-to-actuator-input signal processors.

# A LOOK BACK

*What have we been trying to do, really ?*

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↳ This is the **historical raison d'être** of state models .
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**Algorithmically** well worked out for linear time-invariant systems.
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## Basic observations:

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- An input/output model is simply not a 'map'.
- The state is a construct, and so are the input and output.
- Many technologically very relevant controllers are not sensor-output-to-actuator-input signal processors.
- The behavioral approach is consistent, pedagogically attractive, pragmatic, and practical.

**Thank you**

**Thank you**

**Thank you**

**Thank you**

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**Thank you**

**Thank you**

**End of the Lecture V**