

# Lecture 4a

Wednesday 04-02-2009

09.00-10.30

## Rational symbols

Lecturer: Jan C. Willems

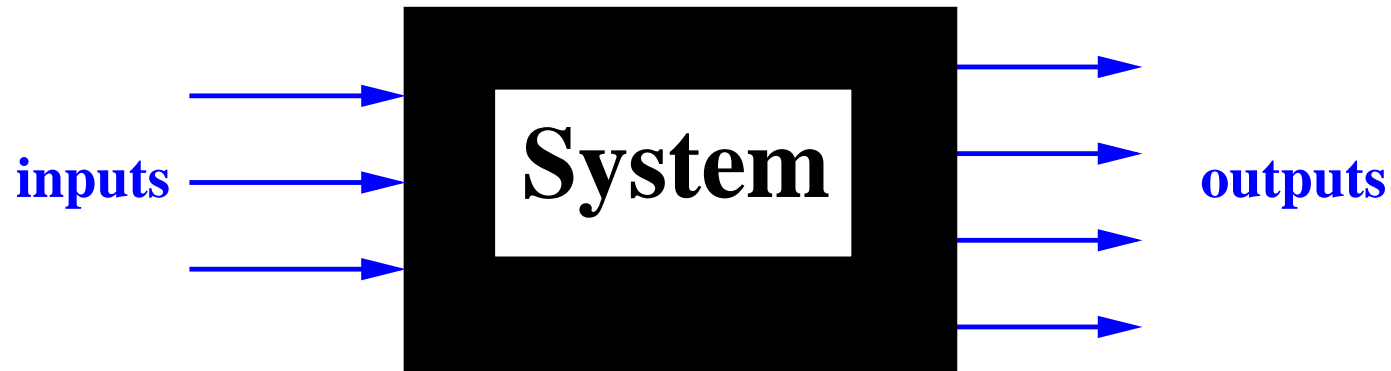
## Outline

- ▶ **Behaviors defined by rational symbols**
- ▶ **Norm preserving representations**
- ▶ **The gap between LITDSs**
- ▶ **Model reduction without stability or i/o partition**

# Introduction

## Theme

In system theory, it is customary to think of dynamical models in terms of inputs and outputs, viz.



often with transfer functions

$$y = F(s)u$$

$F$  a matrix of rational transfer functions.

## Theme

$$y = F(s)u$$

In the present lecture, we will

- ▶ for good physical and system theoretic reasons, not use an input/output partition

↷ system variables

$$w = \begin{bmatrix} u \\ y \end{bmatrix}$$

- ▶ interpret  $F$ , not in terms of Laplace transforms, but in terms of differential equations.

**Important for pedagogical reasons, among other things.**

## Reminder

**LTIDSs:**  $(\mathbb{R}, \mathbb{R}^w, \mathcal{B})$  where

- $T = \mathbb{R}$  ‘time’
- $W = \mathbb{R}^w$  ‘signal space’
- and ‘behavior’  $\mathcal{B} =$  the set of solutions of a system of

**linear constant coefficient ODEs**

$\mathcal{B} =$  the  $\mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w)$ -solutions of

$$R_0 w + R_1 \frac{d}{dt} w + \cdots + R_L \frac{d^L}{dt^L} w = 0, \quad R_0, R_1, \dots \text{ matrices}$$

**Polynomial matrix notation**  $\rightsquigarrow R \left( \frac{d}{dt} \right) w = 0$

$$R \in \mathbb{R}[\xi]^{\bullet \times w}, \quad R(\xi) = R_0 + R_1 \xi + \cdots + R_L \xi^L$$

# Representations of LTIDSs

**Behaviors of LTIDSs allow many useful representations**

▶ **As the set of solutions of  $R \left( \frac{d}{dt} \right) w = 0$   $R \in \mathbb{R} [\xi]^{\bullet \times w}$**

▶ **With input/output partition**

▶ **Input/state/output representation**

$\exists$  **matrices  $A, B, C, D$  such that**

**$\mathcal{B}$  consists of all  $w$ 's generated by**

$$\frac{d}{dt}x = Ax + Bu, \quad y = Cx + Du \quad w \cong \begin{bmatrix} u \\ y \end{bmatrix}$$

▶ ...

▶ **with rational symbols  $\rightsquigarrow$  this lecture**

# **Rational symbols**



## ODEs with rational symbols

Defining what a solution is for ODEs such as

$$R\left(\frac{d}{dt}\right)w = 0 \quad \text{or} \quad \frac{d}{dt}x = Ax + Bu, \quad y = Cx + Du, \quad w = \begin{bmatrix} u \\ y \end{bmatrix}$$

poses no difficulties worth mentioning, but rational functions  
~> Laplace transforms with domains of convergence, etc.

## ODEs with rational symbols

Let  $G \in \mathbb{R}(\xi)^{\bullet \times w}$ , and consider the ‘differential equation’

$$G\left(\frac{d}{dt}\right)w = 0$$

$G$  is called the associated **symbol**

*What do we mean by its solutions*, i.e. by the behavior?

## ODEs with rational symbols

Let  $G \in \mathbb{R}(\xi)^{\bullet \times w}$ , and consider the ‘differential equation’

$$G \left( \frac{d}{dt} \right) w = 0 \quad G \text{ is called the associated } \mathbf{symbol}$$

*What do we mean by its solutions*, i.e. by the behavior?

Recall:

$\llbracket M \text{ left prime (over } \mathbb{R}[\xi]) \rrbracket$

$$:\Leftrightarrow \llbracket \llbracket M = FM' \rrbracket \Rightarrow \llbracket F \text{ unimodular} \rrbracket \rrbracket$$

$\Leftrightarrow \exists H \text{ such that } MH = I.$

In the scalar case,  $M = [m_1 \ m_2 \ \cdots \ m_n]$ , this means:

$m_1, m_2, \cdots, m_n$  have no common root.

## ODEs with rational symbols

Let  $G \in \mathbb{R}(\xi)^{\bullet \times w}$ , and consider the ‘differential equation’

$$G \left( \frac{d}{dt} \right) w = 0 \quad G \text{ is called the associated } \mathbf{\text{symbol}}$$

*What do we mean by its solutions*, i.e. by the behavior?

Let  $(P, Q)$  be a **left coprime** polynomial factorization of  $G$   
i.e.,  $P, Q \in \mathbb{R}[\xi]^{\bullet \times \bullet}$ ,  $\det(P) \neq 0$ ,  $G = P^{-1}Q$ ,  $[P : Q]$  left prime.

In scalar case, this means  $P$  and  $Q$  have no common roots.

## ODEs with rational symbols

Let  $G \in \mathbb{R}(\xi)^{\bullet \times w}$ , and consider the ‘differential equation’

$$G\left(\frac{d}{dt}\right)w = 0 \quad G \text{ is called the associated } \mathbf{\text{symbol}}$$

*What do we mean by its solutions*, i.e. by the behavior?

Let  $(P, Q)$  be a **left coprime** polynomial factorization of  $G$

$$\llbracket G\left(\frac{d}{dt}\right)w = 0 \rrbracket \Leftrightarrow \llbracket P^{-1}Q\left(\frac{d}{dt}\right)w = 0 \rrbracket \Leftrightarrow \llbracket Q\left(\frac{d}{dt}\right)w = 0 \rrbracket$$

**By definition** therefore, the behavior of  $G\left(\frac{d}{dt}\right)w = 0$  is equal to the behavior of  $Q\left(\frac{d}{dt}\right)w = 0$ .

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### Justification:

1.  $G$  proper.  $G(\xi) = C(I\xi - A)^{-1}B + D$  controllable realization. Consider the output nulling inputs:

$$\frac{d}{dt}x = Ax + Bw, \quad 0 = Cx + Dw$$

This set of  $w$ 's are exactly those that satisfy  $G\left(\frac{d}{dt}\right)w = 0$ .

Analogous for  $\frac{d}{dt}x = Ax + Bw, 0 = Cx + D\left(\frac{d}{dt}\right)w, D \in \mathbb{R}[\xi]^{\bullet \times \bullet}$ .

## ODEs with rational symbols

$$\llbracket G\left(\frac{d}{dt}\right)w = 0 \rrbracket \Leftrightarrow \llbracket P^{-1}Q\left(\frac{d}{dt}\right)w = 0 \rrbracket :\Leftrightarrow \llbracket Q\left(\frac{d}{dt}\right)w = 0 \rrbracket$$

**By definition** therefore, the behavior of  $G\left(\frac{d}{dt}\right)w = 0$  is equal to the behavior of  $Q\left(\frac{d}{dt}\right)w = 0$ .

### Justification:

**2. Consider  $y = G(s)w$ . View  $G(s)$  as a transfer f'n. Take your favorite definition of input/output pairs.**

**Output nulling inputs exactly those that satisfy  $G\left(\frac{d}{dt}\right)w = 0$ .**

**3. ...**

## ODEs with rational symbols

$$\left[ \left[ G\left(\frac{d}{dt}\right)w = 0 \right] \Leftrightarrow \left[ P^{-1}Q\left(\frac{d}{dt}\right)w = 0 \right] \right] \Leftrightarrow \left[ \left[ Q\left(\frac{d}{dt}\right)w = 0 \right] \right]$$

**By definition** therefore, the behavior of  $G\left(\frac{d}{dt}\right)w = 0$  is equal to the behavior of  $Q\left(\frac{d}{dt}\right)w = 0$ .

**Note!** With this def., we can deal with transfer functions,

$$y = F\left(\frac{d}{dt}\right)u, \quad \text{i.e.} \quad \begin{bmatrix} F\left(\frac{d}{dt}\right) & \vdots & -I \end{bmatrix} \begin{bmatrix} u \\ y \end{bmatrix} = 0$$

with  $F$  a matrix of rational functions, and completely avoid Laplace transforms, domains of convergence, and such cumbersome, but largely irrelevant, mathematical traps.





# Caveats

$F\left(\frac{d}{dt}\right)$  is not a map!

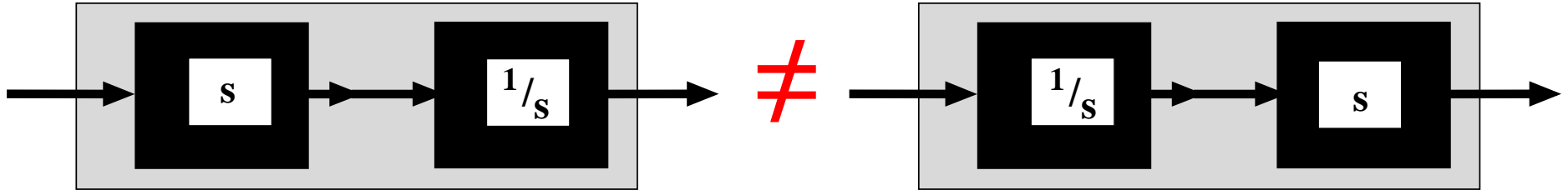
Consider

$$y = F\left(\frac{d}{dt}\right)u$$

We now know what it means that  $(u, y) \in \mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^\bullet)$  satisfies this ‘ODE’.

Given  $u$ ,  $\exists$  solution  $y$ , but not unique, unless  $F$  is polynomial

$G_1\left(\frac{d}{dt}\right)$  and  $G_2\left(\frac{d}{dt}\right)$  do not commute



$$G_1(s) = \frac{1}{s} \quad \text{and} \quad G_2(s) = s$$

**do not commute.**

$$y = \frac{1}{\frac{d}{dt}}v, \quad v = \frac{d}{dt}u \quad \Rightarrow \quad y(t) = u(t) + \text{constant}$$

$$y = \frac{d}{dt}v, \quad v = \frac{1}{\frac{d}{dt}}u \quad \Rightarrow \quad y(t) = u(t)$$

# Representations

## Stable representations

**Linear time-invariant differential systems**  $\Sigma = (\mathbb{R}, \mathbb{R}^w, \mathcal{B})$ .  
 $\mathcal{B} = \text{kernel} \left( R \left( \frac{d}{dt} \right) \right)$  for some  $R \in \mathbb{R}[\xi]^{\bullet \times w}$  **by definition**.

**But we may as well take the representation**  $G \left( \frac{d}{dt} \right) w = 0$  for  
some  $G \in \mathbb{R}(\xi)^{\bullet \times w}$  **as the def. of a LTIDS behavior.**

## Stable representations

**Linear time-invariant differential systems**  $\Sigma = (\mathbb{R}, \mathbb{R}^w, \mathcal{B})$ .  
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**But we may as well take the representation**  $G \left( \frac{d}{dt} \right) w = 0$  for some  $G \in \mathbb{R} (\xi)^{\bullet \times w}$  as the def. of a LTIDS behavior.

**$R$ : all poles at  $\infty$ , we can take  $G$  with no poles at  $\infty$ , or more generally with all poles in some non-empty set - symmetric w.r.t.  $\mathbb{R}$ . In particular (many variations on this theme):**

**Theorem:** Every linear time-invariant differential systems has a representation

$$G \left( \frac{d}{dt} \right) w = 0$$

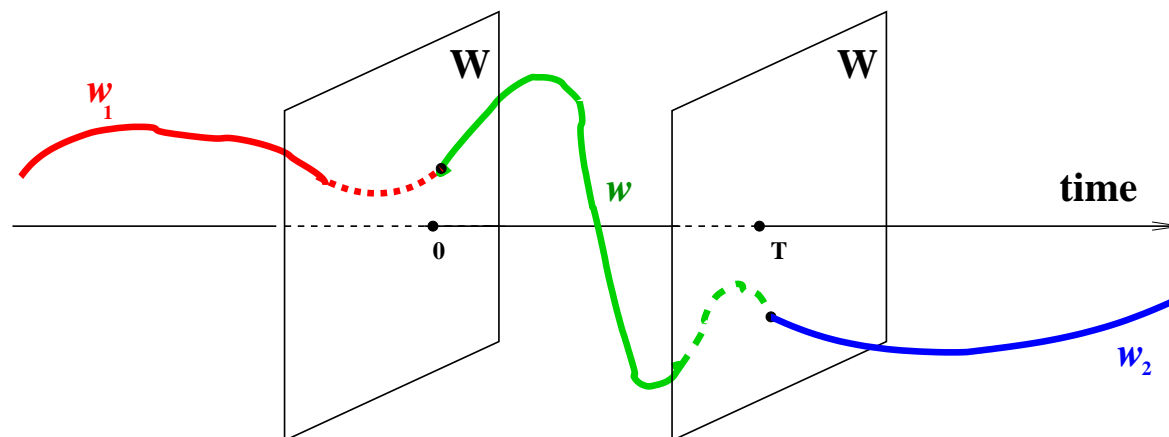
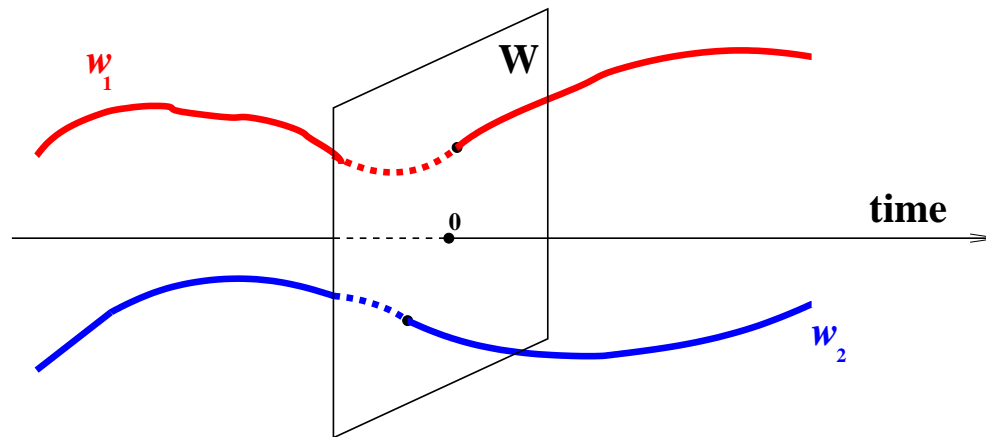
with  $G \in \mathbb{R} (\xi)^{\bullet \times w}$  **strictly proper stable rational**.

**Proof:** Take  $G(s) = \frac{R(s)}{(s+\lambda)^n}$ , suitable  $\lambda \in \mathbb{R}, n \in \mathbb{N}$ .

# Controllability and stabilizability

$\mathcal{B}$  is said to be **controllable**  $:\Leftrightarrow$

$\forall w_1, w_2 \in \mathcal{B}, \exists T \geq 0$  and  $w \in \mathcal{B}$  such that ...

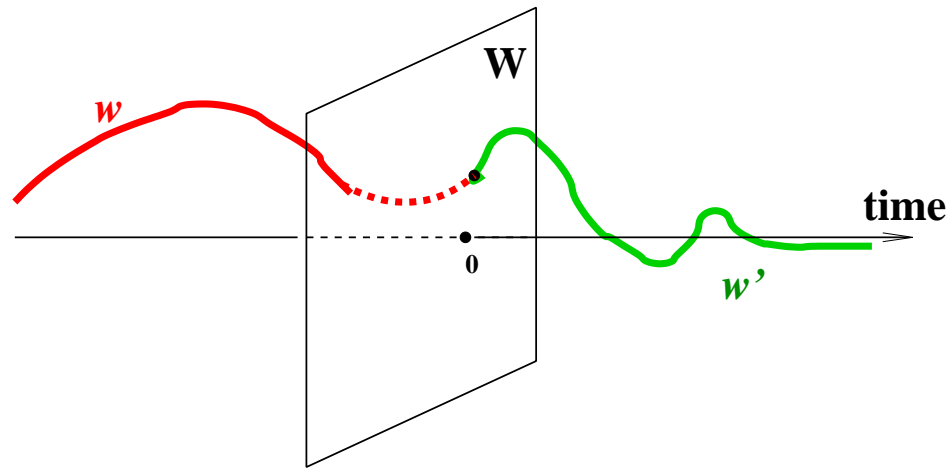


# Controllability and stabilizability

$\mathcal{B}$  is said to be **controllable**  $:\Leftrightarrow$

$\mathcal{B}$  is said to be **stabilizable**  $:\Leftrightarrow$

$$\forall w \in \mathcal{B}, \exists w' \in \mathcal{B} \text{ such that ...}$$



(asymptotic) stability in the sense of **Lyapunov**





## Rational representations

**What properties on  $G$  imply that the system with rational representation**

$$G \left( \frac{d}{dt} \right) w = 0 \quad G \in \mathbb{R}(\xi)^{\bullet \times w}$$

**has any of these properties?**

**Under what conditions on  $G$  does  $G \left( \frac{d}{dt} \right) w = 0$  define a controllable or a stabilizable system?**

## Rational representations

**What properties on  $G$  imply that the system with rational representation**

$$G \left( \frac{d}{dt} \right) w = 0$$

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**has any of these properties?**

**Under what conditions on  $G$  does  $G \left( \frac{d}{dt} \right) w = 0$  define a controllable or a stabilizable system?**

**Can a rational representation be used to put one of these properties in evidence?**

**Theorem: The LTIDS**

$$G\left(\frac{d}{dt}\right)w = 0 \quad G \in \mathbb{R}(\xi)^{\bullet \times w}$$

is **controllable** if and only if

$$G(\lambda) \text{ has the same rank } \forall \lambda \in \mathbb{C}$$

**Interpret carefully in cases like**

$$G(s) = \begin{bmatrix} s & 0 \\ 0 & \frac{1}{s} \end{bmatrix}, G(s) = \begin{bmatrix} s \\ 1 \\ \frac{1}{s} \end{bmatrix}, G(s) = \begin{bmatrix} s & 1 \\ & s \end{bmatrix}$$

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is **stabilizable** if and only if

$$G(\lambda) \text{ has the same rank } \forall \lambda \in \mathbb{C} \text{ with } \text{Realpart}(\lambda) \geq 0$$

## Rational image representations

**Theorem:** A LTIDS is **controllable** if and only if its behavior allows an image representation

$$w = M\left(\frac{d}{dt}\right)\ell \quad M \in \mathbb{R}(\xi)^{w \times \bullet}$$

For example,

$$y = F\left(\frac{d}{dt}\right)u \quad \rightsquigarrow w = \begin{bmatrix} u \\ y \end{bmatrix} = \begin{bmatrix} \ell \\ F\left(\frac{d}{dt}\right)\ell \end{bmatrix}$$

**Systems defined by transfer functions are controllable**

**Transfer functions can only deal with controllable systems**

## Rational image representations

**Theorem:** A LTIDS is **controllable** if and only if its behavior allows an image representation

$$w = M\left(\frac{d}{dt}\right)\ell \quad M \in \mathbb{R}(\xi)^{w \times \bullet}$$

**Theorem:** A LTIDS is **stabilizable** if and only if its behavior allows a kernel representation

$$R\left(\frac{d}{dt}\right)w = 0$$

with  $R \in \mathbb{R}(\xi)^{\bullet \times w}$  left prime  
over the ring of (proper) stable rationals

## Raison d'être of rational representations

**LTIDSs** are **defined** in terms of **polynomial** symbols

$$R \left( \frac{d}{dt} \right) w = 0 \quad R \in \mathbb{R} [\xi]^{\bullet \times w}$$

(behavior  $\mathcal{B} :=$  the  $\mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w)$  solutions) but can also be represented by **rational** symbols

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$$G \left( \frac{d}{dt} \right) w = 0 \quad G \in \mathbb{R}(\xi)^{\bullet \times w}$$

**Behavior** := the set of solutions of

$$Q \left( \frac{d}{dt} \right) w = 0 \quad Q \in \mathbb{R} [\xi]^{\bullet \times w}$$

where  $G = P^{-1}Q$ ,  $P, Q \in \mathbb{R} [\xi]^{\bullet \times \bullet}$ ,  $P$  and  $Q$  left coprime



## Raison d'être of rational representations

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$$G \left( \frac{d}{dt} \right) w = 0 \quad G \in \mathbb{R}(\xi)^{\bullet \times w}$$

This added flexibility  $\rightsquigarrow$  better adapted to certain applications

**e.g. (series, parallel, ...) interconnections**

**e.g. distance between systems**

**e.g. behavioral model reduction**

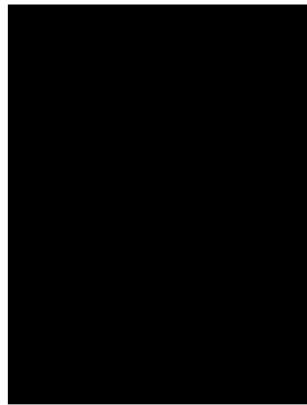
**e.g. parametrization of the set of stabilizing controllers**

# Parametrization of stabilizing controllers

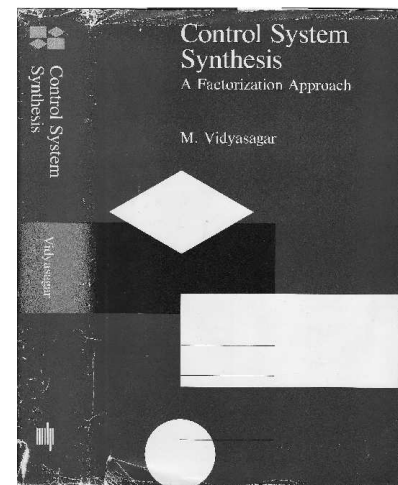
One of the main applications where rational representations are used is for the **Kučera-Youla parametrization of stabilizing controllers** cfr. the book by Vidyasagar



**Vladimir Kučera**



**Dante Youla**



**M. Vidyasagar**

# **Norm-preserving representations**

## Norm-preserving representations

Let  $\mathcal{B}$  be the behavior of a controllable LTIDS.

Then it allows a rational symbol based image representation

$$w = M\left(\frac{d}{dt}\right)\ell \quad \text{with } M \in \mathbb{R}(\xi)^{w \times \bullet} \quad \& \quad M(-\xi)^\top M(\xi) = I$$

**i.e.,**  $\|\ell\|_{\mathcal{L}_2(\mathbb{R}, \mathbb{R}^\bullet)}^2 = \|w\|_{\mathcal{L}_2(\mathbb{R}, \mathbb{R}^w)}^2$  ‘**norm preserving image repr.**’

$$\begin{aligned} \int_{-\infty}^{+\infty} \|w(t)\|^2 dt &= \frac{1}{2\pi} \int_{-\infty}^{+\infty} \|\hat{w}(i\omega)\|^2 d\omega = \\ \frac{1}{2\pi} \int_{-\infty}^{+\infty} \|M(i\omega)\hat{\ell}(i\omega)\|^2 d\omega &= \frac{1}{2\pi} \int_{-\infty}^{+\infty} \|\hat{\ell}(i\omega)\|^2 d\omega = \int_{-\infty}^{+\infty} \|\ell(t)\|^2 dt \end{aligned}$$

**Note:**  $M$  cannot be polynomial, **it must be rational**

**Obviously  $M$  must be proper. Can also make it stable.**

## Norm-preserving representations

Let  $\mathcal{B}$  be the behavior of a controllable LTIDS.

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$$w = M\left(\frac{d}{dt}\right)\ell \quad \text{with } M \in \mathbb{R}(\xi)^{w \times \bullet} \quad \& \quad M(-\xi)^\top M(\xi) = I$$

Idea of proof: Start with obs. polynomial im. representation

$$w = M\left(\frac{d}{dt}\right)\ell.$$

Factor  $M^\top(-\xi)M(\xi) = F^\top(-\xi)F(\xi)$

Now take rational symbol based image representation

$$w = MF^{-1}\left(\frac{d}{dt}\right)\ell$$

# Distance between systems

## Motivation

**What is a good, computable, definition for the distance between two (LTID) systems?**

**Basic issue underlying model simplification, robustness, etc.**

## Motivation

What is a good, computable, definition for the **distance** between two (LTID) systems?

Basic issue underlying model simplification, robustness, etc.

- Approximate a system by a simpler one.
- If a system has a particular property (e.g., stabilized by a controller), will this also hold for close-by systems?
- Does a sequence of systems converge?

What is meant

by ‘approximate’, by ‘close-by’, by ‘converge’?



# The gap

## Distance between linear subspaces

**In the behavioral theory, we identify a dynamical system with its behavior, that is, a set of trajectories. For LTIDSs, with a subspace  $\mathcal{B} \subseteq \mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w)$ .**

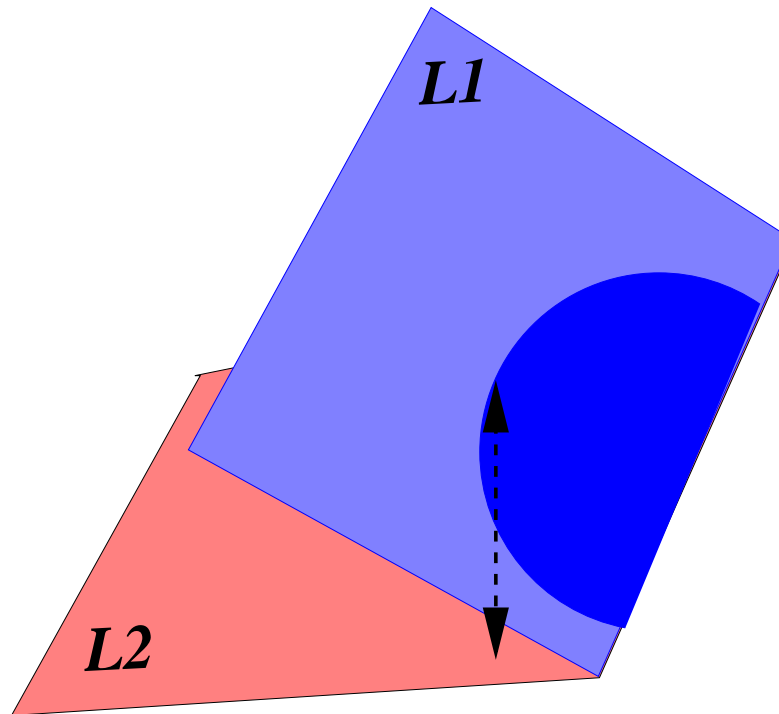
**Distance between systems**

**$\cong$  distance between linear subspaces.**

# Distance between linear subspaces of $\mathbb{R}^n$

$\mathcal{L}_1, \mathcal{L}_2 \subseteq \mathbb{R}^n$ , **linear subspaces**

$$\vec{d}(\mathcal{L}_1, \mathcal{L}_2) := \max_{x_1 \in \mathcal{L}_1, \|x_1\|=1} \min_{x_2 \in \mathcal{L}_2} \|x_1 - x_2\|$$



## Distance between linear subspaces of $\mathbb{R}^n$

$\mathcal{L}_1, \mathcal{L}_2 \subseteq \mathbb{R}^n$ , **linear subspaces**

$$d(\mathcal{L}_1, \mathcal{L}_2) := \max \left( \left\{ \vec{d}(\mathcal{L}_1, \mathcal{L}_2), \vec{d}(\mathcal{L}_2, \mathcal{L}_1) \right\} \right)$$

$$0 \leq d(\mathcal{L}_1, \mathcal{L}_2) \leq 1$$

## Distance between linear subspaces of $\mathbb{R}^n$

$\mathcal{L}_1, \mathcal{L}_2 \subseteq \mathbb{R}^n$ , linear subspaces

$P_{\mathcal{L}} \perp$  projection onto  $\mathcal{L}$

$S_1, S_2$  matrices, columns orthonormal basis for  $\mathcal{L}_1, \mathcal{L}_2$

**Note:**  $S_1 S_1^\top, S_2 S_2^\top$  orthogonal projectors

$$\begin{aligned} d(\mathcal{L}_1, \mathcal{L}_2) &= \|P_{\mathcal{L}_1} - P_{\mathcal{L}_2}\| && \text{‘gap’, ‘aperture’} \\ &= \|S_1 S_1^\top - S_2 S_2^\top\| \\ &= \min_{\text{matrices } U} \|S_1 - S_2 U\| \\ &= \min_{U \text{ such that } U\mathcal{L}_1 = \mathcal{L}_2} \|I - U\| \end{aligned}$$

## Distance between linear subspaces of $\mathbb{R}^n$

$\mathcal{L}_1, \mathcal{L}_2 \subseteq \mathbb{R}^n$ , linear subspaces

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**Therefore,**  $d(\mathcal{L}_1, \mathcal{L}_2) = \|S_1 S_1^\top - S_2 S_2^\top\| \leq \|S_1 - S_2\|$

# Distance between LTIDSs

## Distance between controllable behaviors

$\min \rightarrow \mathbf{inf}$ ,  $\max \rightarrow \mathbf{sup}$ , etc., readily generalized to linear subspaces of Hilbert space, ..... and to LTIDSs.

**Which subspace of which Hilbert space should we associate with a LTIDS with behavior  $\mathcal{B} \subseteq \mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w)$ ?**



## Distance between controllable behaviors

$\min \rightarrow \mathbf{inf}$ ,  $\max \rightarrow \mathbf{sup}$ , etc., readily generalized to linear subspaces of Hilbert space, ..... and to LTIDSs.

Which subspace of which Hilbert space should we associate with a LTIDS with behavior  $\mathcal{B} \subseteq \mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w)$ ?

For LTIDS, behaviors  $\mathcal{B} \mapsto (\mathcal{B} \cap \mathcal{L}_2(\mathbb{R}, \mathbb{R}^w))^{\text{closure}}$

Defines a  $1 \leftrightarrow 1$  relation between controllable systems and ‘certain’ closed subspaces of  $\mathcal{L}_2(\mathbb{R}, \mathbb{R}^w)$ .

## Distance between controllable behaviors

Define the distance between two controllable behaviors as

$$d(\mathcal{B}_1, \mathcal{B}_2) := \text{gap}((\mathcal{B}_1 \cap \mathcal{L}_2(\mathbb{R}, \mathbb{R}^w))^{\text{closure}}, (\mathcal{B}_2 \cap \mathcal{L}_2(\mathbb{R}, \mathbb{R}^w))^{\text{closure}})$$

We consider only the  $\mathcal{L}_2$ -trajectories for measuring distance.

Henceforth, keep notation  $\mathcal{B}$  for  $(\mathcal{B} \cap \mathcal{L}_2(\mathbb{R}, \mathbb{R}^w))^{\text{closure}}$

## Distance between controllable behaviors

Define the distance between two controllable behaviors as

$$d(\mathcal{B}_1, \mathcal{B}_2) := \text{gap}((\mathcal{B}_1 \cap \mathcal{L}_2(\mathbb{R}, \mathbb{R}^w))^{\text{closure}}, (\mathcal{B}_2 \cap \mathcal{L}_2(\mathbb{R}, \mathbb{R}^w))^{\text{closure}})$$

We consider only the  $\mathcal{L}_2$ -trajectories for measuring distance.

$$\forall w_1 \in \mathcal{B}_1, \exists w_2 \in \mathcal{B}_2 \text{ such that } \|w_1 - w_2\| \leq \text{gap}(\mathcal{B}_1, \mathcal{B}_2) \|w_1\|$$

and vice-versa. Small gap  $\Rightarrow$  the models are ‘close’.

- How to compute the gap?
- Model reduce according to the gap!

## The gap and norm-preserving representations

Let  $\mathcal{B}$  be the behavior of a controllable LTIDS.

Then it allows a rational symbol based image representation

$$w = M\left(\frac{d}{dt}\right)\ell \quad \text{with } M \in \mathbb{R}(\xi)^{w \times \bullet} \quad \& \quad M(-\xi)^\top M(\xi) = I$$

**i.e.,**  $\|\ell\|_{\mathcal{L}_2(\mathbb{R}, \mathbb{R}^\bullet)}^2 = \|w\|_{\mathcal{L}_2(\mathbb{R}, \mathbb{R}^w)}^2$  ‘**norm preserving image repr.**’

$\mathcal{B}_1 \mapsto M_1, \mathcal{B}_2 \mapsto M_2$ , both norm preserving & stable, then

$$\begin{aligned} \text{gap}(\mathcal{B}_1, \mathcal{B}_2) &= \|M_1(i\omega)M_1(-i\omega)^\top - M_2(i\omega)M_2(-i\omega)^\top\|_{\mathcal{L}_\infty} \\ &\leq \|M_1(i\omega) - M_2(i\omega)\|_{\mathcal{H}_\infty} \end{aligned}$$

# Model reduction

## Reducing the state dimension

There is an elegant theory for reducing the state space dimension of **stable** LTI **input/output** systems.

Let  $\mathcal{B}$  be described by  $\frac{d}{dt}\mathbf{x} = A\mathbf{x} + B\mathbf{u}$ ,  $\mathbf{y} = C\mathbf{x} + D\mathbf{u}$  with  $A$  Hurwitz ( $:\Leftrightarrow$  eigenvalues in left half plane).

There are effective methods (balancing, AAK) with good error bounds (in terms of the  $\mathcal{H}_\infty$  norm) for approximating  $\mathcal{B}$  by a (stable) system with a lower dimensional state space.

## Reducing the state dimension

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$$\mathbf{w} \cong \begin{bmatrix} \mathbf{u} \\ \mathbf{y} \end{bmatrix}$$

Balanced model reduction  $\Rightarrow$

$$\|F(i\omega) - F_{\text{reduced}}(i\omega)\|_{\mathcal{L}_\infty} \leq 2 \left( \sum_{\text{neglected Hankel SVs}} \sigma_k \right)$$



Keith Glover

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$F(s)$  proper **stable** rational  $\Rightarrow$  reducible.

∴ Extend this to situations where we do not make a distinction between inputs and outputs, and to unstable systems.



# Model reduction by balancing

Start with  $\mathcal{B}$ . Take representation

$$w = M\left(\frac{d}{dt}\right)\ell \quad \text{with } M \in \mathbb{R}(\xi)^{w \times \bullet} \quad \text{norm preserving, stable}$$

Now model reduce  $w = M\left(\frac{d}{dt}\right)\ell$  (viewed as a stable input/output system) using, for example, balancing

$$\rightsquigarrow w = M_{\text{reduced}}\left(\frac{d}{dt}\right)\ell$$

and an error bound

$$\|M - M_{\text{reduced}}\|_{\mathcal{H}_\infty} \leq 2 \left( \sum_{\text{neglected SVs of } M} \sigma_k \right)$$

## Behavioral error bound

Start with stable norm preserving representation of  $\mathcal{B}$

$$w = M\left(\frac{d}{dt}\right)\ell \quad \text{with } M \in \mathbb{R}(\xi)^{w \times \bullet}$$

Model reduce using balancing  $\rightsquigarrow w = M_{\text{reduced}}\left(\frac{d}{dt}\right)\ell$ .

Call behavior  $\mathcal{B}_{\text{reduced}}$ . Error bound

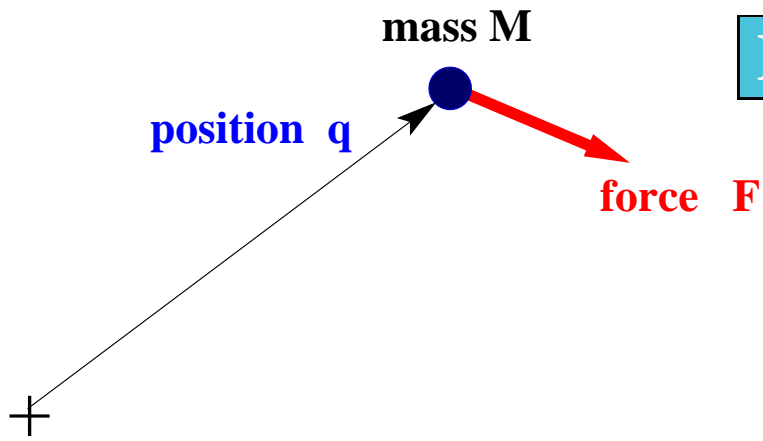
$$\begin{aligned} \text{gap}(\mathcal{B}, \mathcal{B}_{\text{reduced}}) &= \|MM^{\top} - M_{\text{reduced}}M_{\text{reduced}}^{\top}\|_{\mathcal{L}_{\infty}} \\ &\leq \|M - M_{\text{reduced}}\|_{\mathcal{H}_{\infty}} \\ &\leq 2 \left( \sum_{\text{neglected SVs of } M} \sigma_k \right) \end{aligned}$$

$\forall w \in \mathcal{B} \exists w' \in \mathcal{B}_{\text{red}}$  such that  $\|w - w'\| \leq 2 \left( \sum_{\text{neglected SVs}} \sigma_k \right) \|w\|$

and vice-versa.

$\sum_{\text{neglected SVs of } M} \sigma_k$  small  $\Rightarrow$  good approximation in the gap.

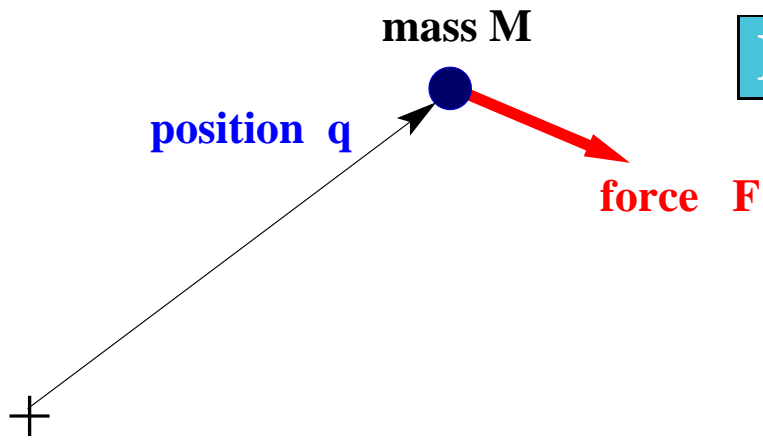
# Examples



**Pointmass**

$$F = M \frac{d^2}{dt^2} q,$$

$$w = \begin{bmatrix} F \\ q \end{bmatrix} \cong \begin{bmatrix} M \frac{d^2}{dt^2} \\ 1 \end{bmatrix} \ell$$



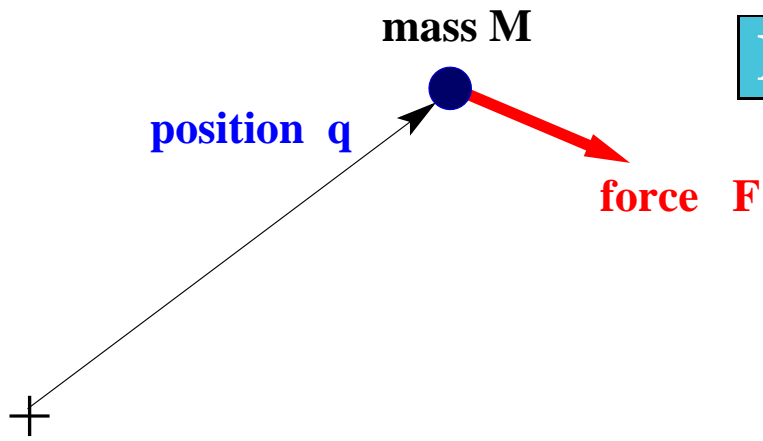
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**Norm preserving, stable**

$$\begin{bmatrix} F \\ q \end{bmatrix} = \begin{bmatrix} \frac{M \frac{d^2}{dt^2}}{M \frac{d^2}{dt^2} + \sqrt{2M} \frac{d}{dt} + 1} \\ \frac{1}{M \frac{d^2}{dt^2} + \sqrt{2M} \frac{d}{dt} + 1} \end{bmatrix} \ell$$



**Pointmass**

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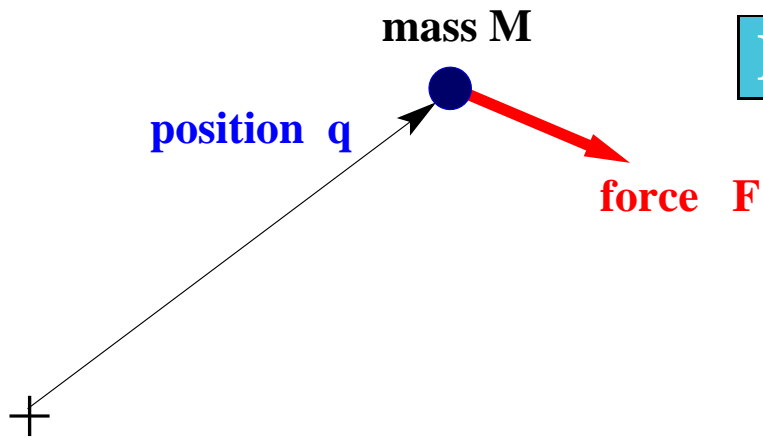
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**reduced model**

$$\begin{bmatrix} F \\ q \end{bmatrix} = \begin{bmatrix} \frac{\sqrt{M} \frac{d}{dt} - \frac{1}{2}}{\sqrt{M} \frac{d}{dt} + \frac{1}{\sqrt{2}}} \\ \frac{\frac{1}{2}}{\sqrt{M} \frac{d}{dt} + \frac{1}{\sqrt{2}}} \end{bmatrix} \ell$$



**Pointmass**

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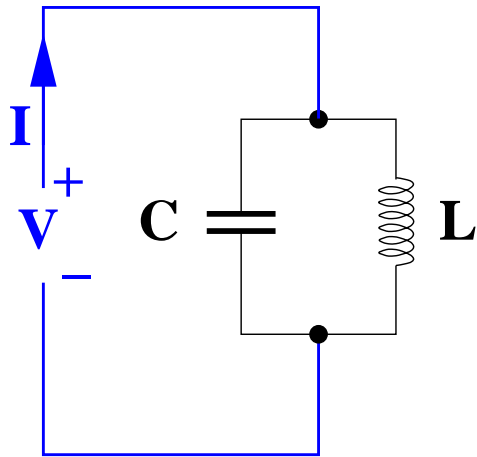
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$F = \frac{d^2}{dt^2} q$  **first order approximation**

$F = 2\sqrt{M} \frac{d}{dt} q - q$

# LC circuit



**kernel**

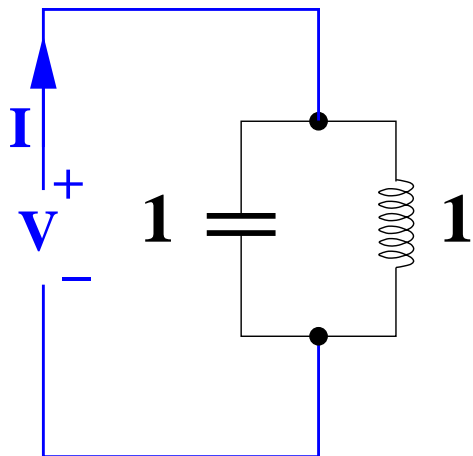
$$\left(1 + LC \frac{d^2}{dt^2}\right) V = C \frac{d}{dt} I$$

**image**

$$\begin{bmatrix} I \\ V \end{bmatrix} = \begin{bmatrix} 1 + LC \frac{d^2}{dt^2} \\ C \frac{d}{dt} \end{bmatrix} \ell$$



## LC circuit



kernel

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image

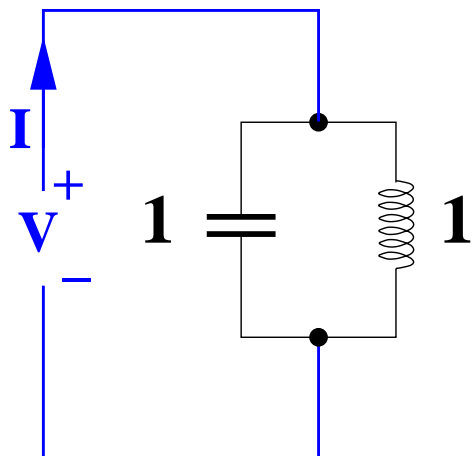
$$\begin{bmatrix} I \\ V \end{bmatrix} = \begin{bmatrix} 1 + LC \frac{d^2}{dt^2} \\ C \frac{d}{dt} \end{bmatrix} \ell$$

Take  $L = C = 1$ .

stable norm-preserving

$$\begin{bmatrix} I \\ V \end{bmatrix} = \frac{1}{\frac{d^2}{dt^2} + \frac{d}{dt} + 1} \begin{bmatrix} \frac{d^2}{dt^2} + 1 \\ \frac{d}{dt} \end{bmatrix} \ell$$

# LC circuit



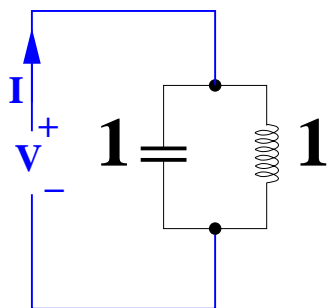
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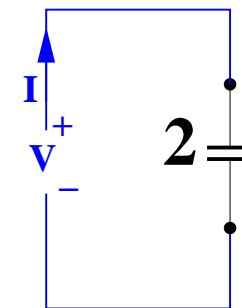
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reduced model order = 1

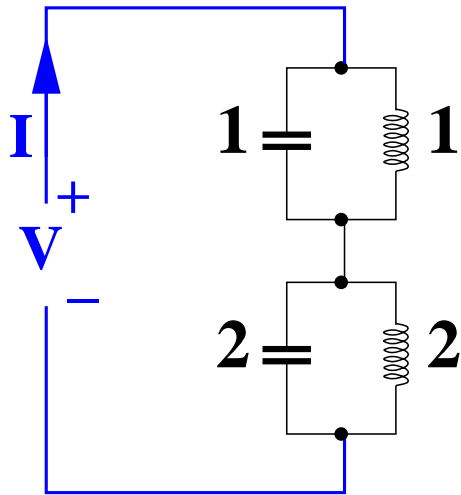
$$\begin{bmatrix} I \\ V \end{bmatrix} = \frac{1}{\frac{d}{dt} + \frac{1}{2}} \begin{bmatrix} \frac{d}{dt} \\ \frac{1}{2} \end{bmatrix} \ell$$



$$\left( \frac{d^2}{dt^2} + 1 \right) V = \frac{d}{dt} I \quad \rightsquigarrow \quad \frac{d}{dt} V = \frac{1}{2} I$$



# LCLC circuit



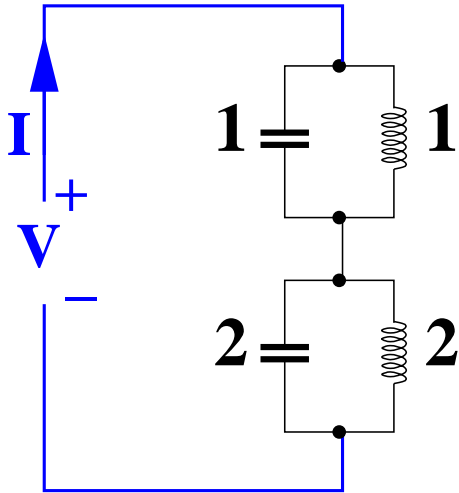
**kernel**

$$\left(1 + 5\frac{d^2}{dt^2} + 4\frac{d^4}{dt^4}\right) V = \left(3\frac{d}{dt} + 6\frac{d^3}{dt^3}\right) I$$

**image**

$$\begin{bmatrix} I \\ V \end{bmatrix} = \begin{bmatrix} 1 + 5\frac{d^2}{dt^2} + 4\frac{d^4}{dt^4} \\ 3\frac{d}{dt} + 6\frac{d^3}{dt^3} \end{bmatrix} \ell$$

# LCLC circuit



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**image** 
$$\begin{bmatrix} I \\ V \end{bmatrix} = \begin{bmatrix} 1 + 5\frac{d^2}{dt^2} + 4\frac{d^4}{dt^4} \\ 3\frac{d}{dt} + 6\frac{d^3}{dt^3} \end{bmatrix} \ell$$

**stable norm-preserving image**

$$\begin{bmatrix} I \\ V \end{bmatrix} = \frac{1}{1 + 3\frac{d}{dt} + 5\frac{d^2}{dt^2} + 6\frac{d^3}{dt^3} + 4\frac{d^4}{dt^4}} \begin{bmatrix} 1 + 5\frac{d^2}{dt^2} + 4\frac{d^4}{dt^4} \\ 3\frac{d}{dt} + 6\frac{d^3}{dt^3} \end{bmatrix} \ell$$

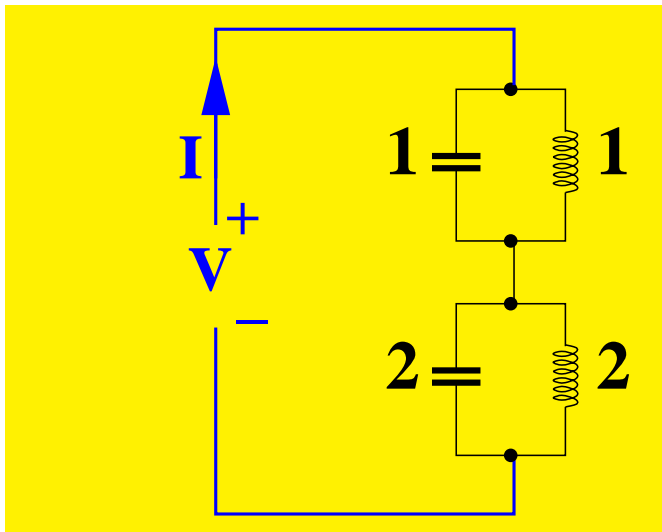
# LCLC circuit

**stable norm-preserving image**

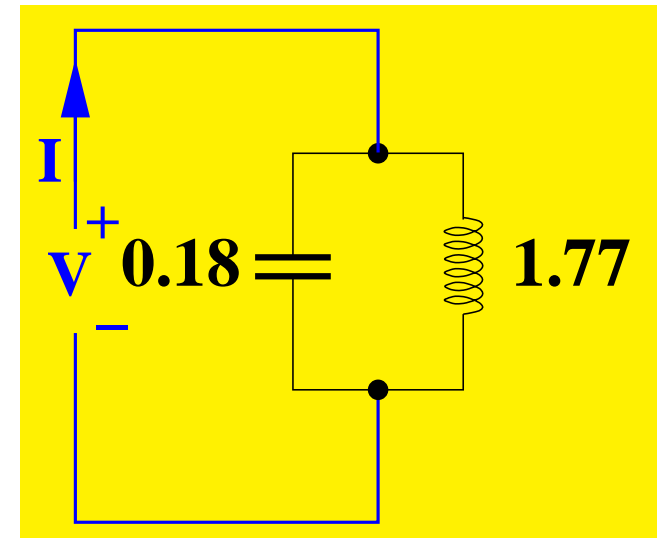
$$\begin{bmatrix} I \\ V \end{bmatrix} = \frac{1}{1 + 3\frac{d}{dt} + 5\frac{d^2}{dt^2} + 6\frac{d^3}{dt^3} + 4\frac{d^4}{dt^4}} \begin{bmatrix} 1 + 5\frac{d^2}{dt^2} + 4\frac{d^4}{dt^4} \\ 3\frac{d}{dt} + 6\frac{d^3}{dt^3} \end{bmatrix} \ell$$

**red. order = 2**

$$\begin{bmatrix} I \\ V \end{bmatrix} = \frac{1}{\frac{d^2}{dt^2} + 0.1861\frac{d}{dt} + 0.3298} \begin{bmatrix} \frac{d^2}{dt^2} + 0.3298 \\ 0.1861\frac{d}{dt} \end{bmatrix} \ell$$



~>



# Summary of Lecture 4a

## The main points

- ▶  $G\left(\frac{d}{dt}\right)w = 0$  defined in terms left-coprime factorization of rational  $G$ .
- ▶  $y = G\left(\frac{d}{dt}\right)u$  does not require Laplace transform.

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- ▶ Numerous other applications of rational symbols

**End of Lecture 4a**