

European Embedded Control Institute

Graduate School on Control — Spring 2010

The Behavioral Approach to Modeling and Control

Lecture VIII

SYSTEM INTERCONNECTION

Theme

How are systems interconnected?

How are interconnected systems modeled?

How does control fit in?

Theme

How are systems **interconnected** ?

How are interconnected systems modeled?

How does control fit in?

We deal with very simple examples,
mainly electrical circuits and
1-dimensional mechanical systems.

Other applications: hydraulic systems
chemical systems
thermal systems, ...

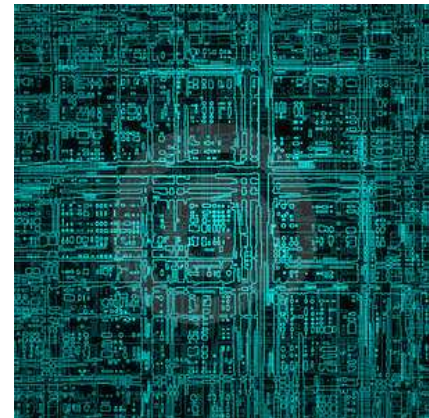
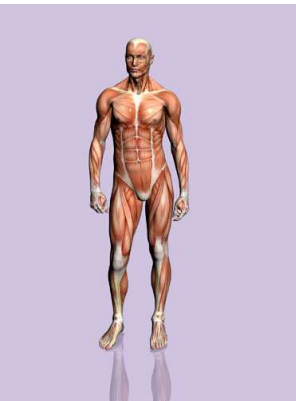
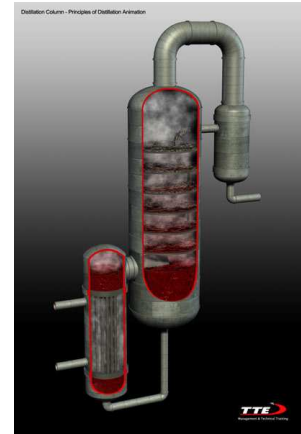
Outline

- ▶ **Motivation**
- ▶ **Modeling by tearing, zooming, and linking**
- ▶ **An example**
- ▶ **Control as interconnection**
- ▶ **Pole assignment and stabilization**

Systems



OIL REFINERY (GVG / PD)



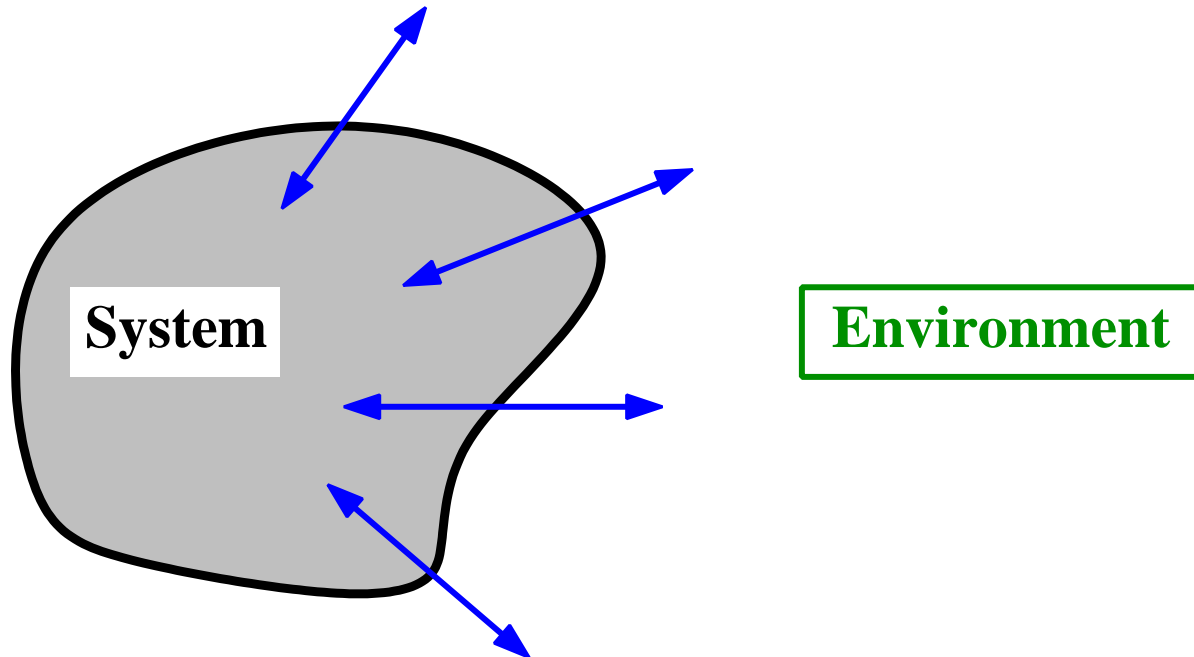
Features

- ▶ **Open**
- ▶ **Interconnected**
- ▶ **Modular**
- ▶ **Dynamic**

The ever-increasing computing power allows to model such complex interconnected systems accurately by tearing, zooming, and linking.

~> **Simulation, model based design, ...**

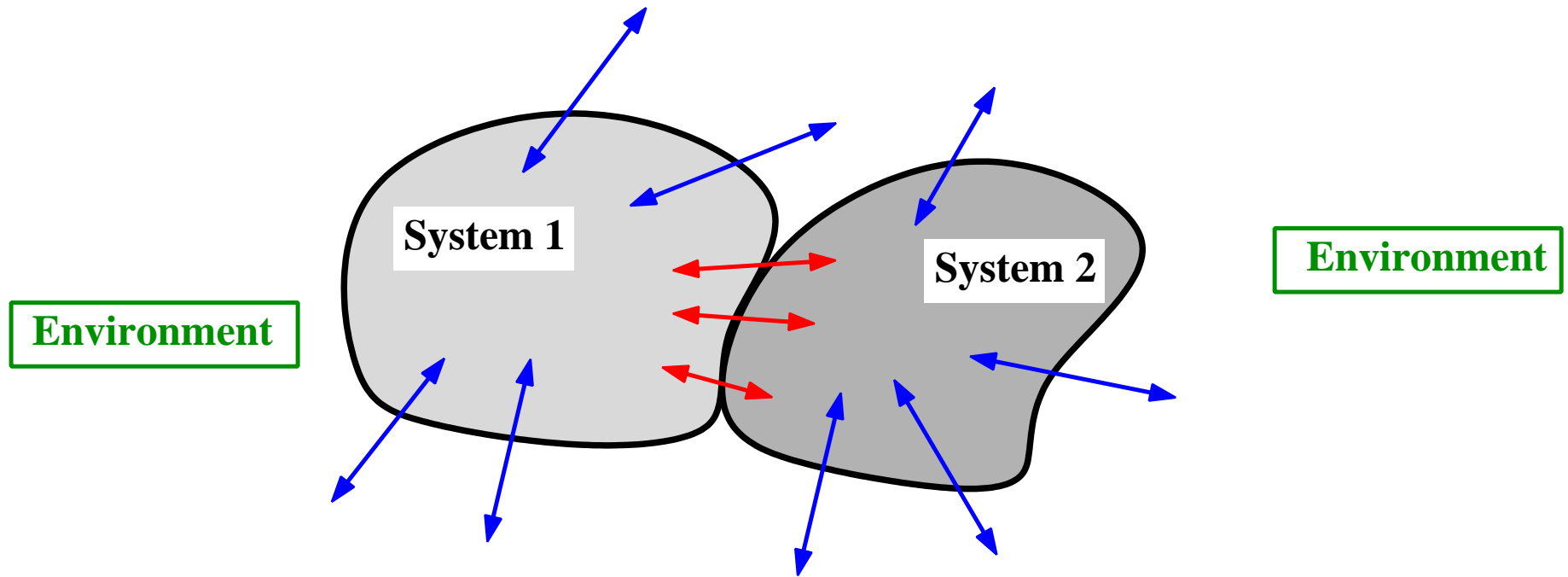
Open systems



Systems are ‘open’, they interact with their environment.

In the previous lectures, we have seen that thinking of systems in terms of their **behavior captures the ‘open’ nature of systems very well.**

Interacting systems



Interconnected systems interact.

How is interaction formalized?

Motivation

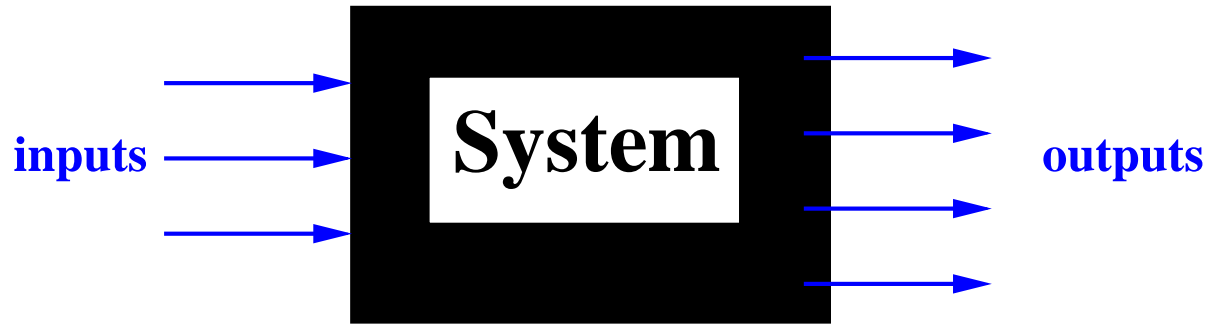
The ever-increasing computing power allows to model complex interconnected systems **accurately.**

Requires the **right mathematical concepts**

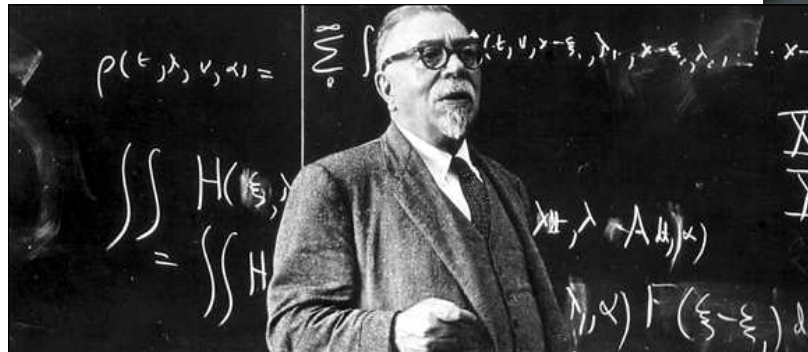
- ▶ for dynamical system (the behavior),
- ▶ for interconnection (this lecture),
- ▶ for interconnection architecture (this lecture).

Classical view

Input/output systems



Oliver Heaviside
(1850-1925)

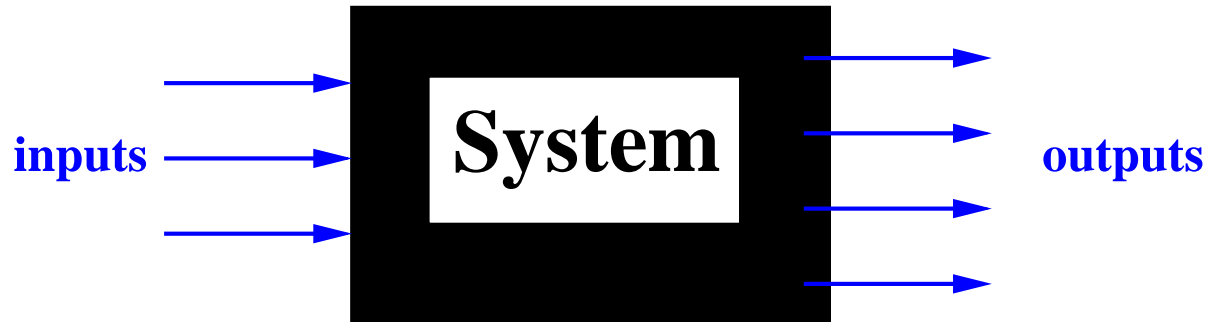


Norbert Wiener
(1894-1964)



Rudy Kalman
(1930-)

Input/output systems

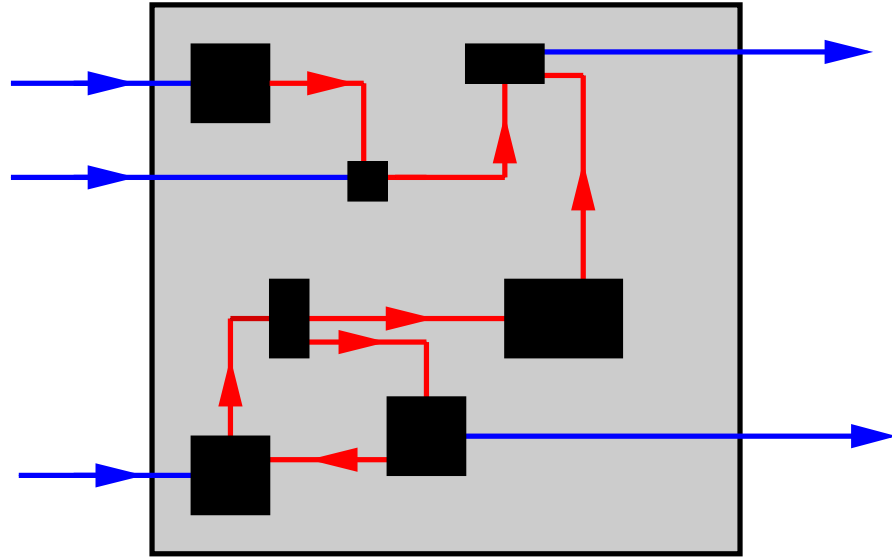


Input/output thinking is *inappropriate* for describing the functioning of physical systems.

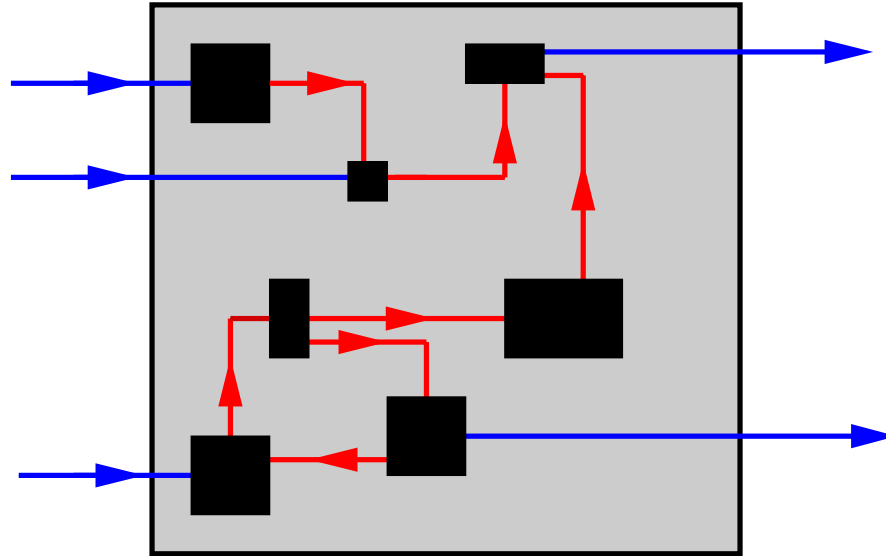
A physical system is not a signal processor.

Better concept: the behavior.

Signal flow graphs



Signal flow graphs



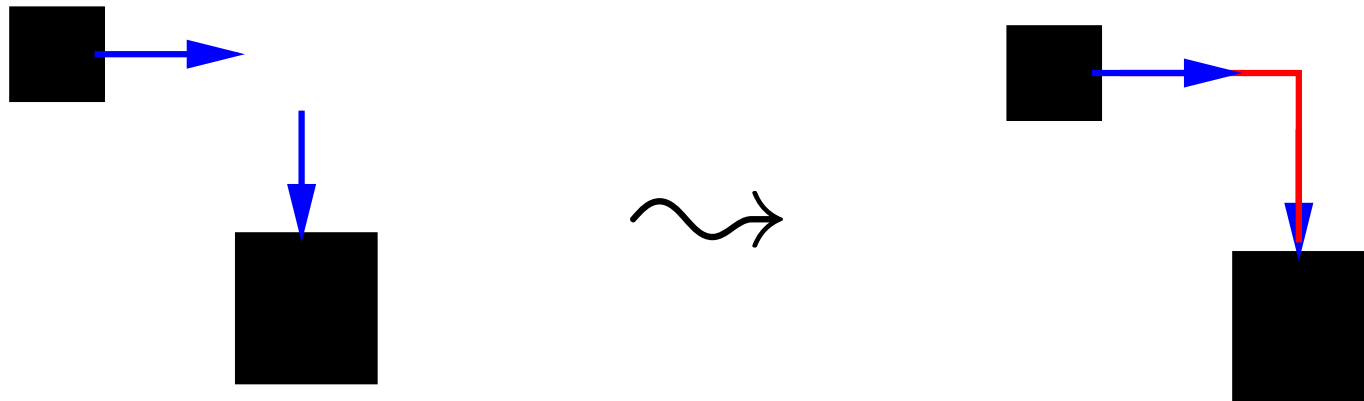
Signal flow graphs are *inappropriate* for describing the interaction architecture of physical systems.

A physical system is not a signal processor.

Better concept: graph with leaves.

Interconnection

Interconnection as output-to-input assignment.

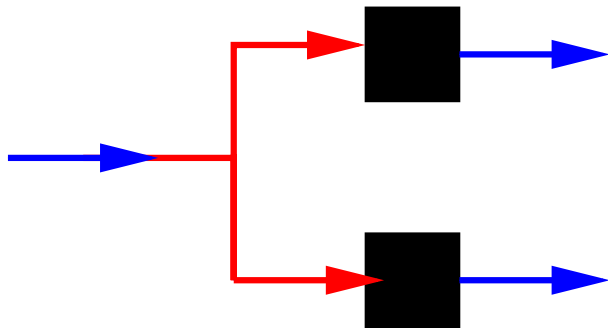
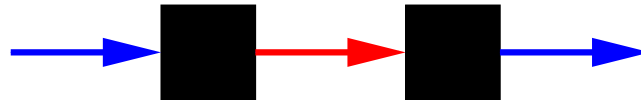


Interconnection

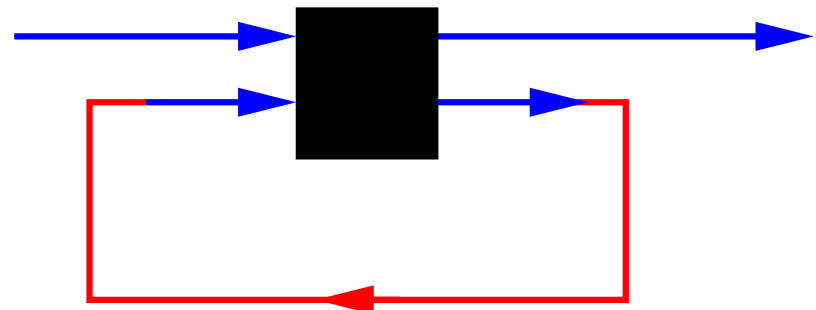
Interconnection as output-to-input assignment.

Examples:

series



parallel



feedback

Interconnection

Interconnection as output-to-input assignment.

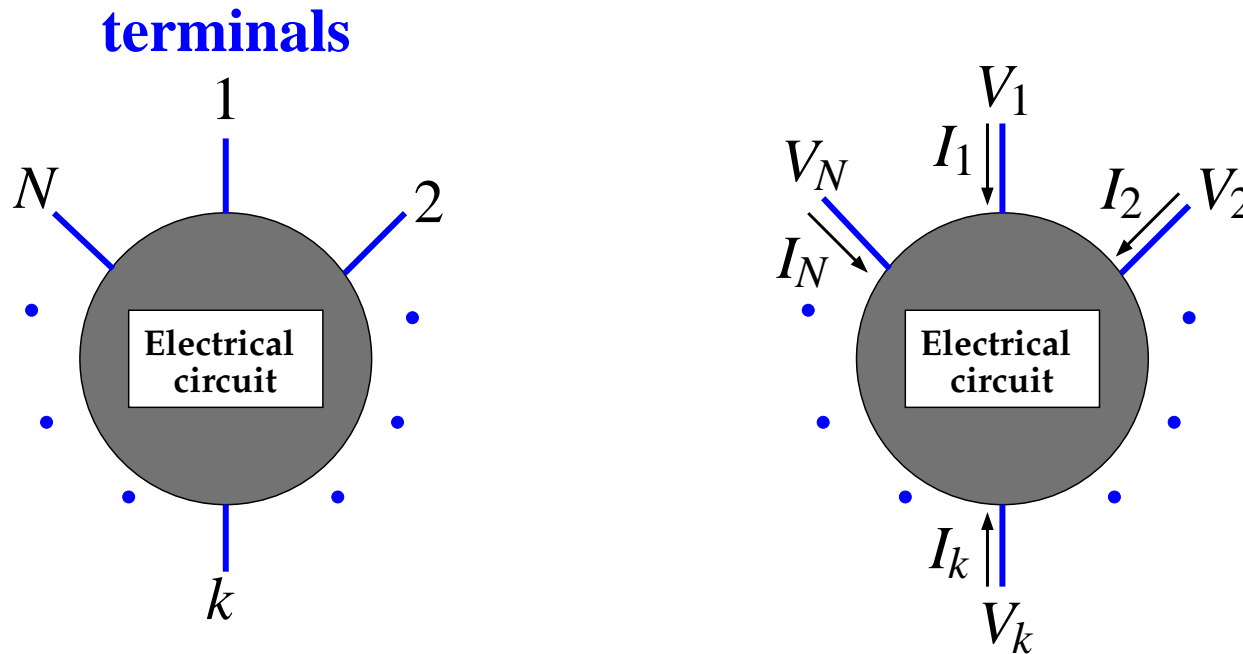
Output-to-input assignment is *inappropriate* for describing the interconnection of physical systems.

A physical system is not a signal processor.

Better concept: variable sharing.

Examples

Electrical circuit

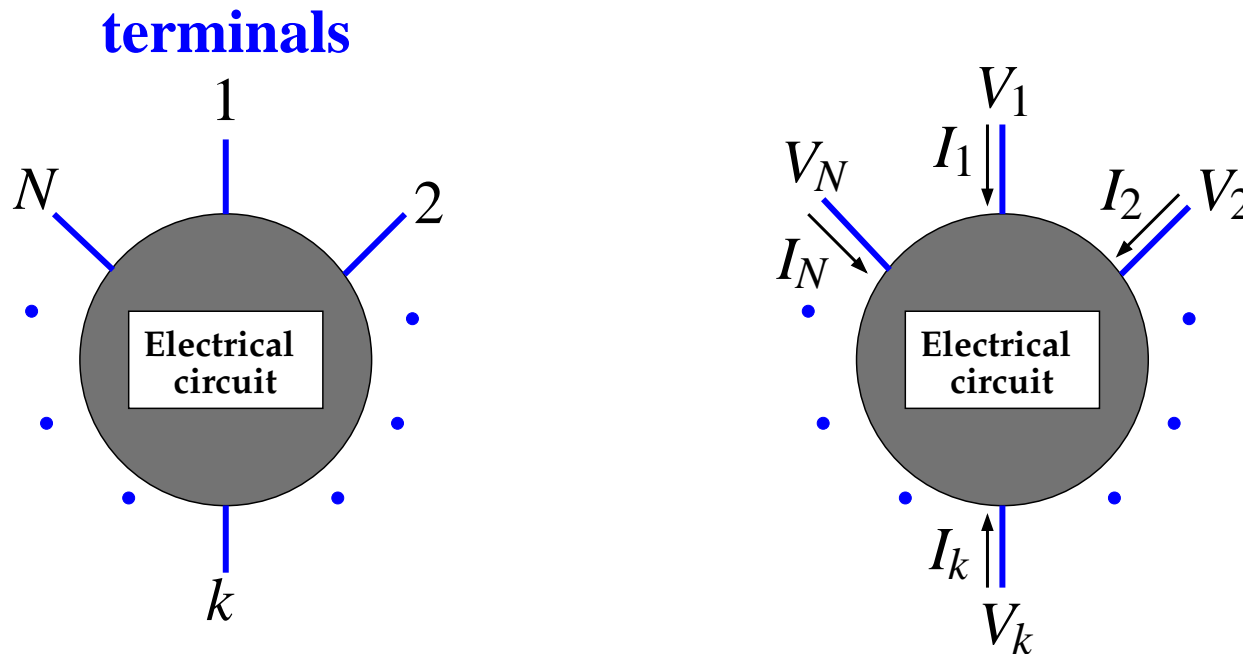


At each terminal:

a **potential (!)** and a **current** (counted > 0 into the circuit),

The relation between potentials of the terminals and voltages across the terminals is discussed elsewhere.

Electrical circuit



At each terminal:

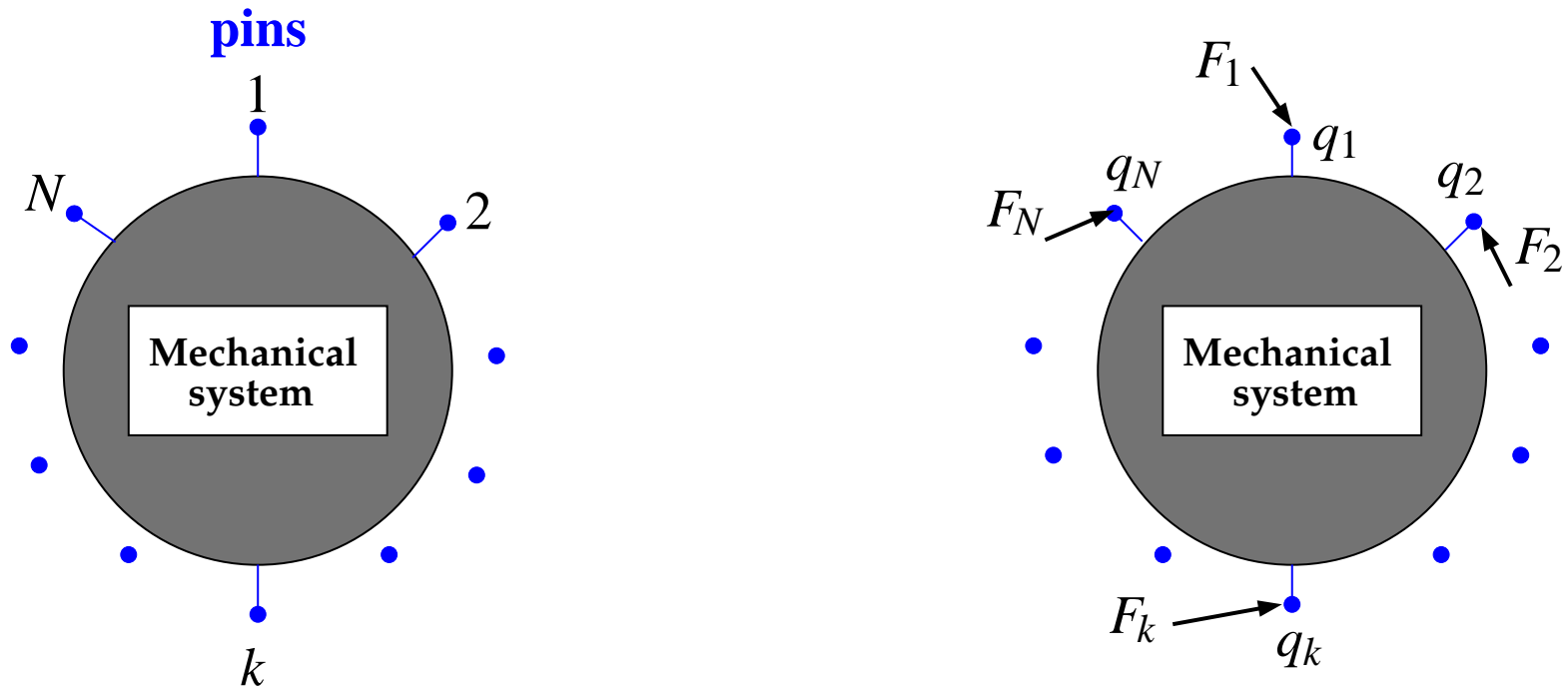
a **potential (!)** and a **current** (counted > 0 into the circuit),

\rightsquigarrow **behavior** $\mathcal{B} \subseteq (\mathbb{R}^N \times \mathbb{R}^N)^{\mathbb{R}}$.

$(V_1, V_2, \dots, V_N, I_1, I_2, \dots, I_N) \in \mathcal{B}$ means:

this potential/current trajectory is compatible with the circuit architecture and its element values.

Mechanical device



At each terminal: a **position** and a **force**.

\rightsquigarrow position/force trajectories $(q, F) \in \mathcal{B} \subseteq ((\mathbb{R}^\bullet)^{2N})^{\mathbb{R}}$.

More generally, a **position**, **force**, **angle**, and **torque**.

Other domains

▶ Thermal systems:

At each terminal: a **temperature** and a **heat flow**.

▶ Hydraulic systems:

At each terminal: a **pressure** and a **mass flow**.

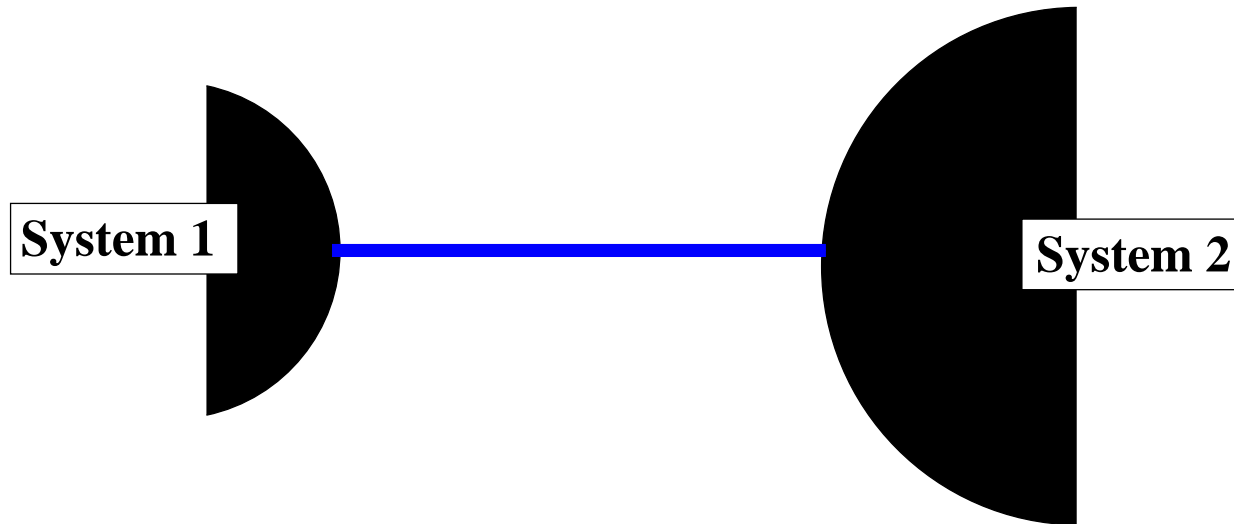
▶ Multidomain systems:

Systems with terminals of different types,
as motors, pumps, etc.

▶ ...

Interconnection

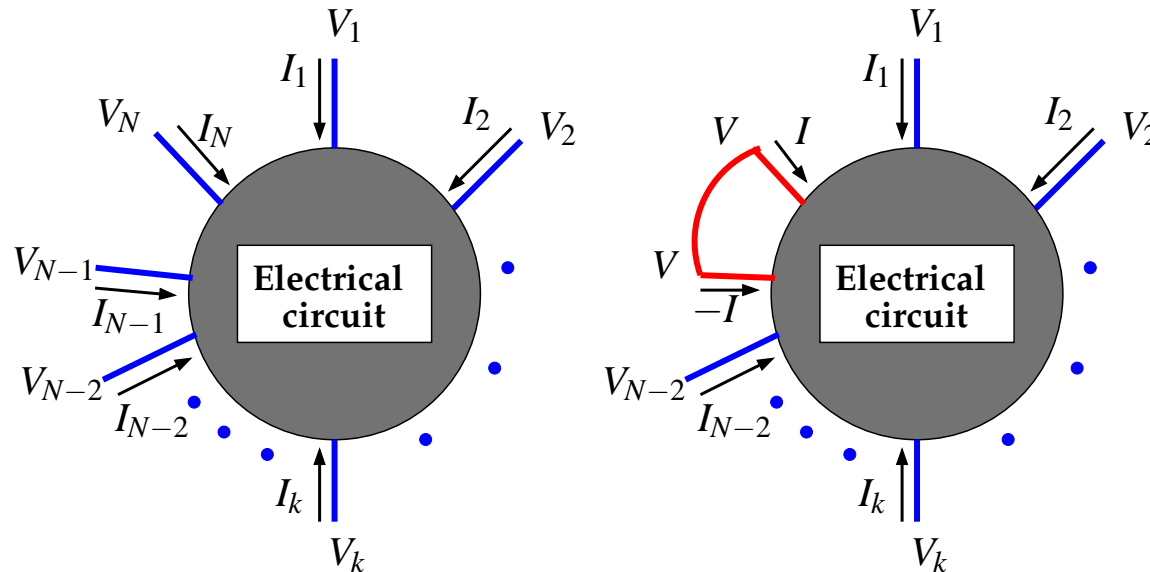
Connection of terminals



By interconnecting, the terminal variables are equated.

Connection of circuit terminals

Interconnection = connecting terminals, like soldering wires together.

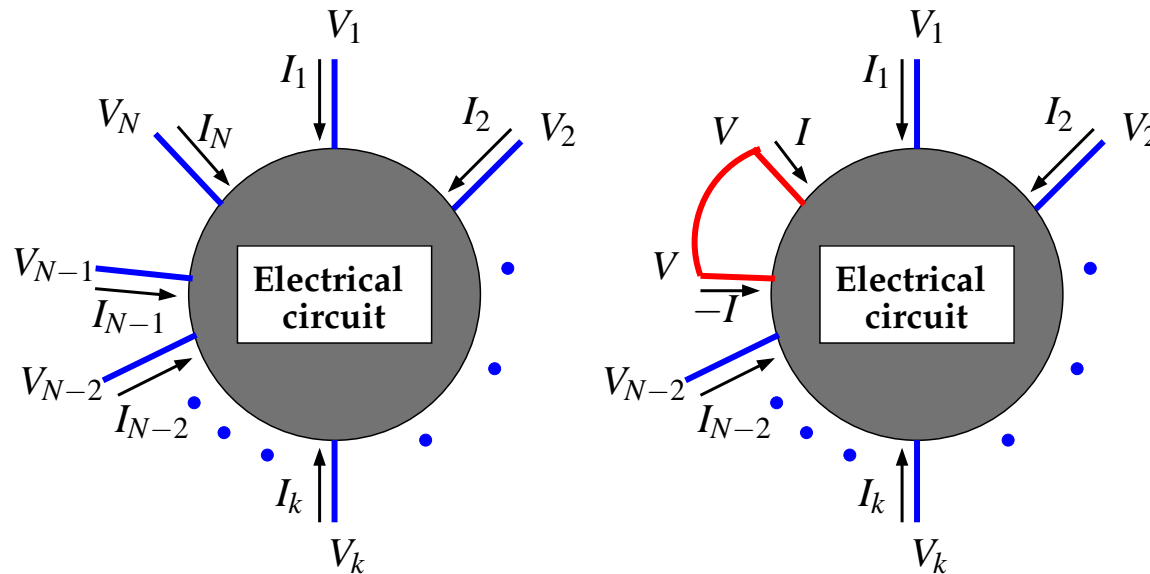


Connecting terminals $N - 1$ and N leads to

$$V_{N-1} = V_N, \quad I_{N-1} + I_N = 0.$$

After interconnection the terminals share the variables V_{N-1}, V_N , and I_{N-1}, I_N (up to a sign).

Connection of circuit terminals



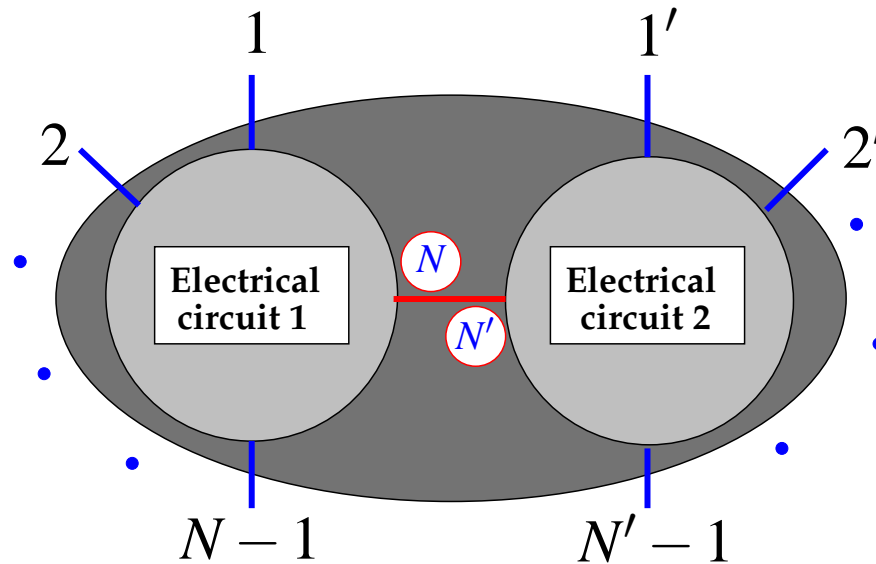
Connecting terminals $N - 1$ and N leads to

$$V_{N-1} = V_N, \quad I_{N-1} + I_N = 0.$$

The interconnected circuit has $N - 2$ terminals. Its behavior =

$$\mathcal{B}' = \{(V_1, I_1, V_2, I_2, \dots, V_{N-2}, I_{N-2}) : \mathbb{R} \rightarrow \mathbb{R}^{2(N-2)} \mid \exists \mathbf{V}, \mathbf{I} \text{ such that } (V_1, I_1, V_2, I_2, \dots, V_{N-2}, I_{N-2}, \mathbf{V}, \mathbf{I}, \mathbf{V}, -\mathbf{I}) \in \mathcal{B}\}.$$

Interconnection of circuits



$$V_N = V_{N'} \quad \text{and} \quad I_N + I_{N'} = 0.$$

Behavior after interconnection:

$$\mathcal{B}_1 \sqcap \mathcal{B}_2$$

$$:= \left\{ (V_1, \dots, V_{N-1}, V_{1'}, \dots, V_{N'-1}, I_1, \dots, I_{N-1}, I_{1'}, \dots, I_{N'-1}) \mid \right.$$

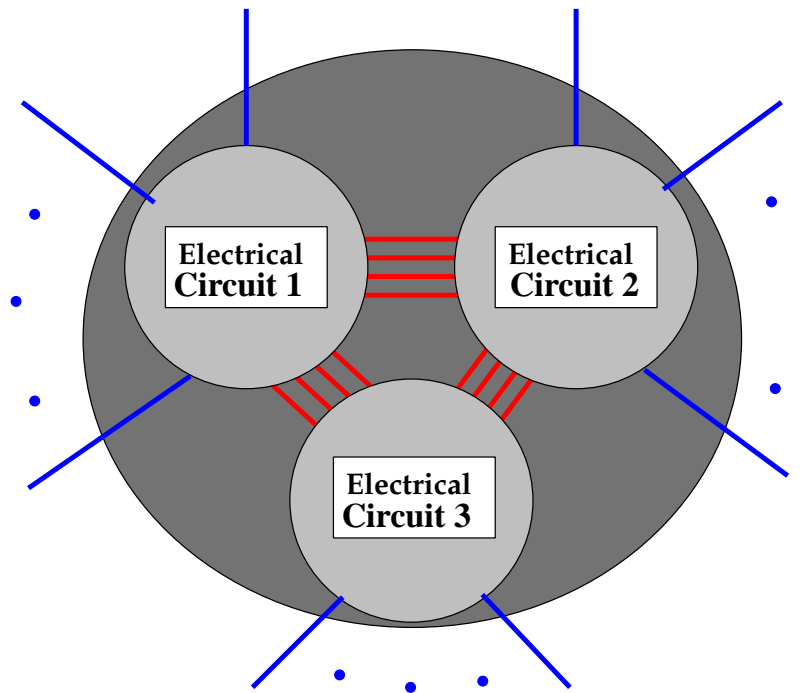
$\exists V, I$ such that

$$(V_1, \dots, V_{N-1}, V, I_1, \dots, I_{N-1}, I) \in \mathcal{B}_1 \quad \text{and}$$

$$(V_{1'}, \dots, V_{N'-1}, V, I_{1'}, \dots, I_{N'-1}, -I) \in \mathcal{B}_2 \}.$$

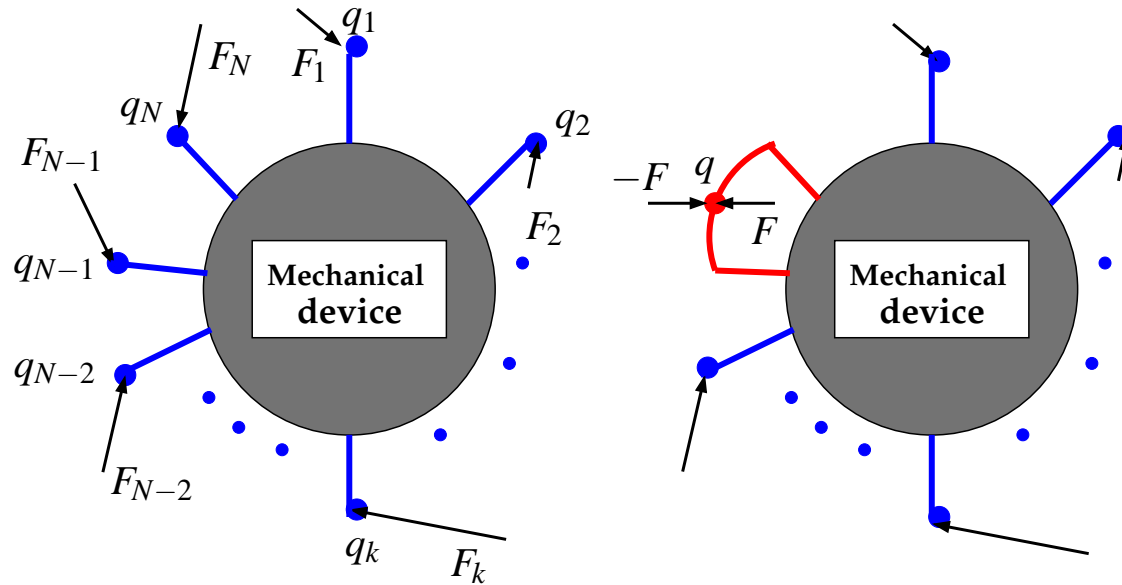
Interconnection of circuits

~> more terminals and more circuits connected



Connection of mechanical terminals

Interconnection = connecting terminals, like screwing pins together.

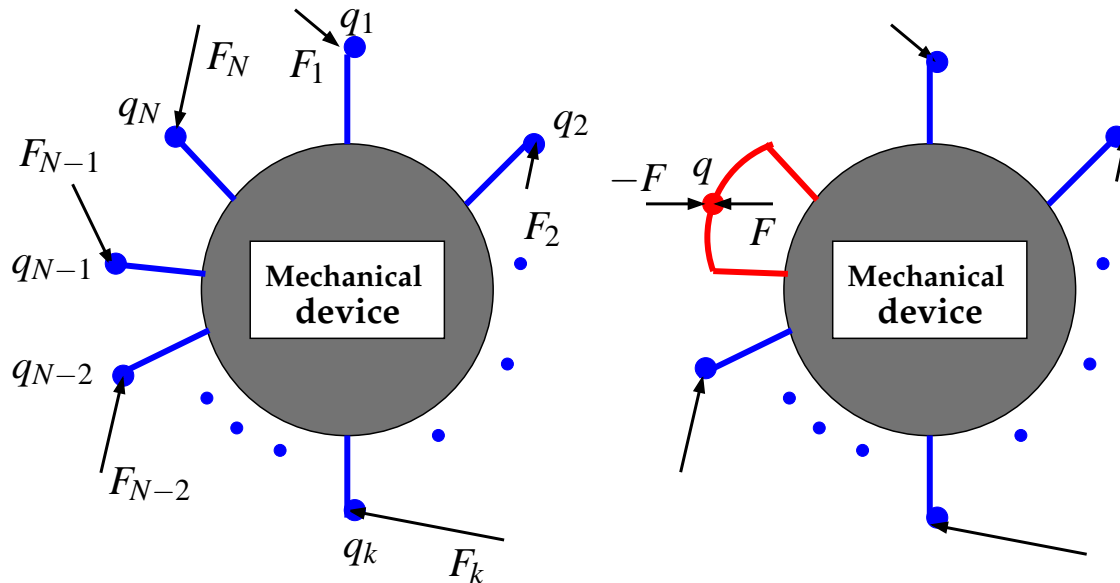


Connecting terminals $N - 1$ and N leads to

$$q_{N-1} = q_N, \quad F_{N-1} + F_N = 0.$$

After interconnection the terminals share the variables q_{N-1}, q_N , and F_{N-1}, F_N (up to a sign).

Connection of mechanical terminals



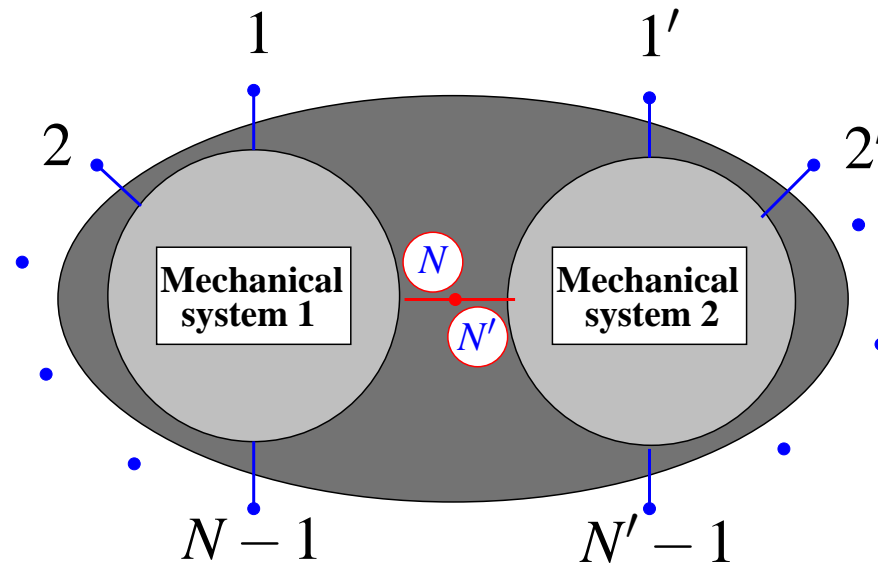
Connecting terminals $N - 1$ and N leads to

$$q_{N-1} = q_N, \quad F_{N-1} + F_N = 0.$$

The interconnected circuit has $N - 2$ terminals. Its behavior =

$$\mathcal{B}' = \{ (q_1, F_1, q_2, F_2, \dots, q_{N-2}, F_{N-2}) : \mathbb{R} \rightarrow \mathbb{R}^{2(N-2)} \mid \exists q, F \text{ such that } (q_1, F_1, q_2, F_2, \dots, q_{N-2}, F_{N-2}, q, F, q, -F) \in \mathcal{B} \}.$$

Interconnection of mechanical systems



$$q_N = q_{N'} \quad \text{and} \quad F_N + F_{N'} = 0.$$

Other terminal types

▶ Thermal systems:

At each terminal: a temperature and a heat flow.

$$T_N = T_{N'} \quad \text{and} \quad Q_N + Q_{N'} = 0.$$

▶ Hydraulic systems:

At each terminal: a pressure and a mass flow.

$$p_N = p_{N'} \quad \text{and} \quad f_N + f_{N'} = 0.$$

▶ ...

Sharing variables

$$V_N = V_{N'} \quad \text{and} \quad I_N + I_{N'} = 0,$$

$$q_N = q_{N'} \quad \text{and} \quad F_N + F_{N'} = 0,$$

$$T_N = T_{N'} \quad \text{and} \quad Q_N + Q_{N'} = 0,$$

$$p_N = p_{N'} \quad \text{and} \quad f_N + f_{N'} = 0,$$

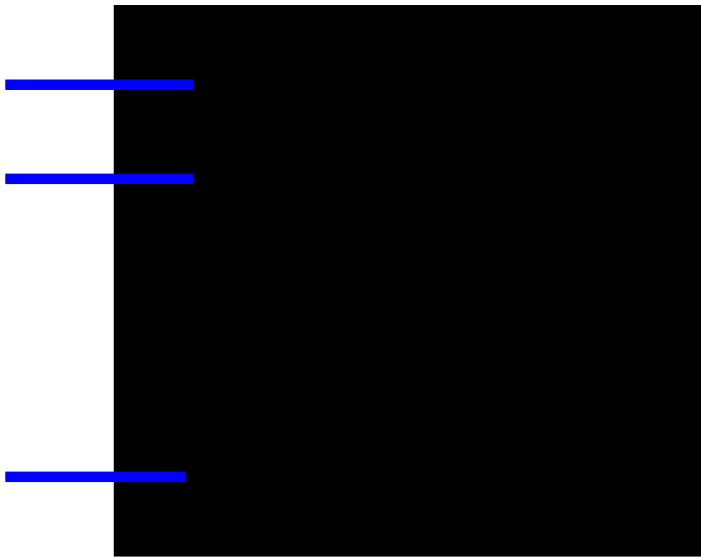
⋮

Interconnection means variable sharing.

Tearing, zooming, and linking

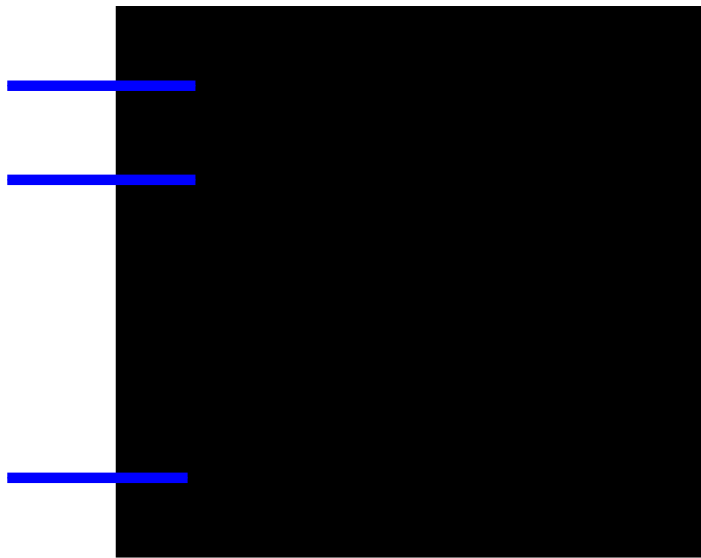
Tearing

∴ Model the behavior of selected variables !!

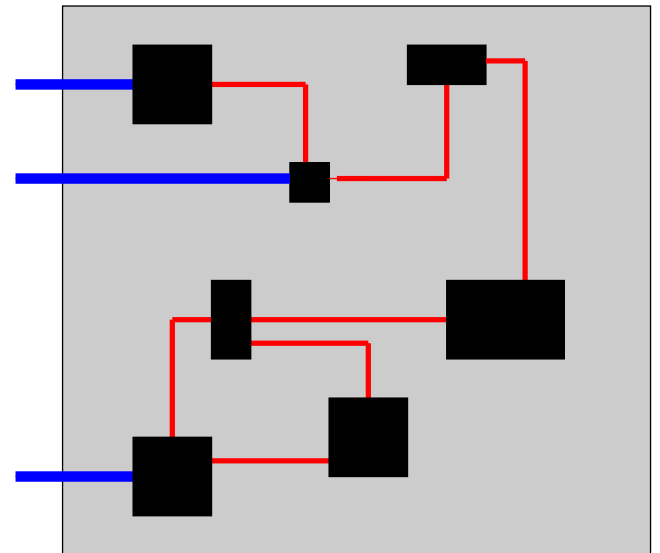


Tearing

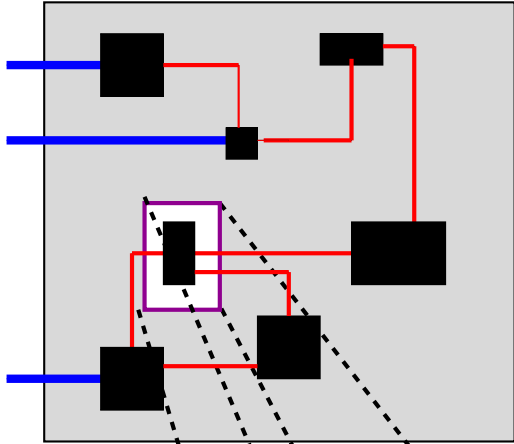
∴ Model the behavior of selected variables !!



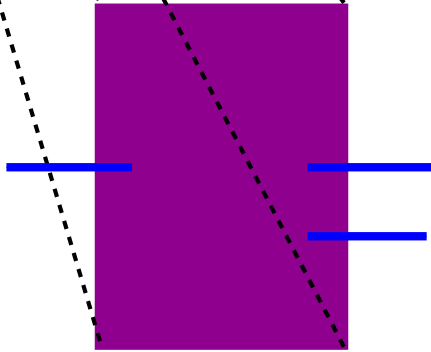
Tear



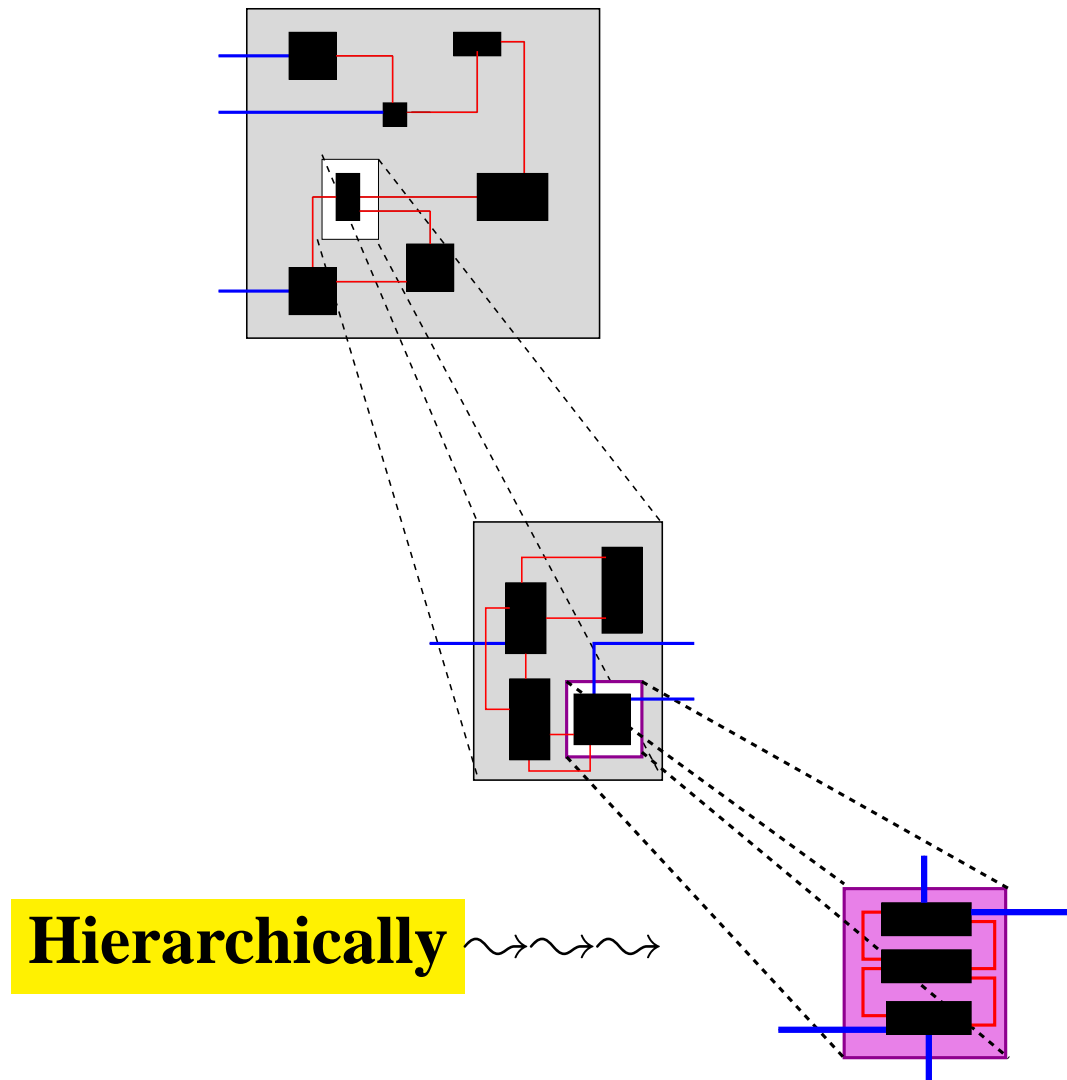
Zooming



Zoom →

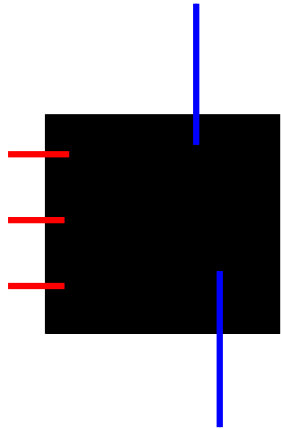
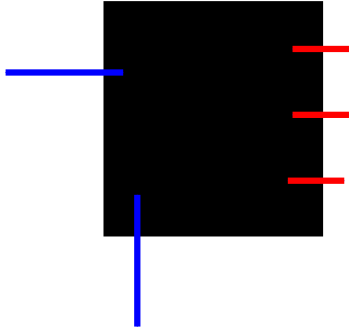


Zooming

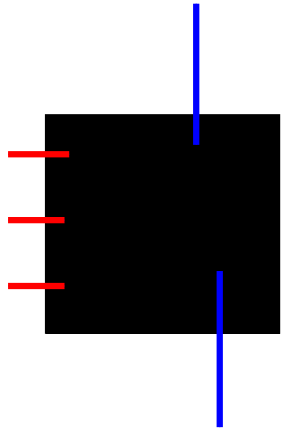
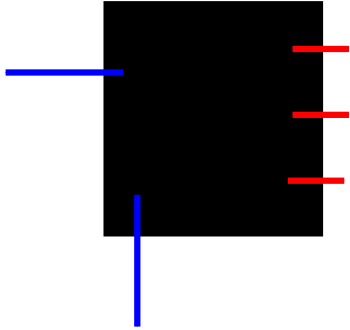


Proceed until subsystems (‘modularity’) are obtained whose model is known, from first principles, or stored in a database.

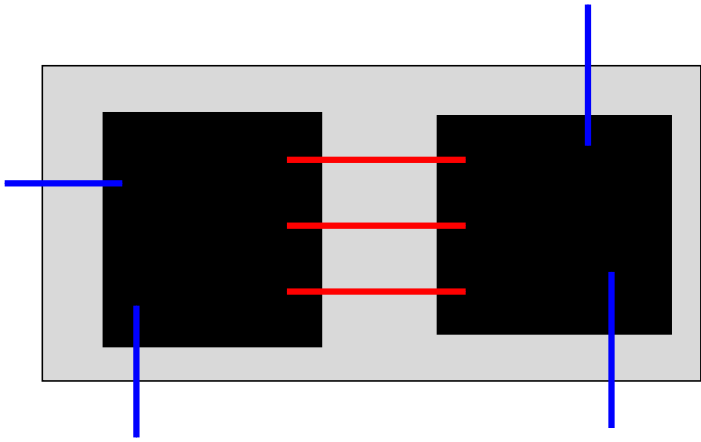
Linking



Linking



Link ~~~~~>



Interconnection architecture

Graph with leaves

A graph with leaves $:\Leftrightarrow$

$$\mathcal{G} = (\mathbb{V}, \mathbb{E}, \mathbb{L}, f_{\mathbb{E}}, f_{\mathbb{L}})$$

\mathbb{V} a finite set of *vertices*,

\mathbb{E} a finite set of *edges*,

\mathbb{L} a finite set of *leaves*,

$f_{\mathbb{E}}$ the edge incidence map,

$f_{\mathbb{L}}$ the leaf incidence map.

$f_{\mathbb{E}}$ maps each element $e \in \mathbb{E}$ into

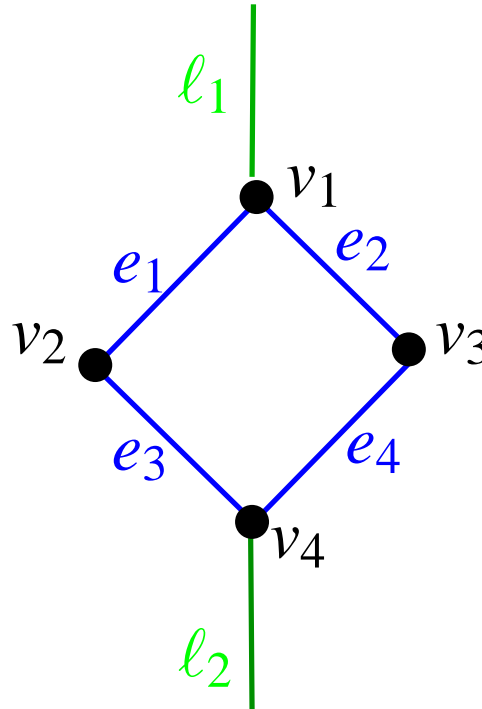
an unordered pair $[v_1, v_2]$, with $v_1, v_2 \in \mathbb{V}$,

$f_{\mathbb{L}}$ is a map from \mathbb{L} to \mathbb{V} , it maps each element

$\ell \in \mathbb{L}$ into an element $v \in \mathbb{V}$.

Graph with leaves

Example:



$$f_{\mathbb{E}} : e_1 \mapsto [v_1, v_2], e_2 \mapsto [v_1, v_3], e_3 \mapsto [v_2, v_4], e_4 \mapsto [v_3, v_4].$$

$$f_{\mathbb{L}} : l_1 \mapsto v_1, l_2 \mapsto v_4.$$

Formalization of interconnected system

An interconnected system is identified with a graph with leaves

$$\mathcal{G} = (\mathbb{V}, \mathbb{E}, \mathbb{L}, f_{\mathbb{E}}, f_{\mathbb{L}}).$$

The vertices \leftrightarrow **subsystems**
The edges \leftrightarrow **connections,**
The leaves \leftrightarrow **external terminals.**

Model specification

A model is obtained as follows.

- ▶ **For each subsystem, specify the behavior of the variables on its terminals, i.e. on the edges and the leaves that are incident to the vertex corresponding to the subsystem.**
- ▶ **For each connection, specify the sharing variable conditions on the connected terminals. I.e., for each edge, specify the interconnection constraints on the variables of the subsystem terminals that correspond to the edges.**
- ▶ **Specify the manifest variables.**

Subsystems in the vertices.

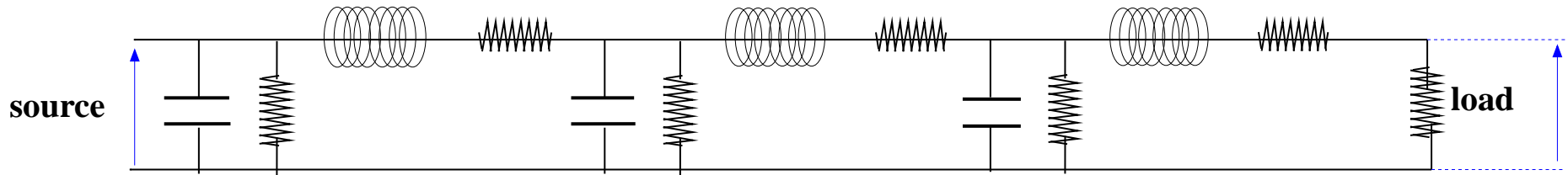
Connections in the edges.

External terminals in the leaves.

Example

A transmission line

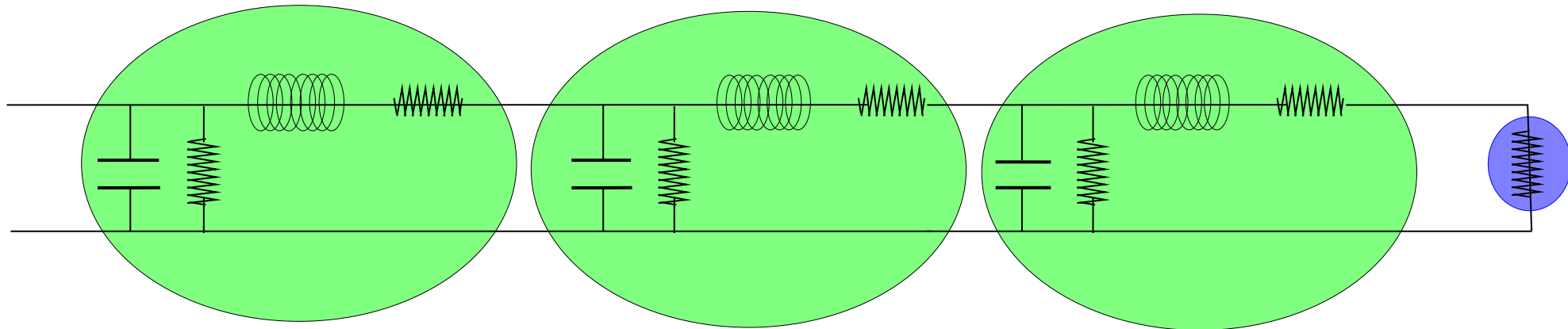
Consider the transmission line shown below.



The aim is to model the relation between the voltage of the source on the left and the voltage across the load on the right.

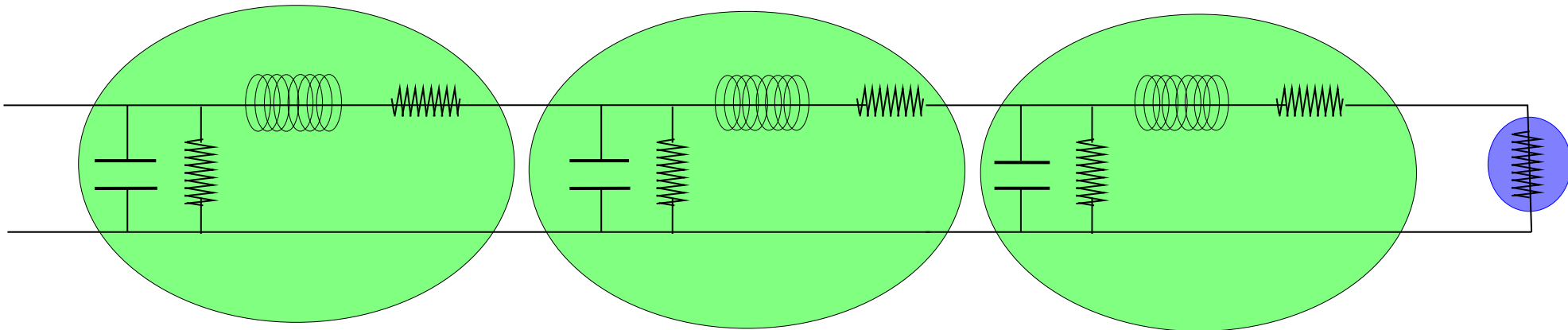
A transmission line

View the system as an interconnection of 4 subsystems.

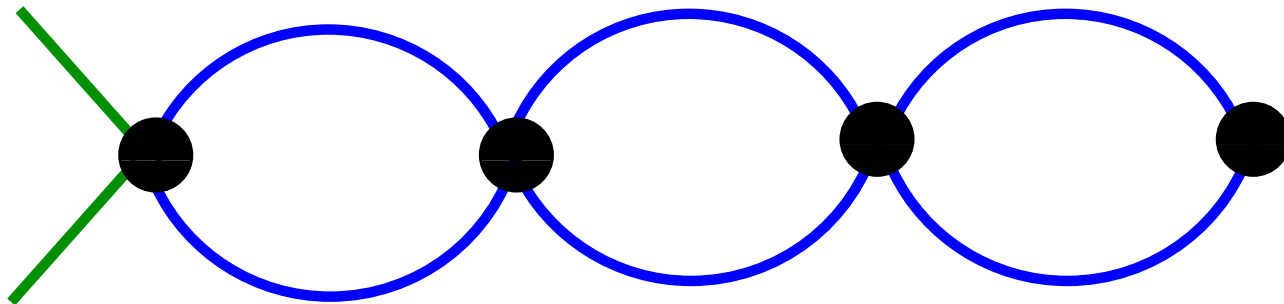


A transmission line

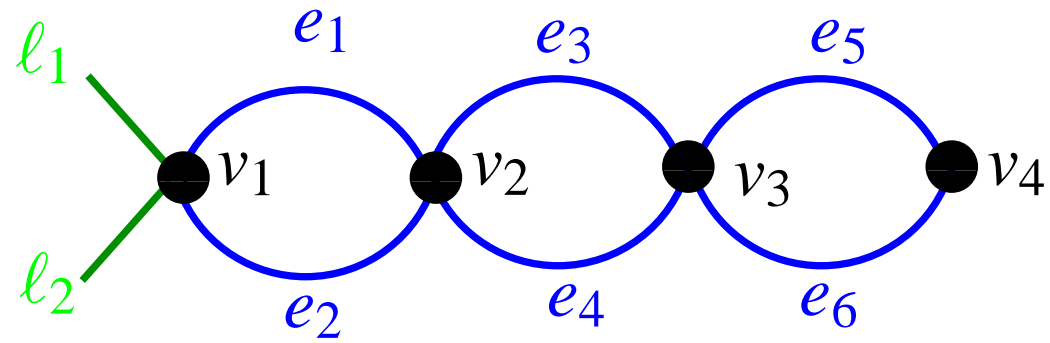
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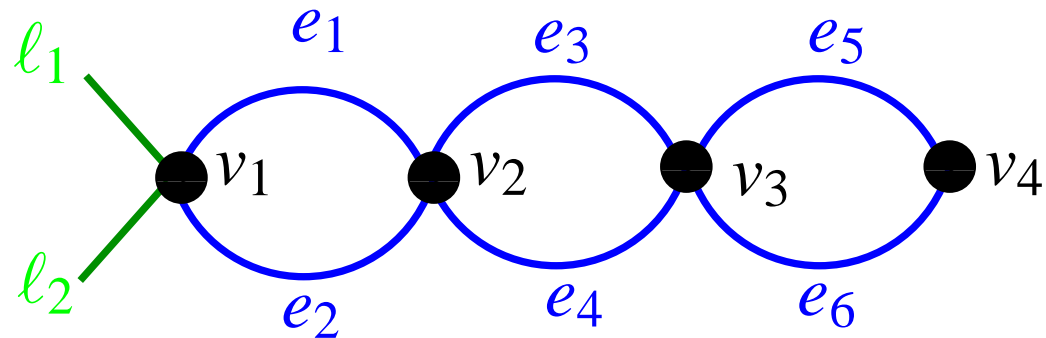
The interconnection architecture \rightsquigarrow the graph with leaves



A transmission line

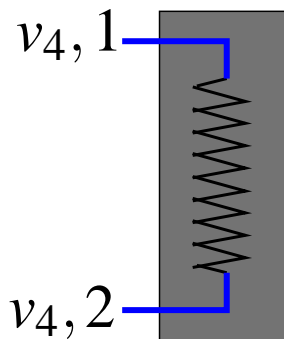


A transmission line



**In vertices v_1, v_2, v_3 we have identical subsystems.
We deal with them later.**

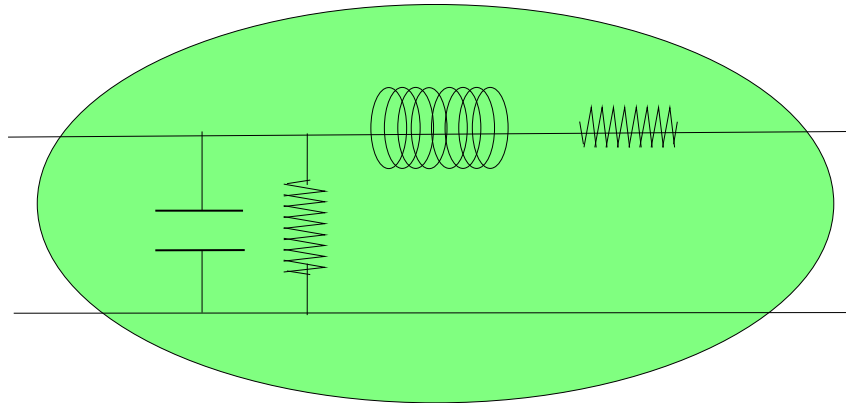
In vertex v_4 there is a resistor \rightsquigarrow



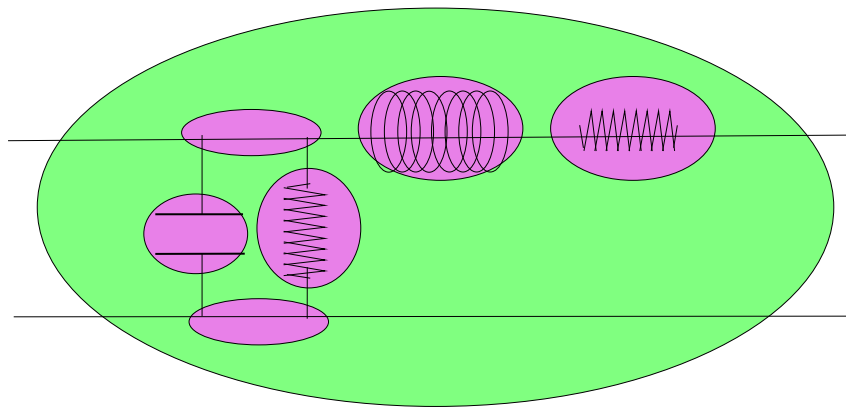
$$V_{v_{4,1}} - V_{v_{4,2}} = R I_{v_{4,1}}, \quad I_{v_{4,1}} + I_{v_{4,2}} = 0.$$

A transmission line section

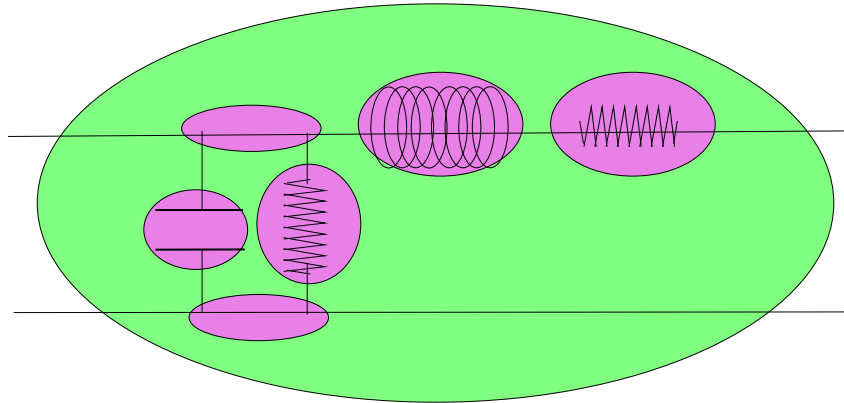
In each of the vertices v_1, v_2, v_3 we have:



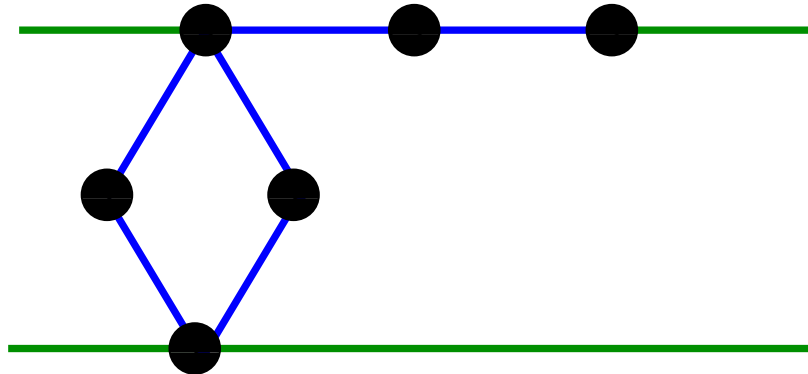
This system can be viewed as the interconnection of 6 subsystems:



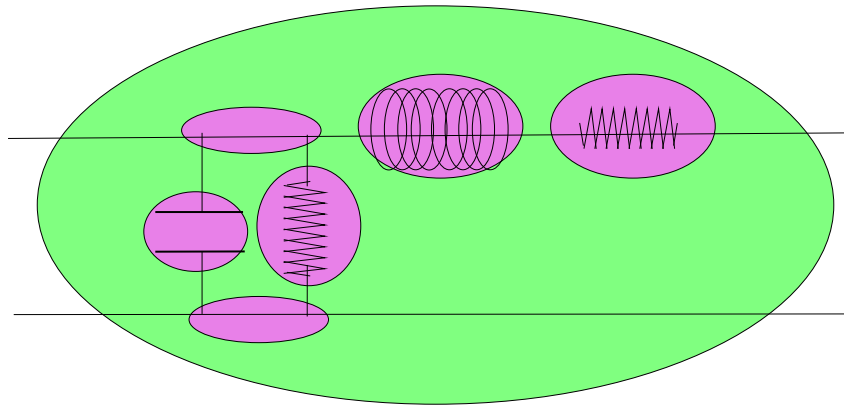
A transmission line section



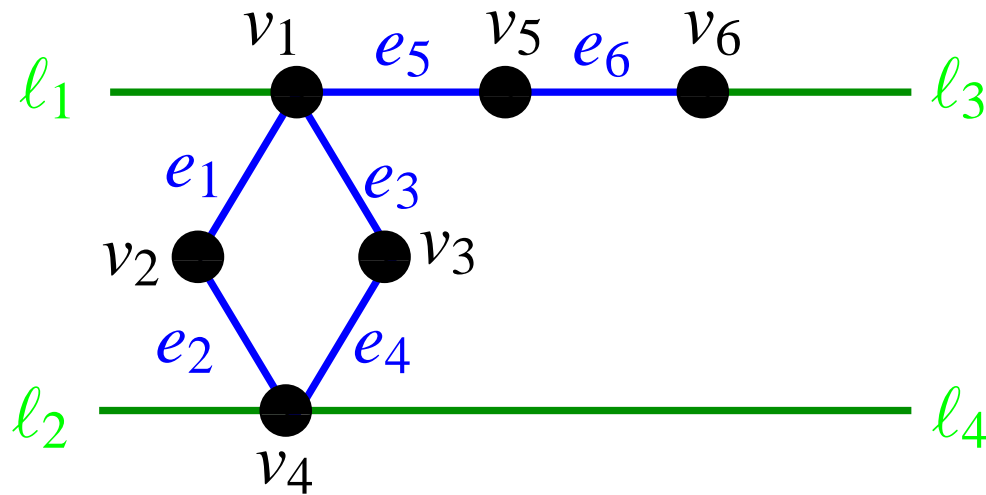
The associated interconnection architecture is



A transmission line section



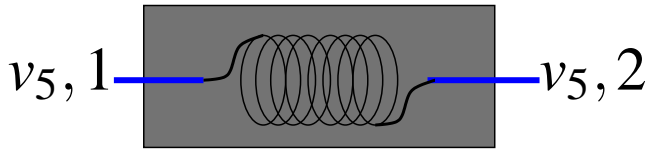
The associated interconnection architecture is



Modeling the transmission line section

Subsystems: 2 resistors, 1 inductor, 1 capacitor, 2 connectors.

The inductor in vertex v_5 , for example, \rightsquigarrow

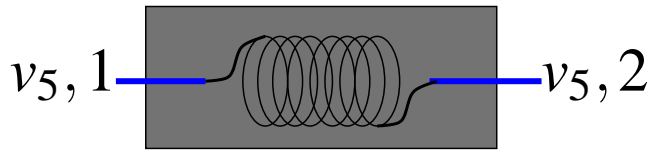


$$V_{v_{5,1}} - V_{v_{5,2}} = L \frac{d}{dt} I_{v_{5,1}}, \quad I_{v_{5,1}} + I_{v_{5,2}} = 0.$$

Modeling the transmission line section

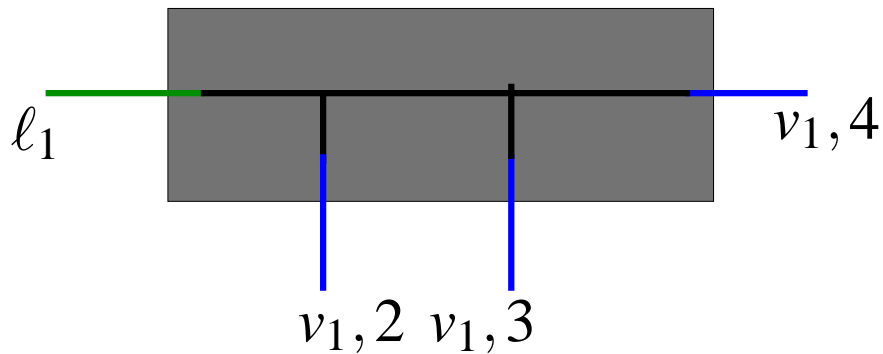
Subsystems: 2 resistors, 1 inductor, 1 capacitor, 2 connectors.

The inductor in vertex v_5 , for example, \rightsquigarrow



$$V_{v_5,1} - V_{v_5,2} = L \frac{d}{dt} I_{v_5,1}, \quad I_{v_5,1} + I_{v_5,2} = 0.$$

The connector in vertex v_1 , for example, \rightsquigarrow



$$V_{l_1} = V_{v_1,2} = V_{v_1,3} = V_{v_1,4}, \quad I_{l_1} + I_{v_1,2} + I_{v_1,3} + I_{v_1,4} = 0.$$

For each vertex we obtain a set of equations.

Modeling the transmission line section

The connection equations for edge e_5 , for example, \rightsquigarrow

$$V_{v_1,4} = V_{v_5,1}, \quad I_{v_1,4} + I_{v_5,1} = 0.$$

For each edge we obtain two such equations.

Modeling the transmission line section

**For each vertex and for each edge we obtain a set of equations.
The manifest variables for this subsystem are**

$$V_{l_1}, I_{l_1}, V_{l_2}, I_{l_2}, V_{l_3}, I_{l_3}, V_{l_4}, I_{l_4}.$$

The latent variables are

$$V_{v_1,1}, I_{v_1,1}, \dots, V_{v_6,1}, I_{v_6,1}.$$

Modeling the transmission line section

**For each vertex and for each edge we obtain a set of equations.
The manifest variables for this subsystem are**

$$V_{l_1}, I_{l_1}, V_{l_2}, I_{l_2}, V_{l_3}, I_{l_3}, V_{l_4}, I_{l_4}.$$

The latent variables are

$$V_{v_1,1}, I_{v_1,1}, \dots, V_{v_6,1}, I_{v_6,1}.$$

Eliminating the latent variables

\rightsquigarrow a LTIDS with (4) ODEs in the variables

$$V_{l_1}, I_{l_1}, V_{l_2}, I_{l_2}, V_{l_3}, I_{l_3}, V_{l_4}, I_{l_4}.$$

Modeling the transmission line section

Denote these equations as

$$R \left(\frac{d}{dt} \right) \begin{bmatrix} V_{l_1} \\ I_{l_1} \\ V_{l_2} \\ I_{l_2} \\ V_{l_3} \\ I_{l_3} \\ V_{l_4} \\ I_{l_4} \end{bmatrix} = 0.$$

Note: For RLC-circuits, as is the case here, there are more efficient ways to arrive at these equations (see Lecture IV).

A transmission line

Going back to the transmission line yields the subsystem equations

$$R \left(\frac{d}{dt} \right) \begin{bmatrix} \begin{bmatrix} V_{\ell_1} \\ I_{\ell_1} \\ V_{v_1,2} \\ I_{v_1,2} \\ V_{v_1,3} \\ I_{v_1,3} \\ V_{v_1,4} \\ I_{v_1,4} \end{bmatrix} \\ \begin{bmatrix} V_{v_2,1} \\ I_{v_2,2} \\ V_{v_2,2} \\ I_{v_2,2} \\ V_{v_2,3} \\ I_{v_2,3} \\ V_{v_2,4} \\ I_{v_2,4} \end{bmatrix} \\ \begin{bmatrix} V_{v_3,1} \\ I_{v_3,2} \\ V_{v_3,2} \\ I_{v_3,2} \\ V_{v_3,3} \\ I_{v_3,3} \\ V_{v_3,4} \\ I_{v_3,4} \end{bmatrix} \end{bmatrix} = 0,$$

$$V_{v_4,1} - V_{v_4,2} = R I_{v_4,1}, \quad I_{v_4,1} + I_{v_4,2} = 0,$$

A transmission line

and the interconnection equations

$$\begin{aligned}V_{v_1,3} &= V_{v_2,1}, & I_{v_1,3} + I_{v_2,1} &= 0, \\V_{v_1,4} &= V_{v_2,2}, & I_{v_1,4} + I_{v_2,2} &= 0, \\V_{v_2,3} &= V_{v_3,1}, & I_{v_2,3} + I_{v_3,1} &= 0, \\V_{v_2,4} &= V_{v_3,2}, & I_{v_2,4} + I_{v_3,2} &= 0, \\V_{v_3,3} &= V_{v_4,1}, & I_{v_4,3} + I_{v_4,1} &= 0, \\V_{v_3,4} &= V_{v_4,2}, & I_{v_3,4} + I_{v_4,2} &= 0.\end{aligned}$$

A transmission line

and the interconnection equations

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Finally, there is the manifest variable assignment

$$w_1 = V_{l_1} - V_{l_2}, \quad w_2 = V_{v_4,1} - V_{v_4,2}.$$

A transmission line

After elimination of the latent variables, we obtain the desired differential equation that describes the behavior of (w_1, w_2)

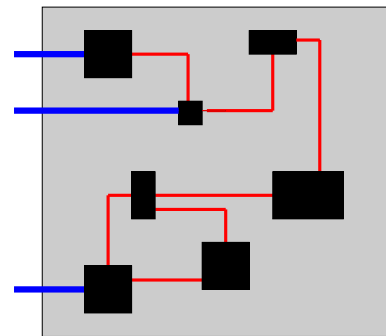
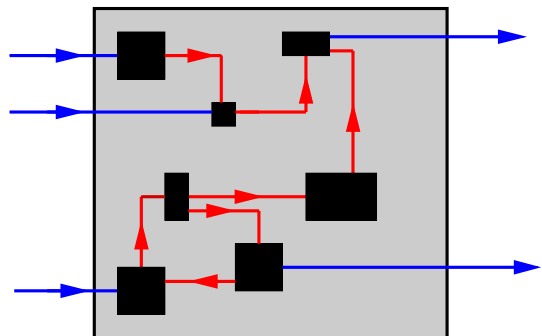
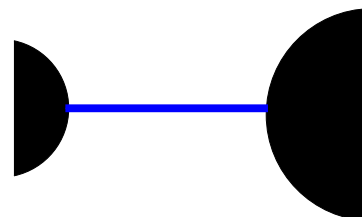
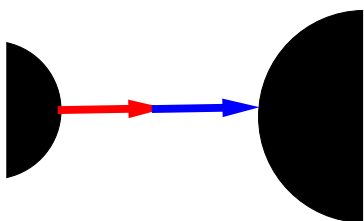
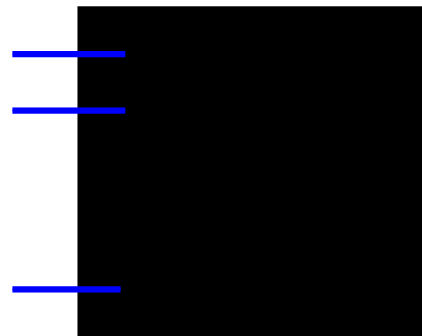
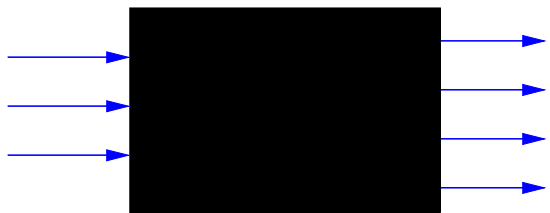
$$r_1 \left(\frac{d}{dt} \right) w_1 = r_2 \left(\frac{d}{dt} \right) w_2.$$

In practice, all these steps need to be carried out more explicitly, faster, better, and more reliably with the help of software and a computer toolbox.

Input/output

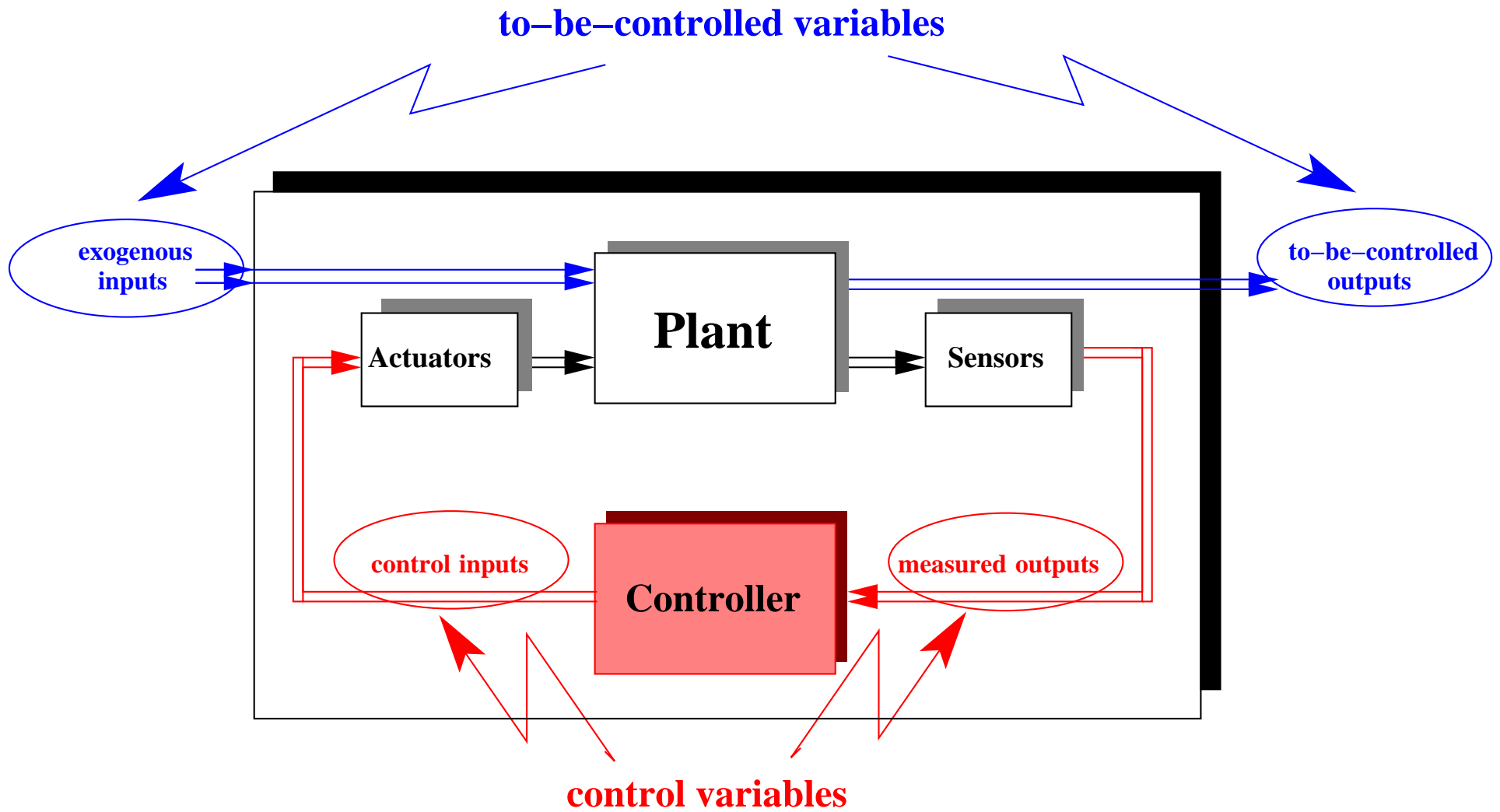
versus

tearing, zooming, and linking

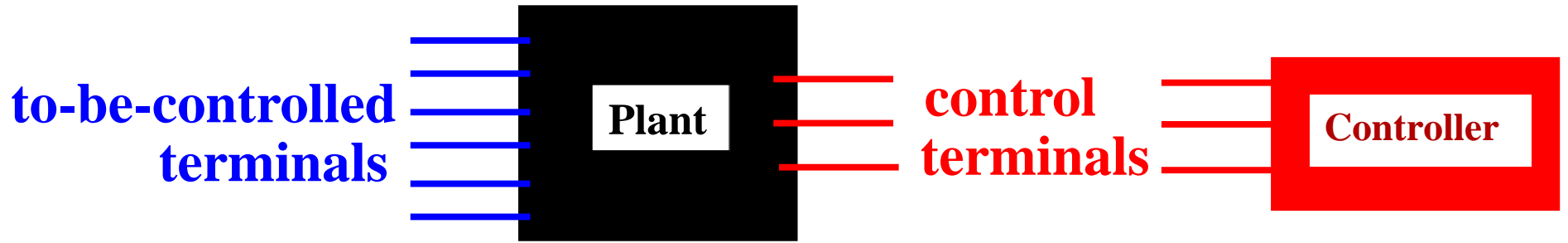


Control as interconnection

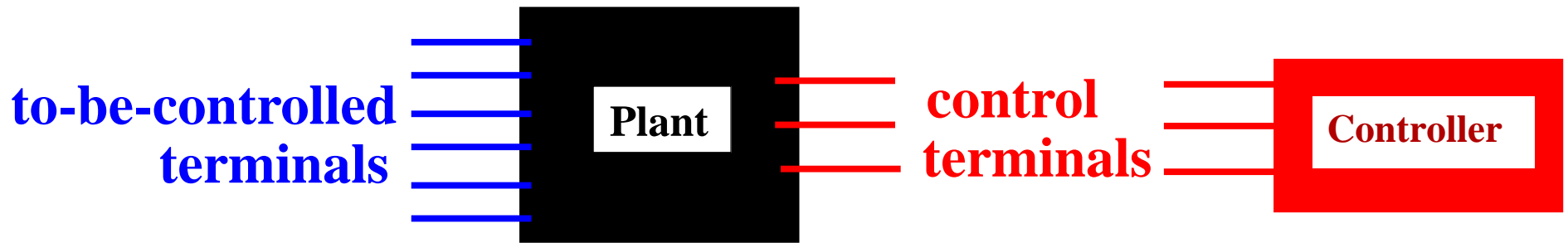
Feedback control



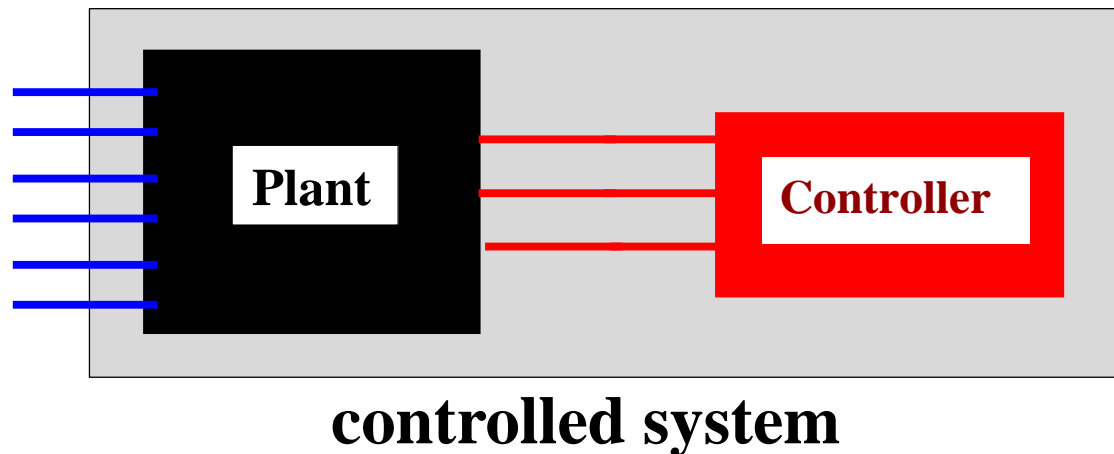
Behavioral control



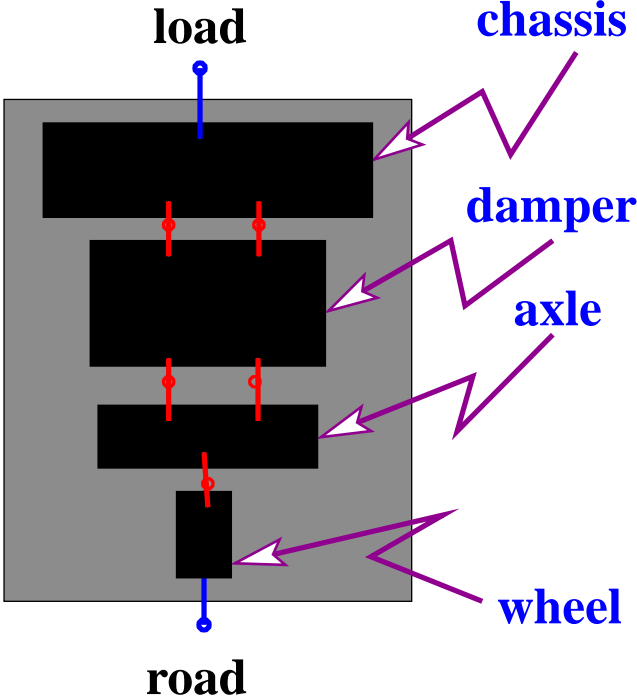
Behavioral control



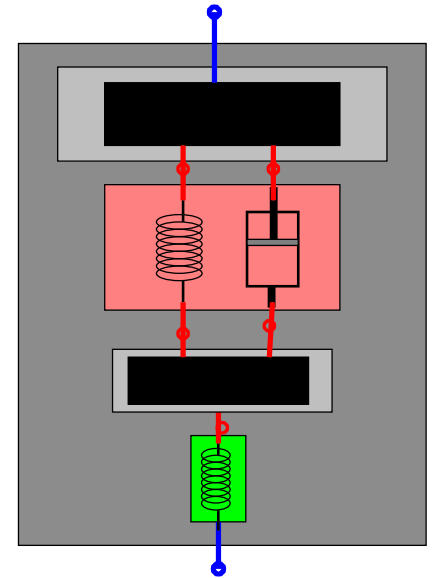
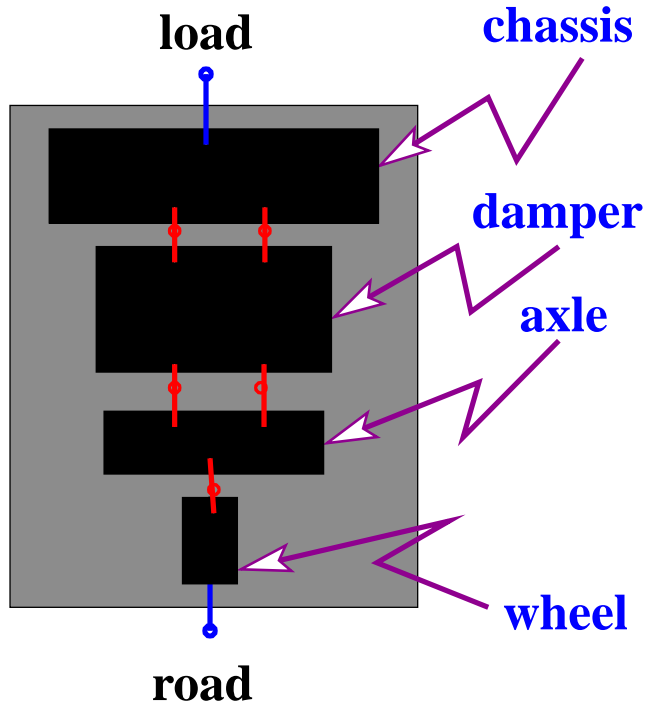
control = interconnection.



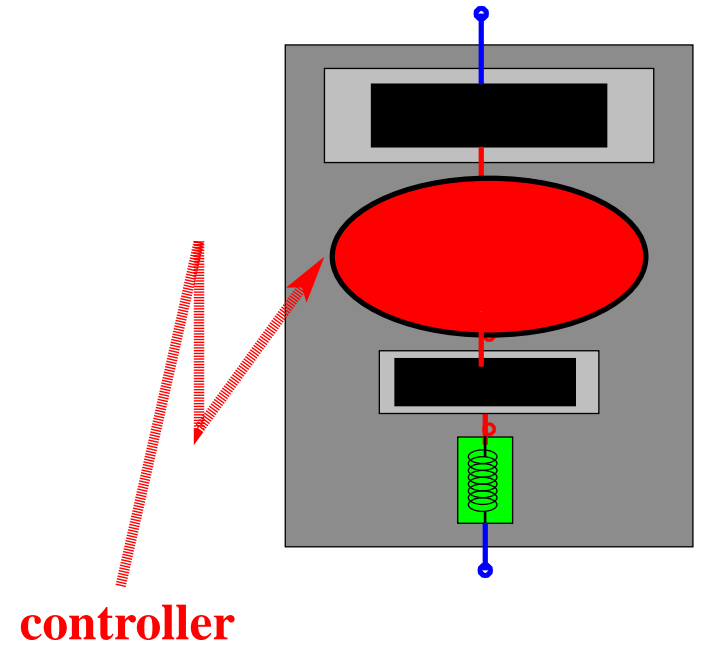
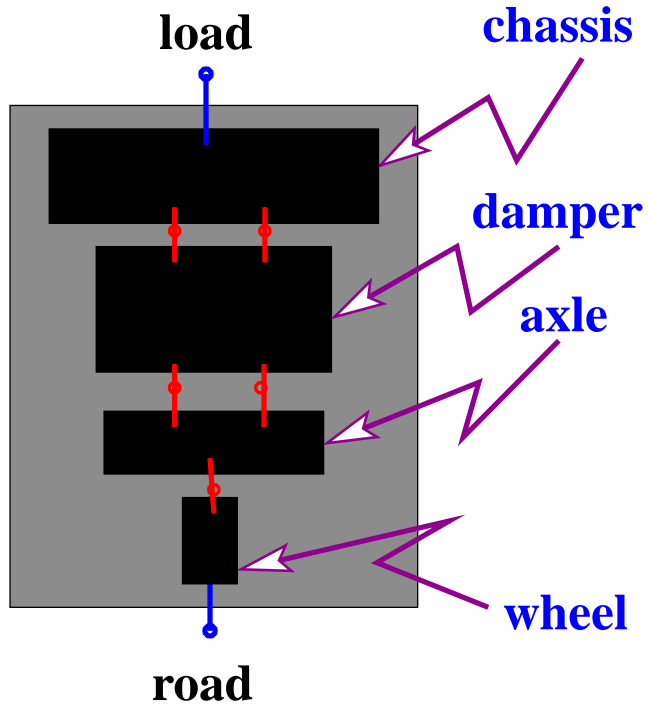
A 'quarter car'



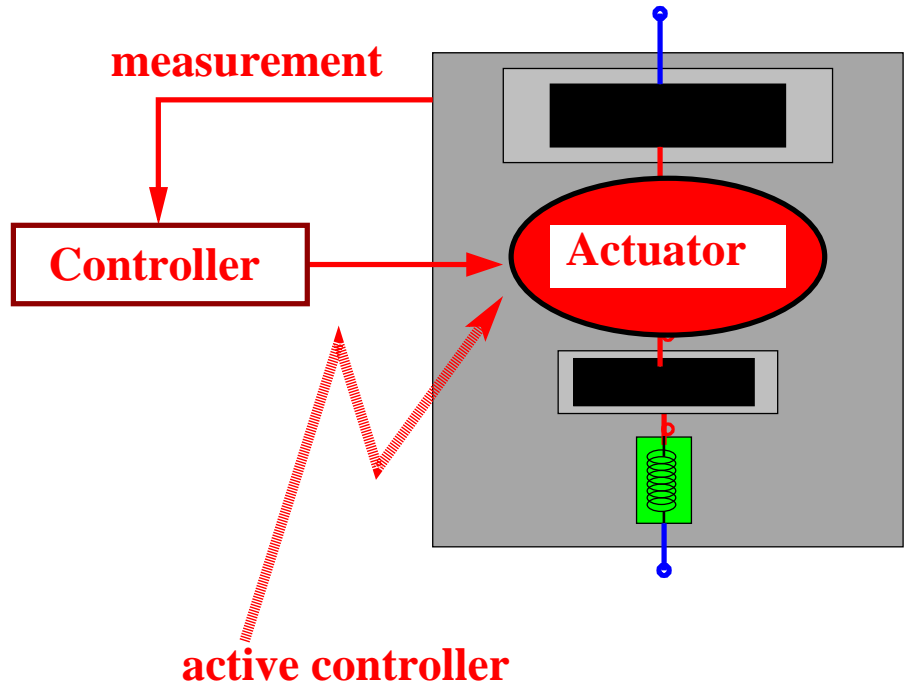
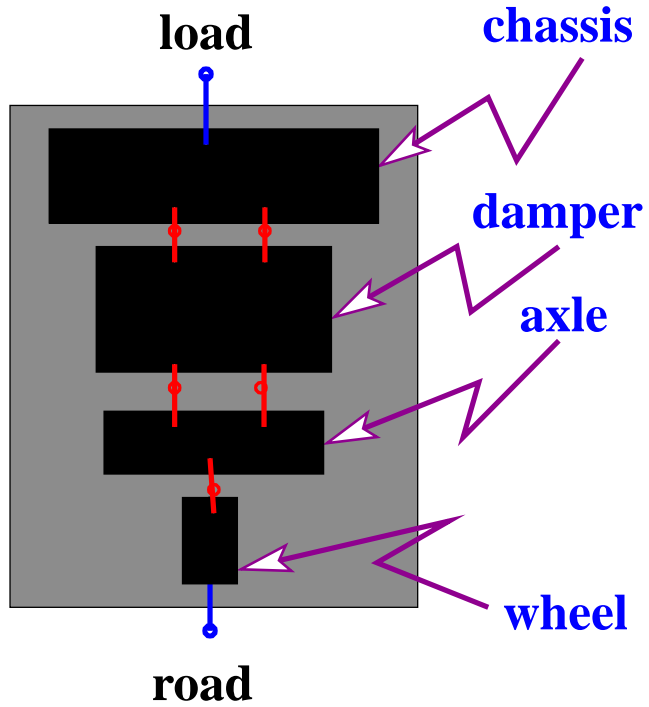
A 'quarter car'



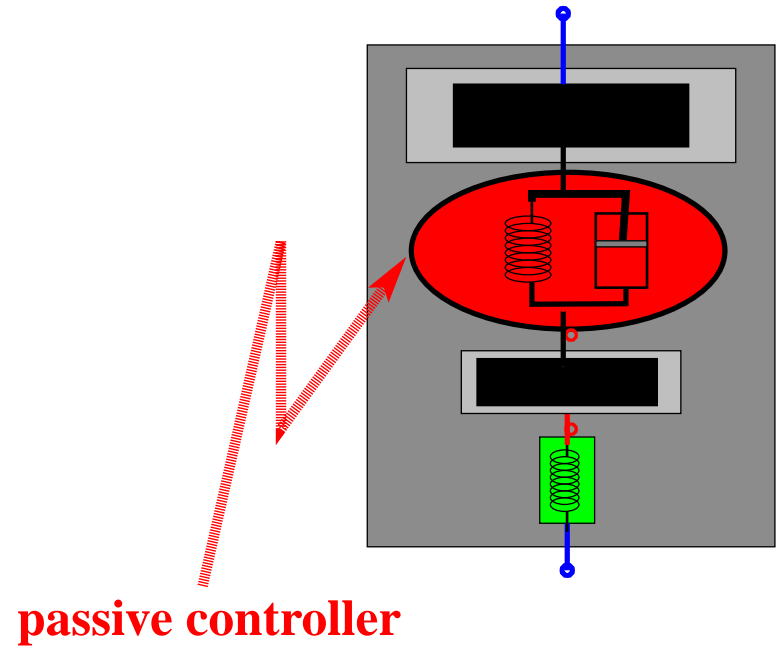
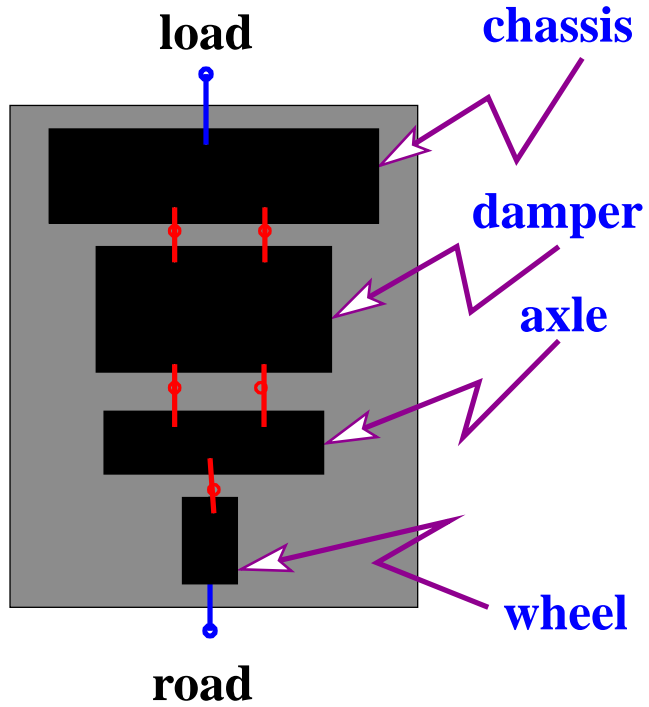
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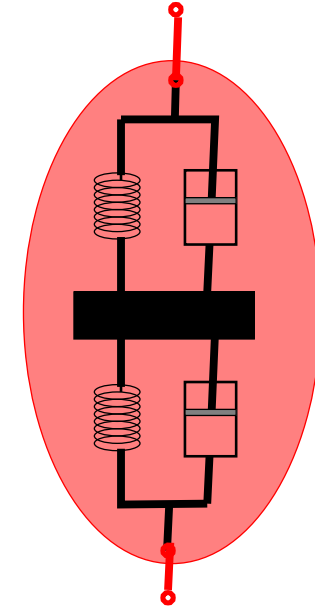
Suspension controllers in Formula 1



Nigel Mansell victorious in 1992 with an active damper suspension.

Active dampers were banned in 1994 to break the dominance of the Williams team.

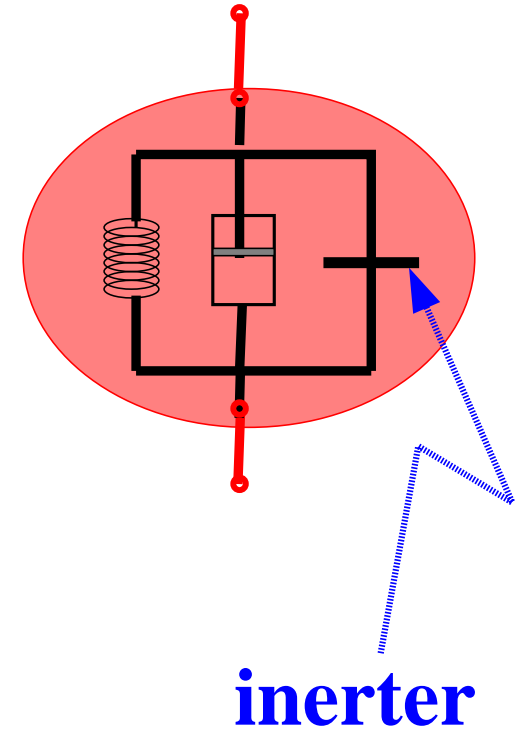
Suspension controllers in Formula 1



Renault successfully uses a passive ‘tuned mass damper’ in 2005/2006.

Banned in 2006, under the ‘movable aerodynamic devices’ clause.

Suspension controllers in Formula 1

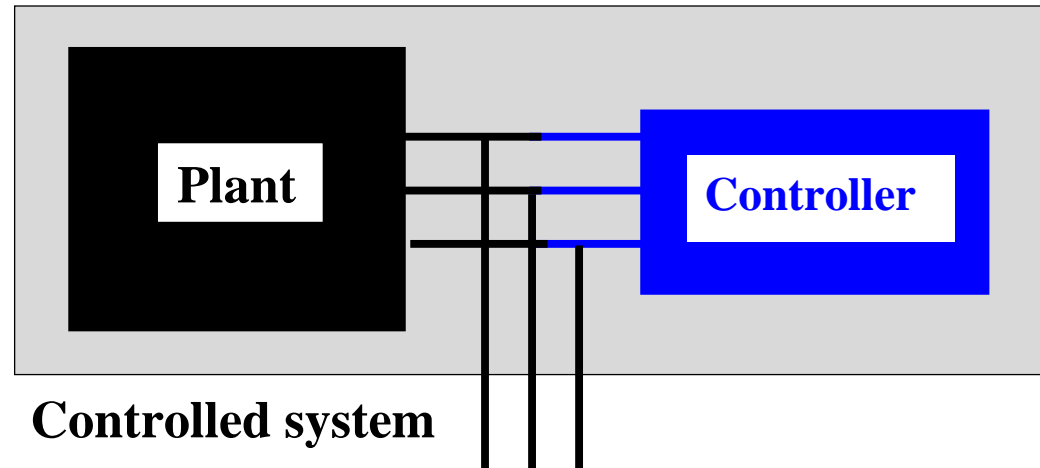


Kimi Räikkönen wins the 2005 Grand Prix in Spain with McLaren's 'J-damper' (see Lecture V).

Pole placement and stabilization

LTIDSs as controllers

We consider only ‘full control’.



and controllers such that the controlled system is autonomous.

LTIDSs as controllers

Consider a LTIDS $\mathcal{P} \in \mathcal{L}^w$, the *plant*,
and a LTIDS $\mathcal{C} \in \mathcal{L}^w$ the *controller*.

[[\mathcal{C} is called a *regular controller* for \mathcal{P}]] $:\Leftrightarrow$

[[(i) $p(\mathcal{C}) = m(\mathcal{P})$ and (ii) $\mathcal{P} \cap \mathcal{C}$ is autonomous]].

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In terms of minimal kernel representations

$$\begin{aligned} P \left(\frac{d}{dt} \right) w &= 0 \\ C \left(\frac{d}{dt} \right) w &= 0 \end{aligned}$$

for \mathcal{P} and \mathcal{C} , regularity $\Leftrightarrow \begin{bmatrix} P \\ C \end{bmatrix}$ square and nonsingular.

The characteristic polynomial

Let $\mathcal{B} \in \mathcal{L}^w$ be autonomous, and $R \left(\frac{d}{dt} \right) w = 0$ be a minimal kernel representation of \mathcal{B} .

Assume that $\text{determinant}(R)$ is monic (otherwise change $R \rightarrow \alpha R$ with $0 \neq \alpha \in \mathbb{R}$ suitably chosen).

The **characteristic polynomial** of \mathcal{B} , $\chi_{\mathcal{B}} \in \mathbb{R}[\xi]$, is defined as

$$\chi_{\mathcal{B}} := \text{determinant}(R).$$

Note: $\chi_{\mathcal{B}}$ is independent of the R chosen to represent \mathcal{B} .

The characteristic polynomial

Let $\mathcal{B} \in \mathcal{L}^w$ be autonomous, and $R \left(\frac{d}{dt} \right) w = 0$ be a minimal kernel representation of \mathcal{B} .

The **minimal polynomial** of \mathcal{B} , $\mu_{\mathcal{B}} \in \mathbb{R}[\xi]$, is defined as the monic polynomial of least degree that annihilates \mathcal{B} , i.e.

$$\mu_{\mathcal{B}} \left(\frac{d}{dt} \right) \mathcal{B} = 0,$$

$$\left[p_{\mathcal{B}} \left(\frac{d}{dt} \right) \mathcal{B} = 0, 0 \neq p \in \mathbb{R}[\xi] \right] \Rightarrow \left[\text{degree}(p) \geq \text{degree}(\mu_{\mathcal{B}}) \right].$$

Note: $\mu_{\mathcal{B}}$ is a factor of $\chi_{\mathcal{B}}$.

For \mathcal{B} described by $\frac{d}{dt}x = Ax, w = Cx$ with (A, C) observable, $\chi_{\mathcal{B}}$ and $\mu_{\mathcal{B}}$ are equal to the characteristic and minimal polynomial of A .

Pole placement

Theorem

Let $\mathcal{P} \in \mathcal{L}^w$ be controllable. Then for all monic polynomials $\pi \in \mathbb{R}[\xi]$, there exists a regular controller $\mathcal{C} \in \mathcal{L}^w$ such that

$$\chi_{\mathcal{P}\mathcal{C}} = \pi.$$

Pole placement

Theorem

Let $\mathcal{P} \in \mathcal{L}^w$ be controllable. Then for all monic polynomials $\pi \in \mathbb{R}[\xi]$, there exists a regular controller $\mathcal{C} \in \mathcal{L}^w$ such that

$$\chi_{\mathcal{P} \cap \mathcal{C}} = \pi.$$

Proof: Let $P \left(\frac{d}{dt} \right) w = 0$ be a minimal kernel representation of \mathcal{P} . The Smith form yields $P = U \begin{bmatrix} 0_{p(\mathcal{P}) \times m(\mathcal{P})} & I_{p(\mathcal{P}) \times p(\mathcal{P})} \end{bmatrix} V$, with U and V unimodular. Take for \mathcal{C} the system $C \left(\frac{d}{dt} \right) w = 0$, with $C = \begin{bmatrix} \mathbf{diag}(\pi, 1, \dots, 1) & 0_{m(\mathcal{P}) \times p(\mathcal{P})} \end{bmatrix} V$.

Pole placement

The above proof can be refined in numerous directions (see Exercise IV.5).

In particular, we obtain

Theorem

Let $\mathcal{P} \in \mathcal{L}^w$ be controllable. Then for all monic polynomials $v \in \mathbb{R}[\xi]$, there exists a regular controller $\mathcal{C} \in \mathcal{L}^w$ such that

$$\mu_{\mathcal{P}n\mathcal{C}} = v.$$

Stabilization

and

Theorem

Let $\mathcal{P} \in \mathcal{L}^w$ be stabilizable. Then there exists a regular controller $\mathcal{C} \in \mathcal{L}^w$ such that the autonomous system

$\mathcal{P} \cap \mathcal{C}$ is stable.

Recapitulation

Summary

- ▶ **Complex systems consist of interconnections of subsystems.**

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- ▶ **Input/output thinking and signal flow graphs provide a very limited view of interconnected physical systems.**

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Summary

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- ▶ **Control as interconnection is an effective way of thinking about control, with feedback control as a very useful special case.**
- ▶ **Controllable LTIDSs allow regular controllers that achieve an arbitrary characteristic polynomial for the controlled system. Stabilizable LTIDSs can be stabilized by regular controllers.**

End of Lecture VIII