



SYNTHESIS of RECIPROCAL CIRCUITS

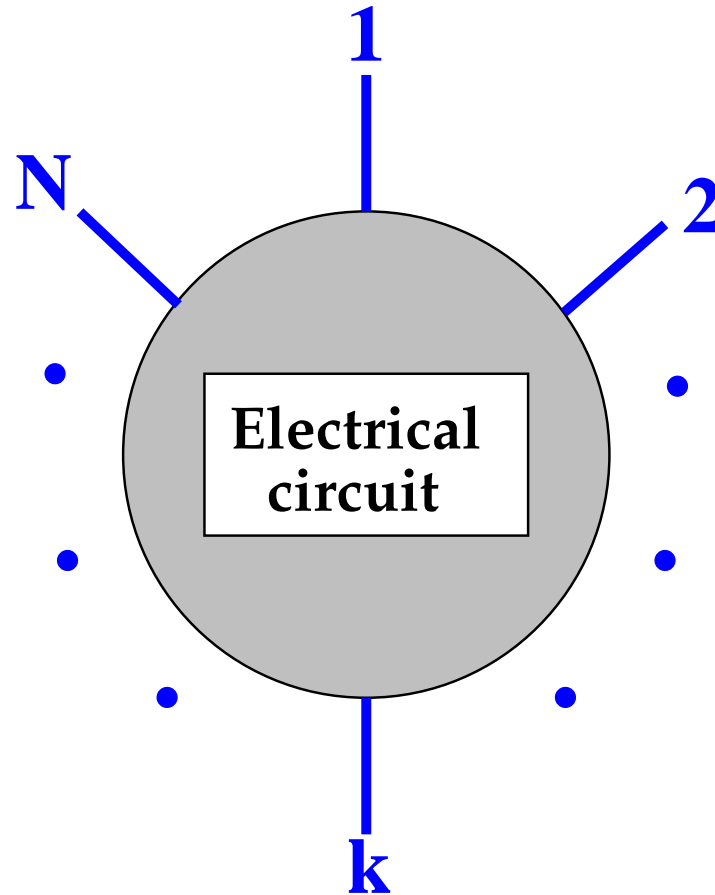
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Second Workshop on Mathematical Aspects of Network Synthesis

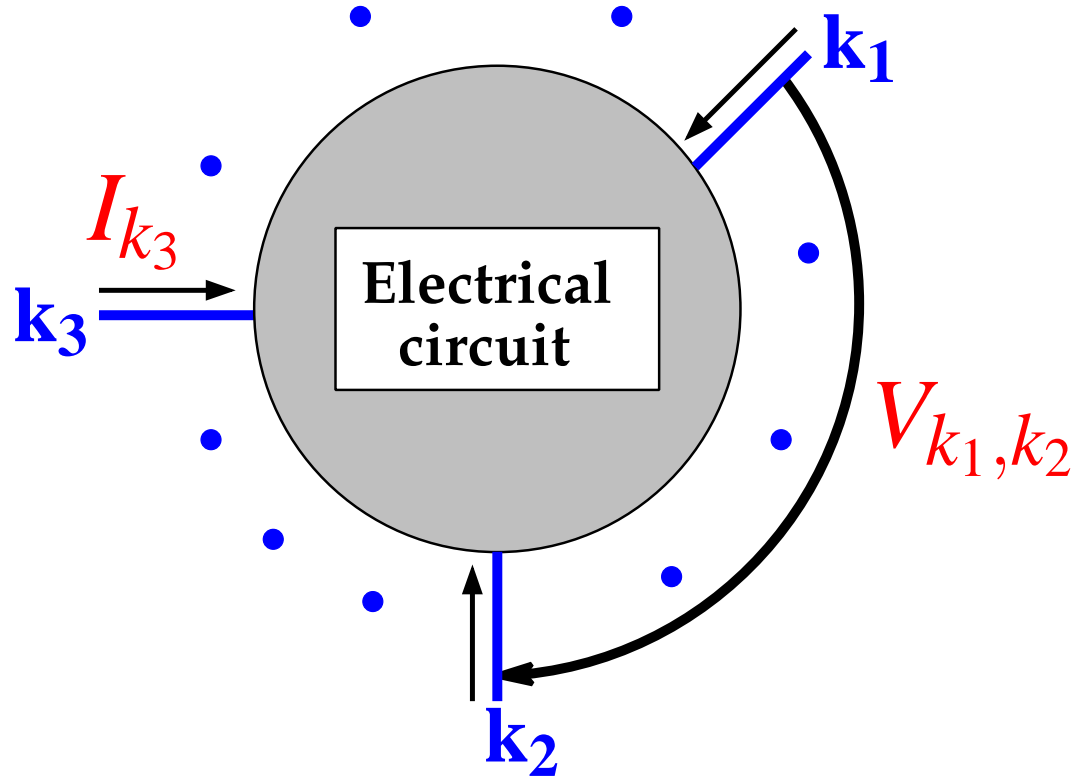
Circuits

Electrical circuits



An N -terminal circuit

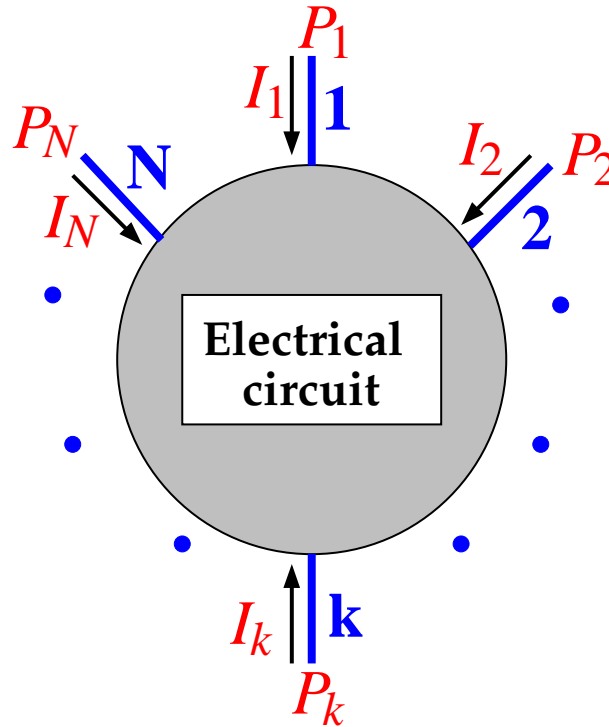
Electrical circuits



Variables: terminal currents and voltages

Electrical circuits

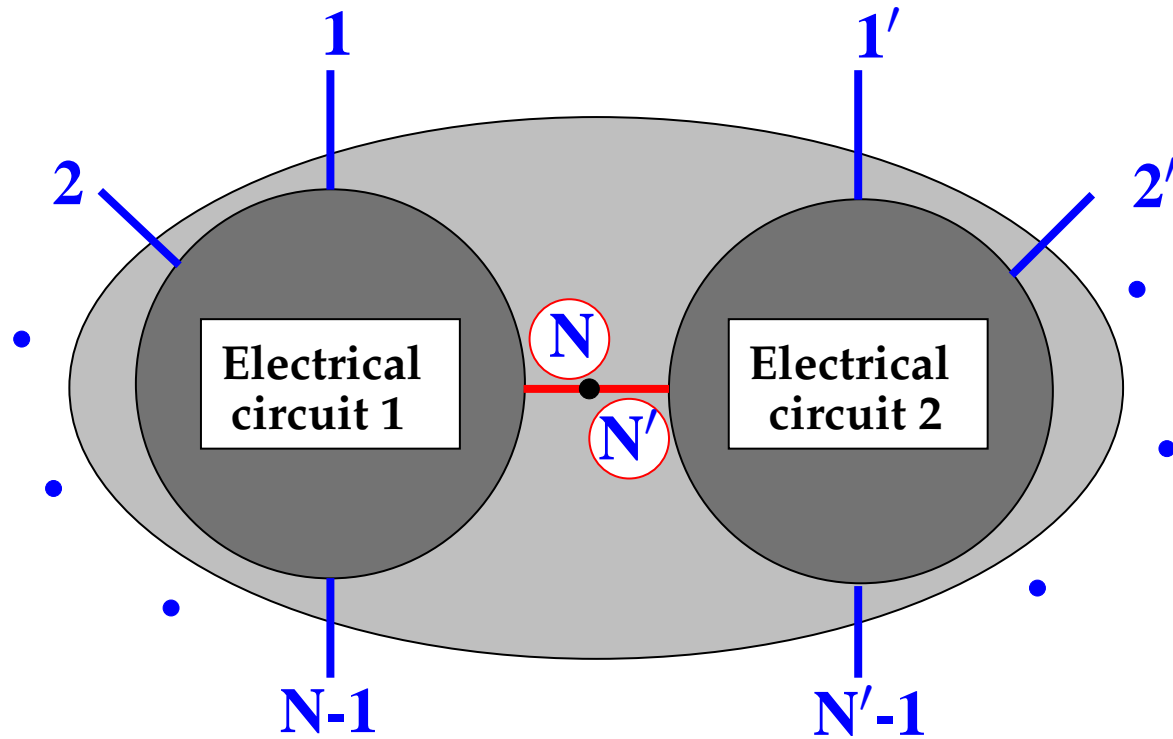
Assume KVL: $V_{k_1,k_2} + V_{k_2,k_3} + V_{k_3,k_1} = 0.$



Variables: terminal currents and potentials

$$V_{k_1,k_2} = P_{k_1} - P_{k_2}.$$

Terminal connection



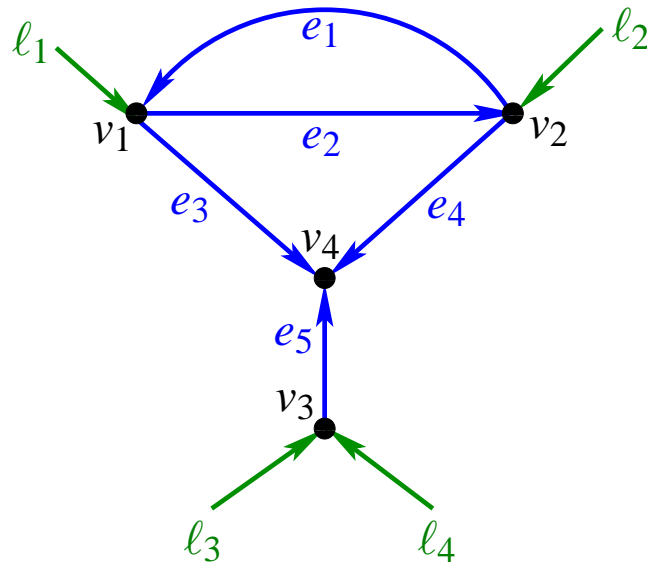
Imposes :

$$I_N + I_{N'} = 0, \quad P_N = P_{N'}.$$

Extendable to more than 2 terminals, than 2 circuits.

Circuit architecture

Digraph with leaves



vertices: subcircuits, ‘building blocks’,

edges: connections,

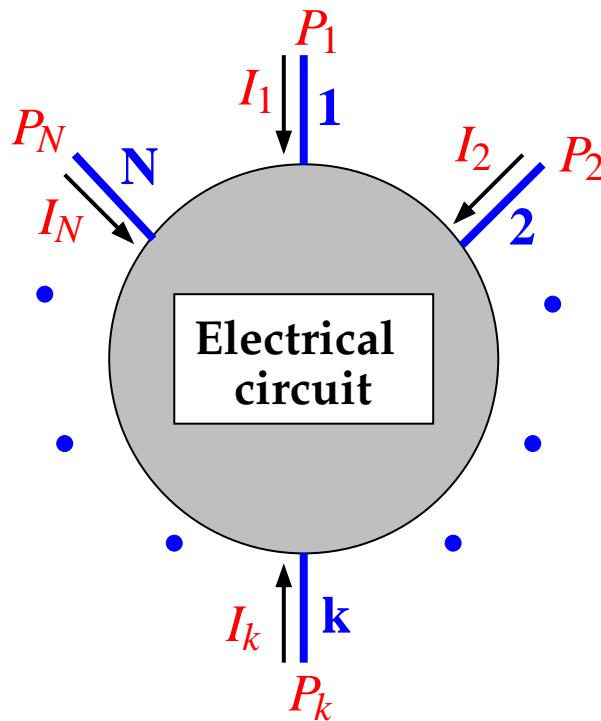
leaves: external terminals.

The terminal behavior

The current/potential behavior

$$\mathcal{B}_{IP} \subseteq (\mathbb{R}^N \times \mathbb{R}^N)^{\mathbb{R}}.$$

$\mathcal{B}_{IP} =$ **all possible** $(I_1, P_1, I_2, P_2, \dots, I_N, P_N) : \mathbb{R} \rightarrow \mathbb{R}^N \times \mathbb{R}^N$.



KCL: $I_1 + I_2 + \dots + I_N = 0.$

Analysis & synthesis

Analysis:

Given circuit architecture and element laws,
find \mathcal{B}_{IP} .

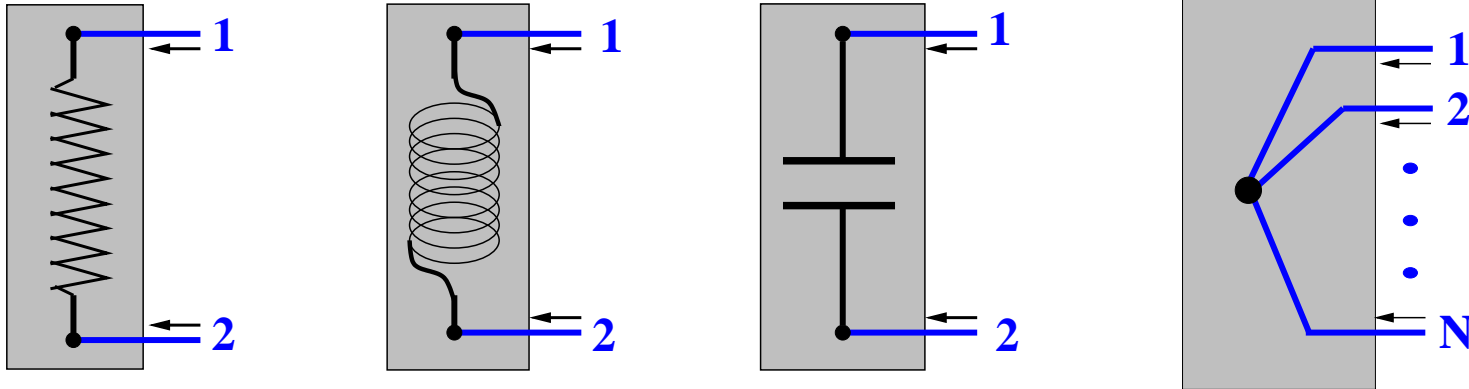
Synthesis:

Given \mathcal{B}_{IP} and building blocks,
find circuit architecture.

Which \mathcal{B}_{IP} 's can be synthesized?

Building blocks

Resistor, inductor, capacitor, connector

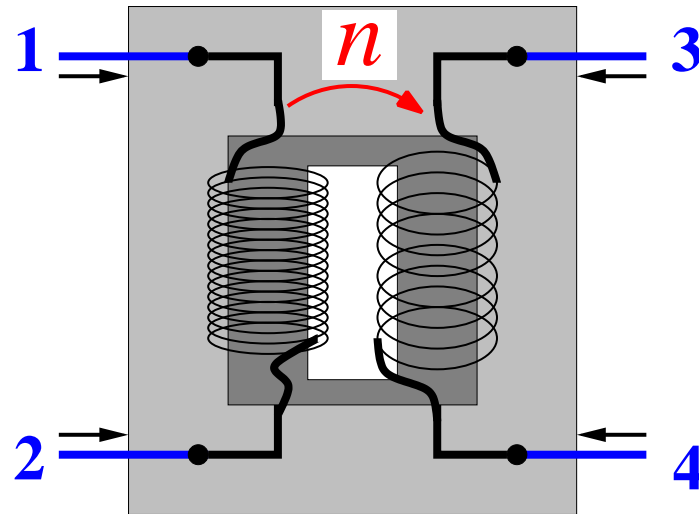


Element laws, e.g. inductor

$$I_1 + I_2 = 0, \quad P_1 - P_2 = L \frac{d}{dt} I_1.$$

L = inductance.

Transformer

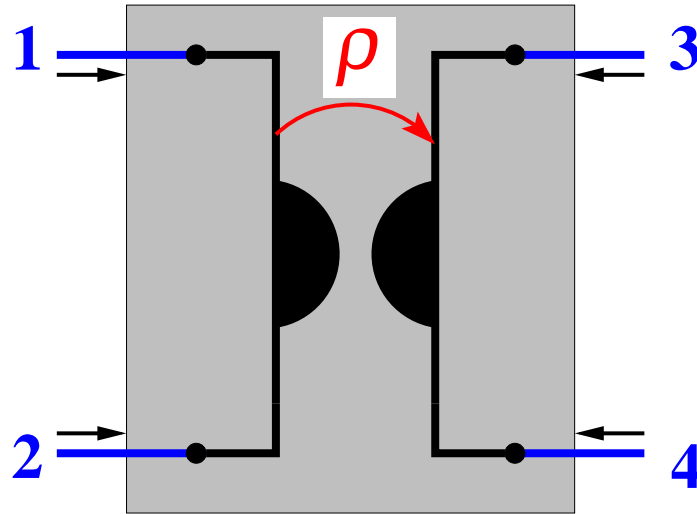


$$I_1 + I_2 = 0, I_3 + I_4 = 0,$$

$$I_3 = -n I_1, \quad (P_1 - P_2) = n (P_3 - P_4).$$

n = 'turns ratio'.

Gyrator



$$I_1 + I_2 = 0, I_3 + I_4 = 0,$$

$$(P_1 - P_2) = \rho I_3, \quad (P_3 - P_4) = -\rho I_1.$$

ρ = ‘gyrator resistance’.

Tellegen B.D.H. : ‘The gyrator, a new electric network element’,
Philips Research Reports, 3, pp. 81-101, 1948.

RLCTG Synthesis

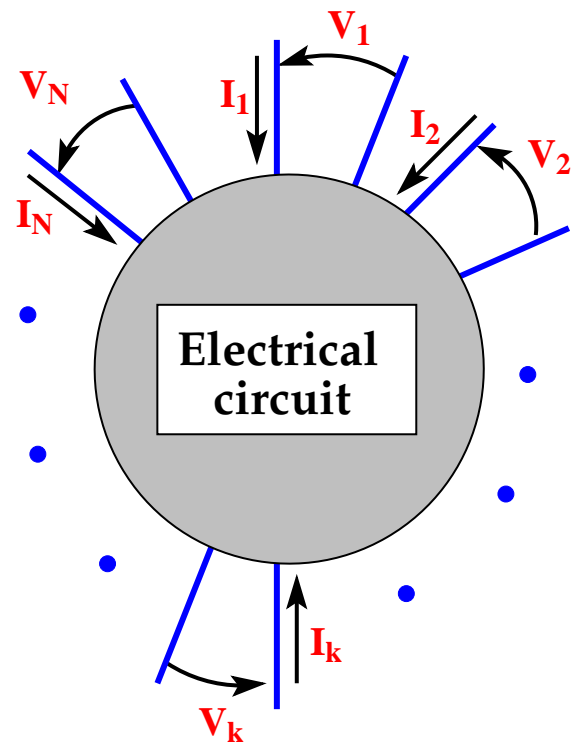
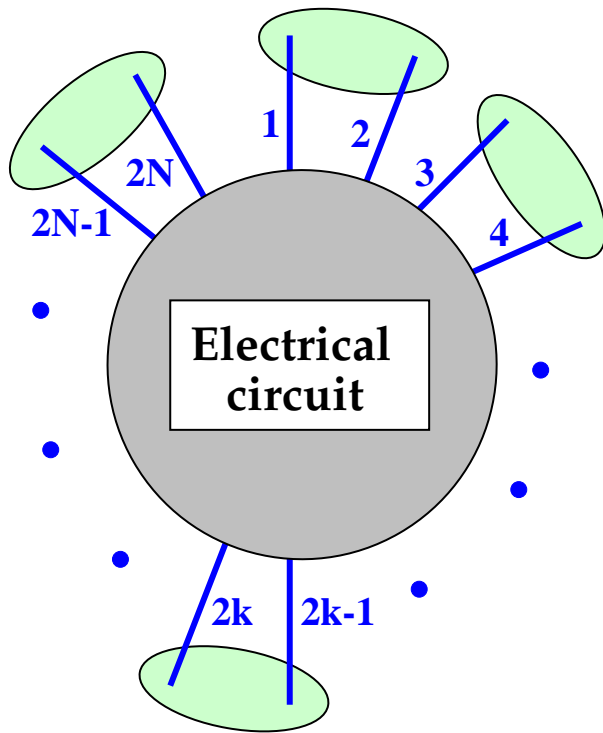
Terminal synthesis

Determine the terminal behaviors \mathcal{B}_{IP} that are achievable by interconnecting a finite number of

positive resistors,
positive capacitors,
positive inductors,
transformers,
and gyrators.

Ports

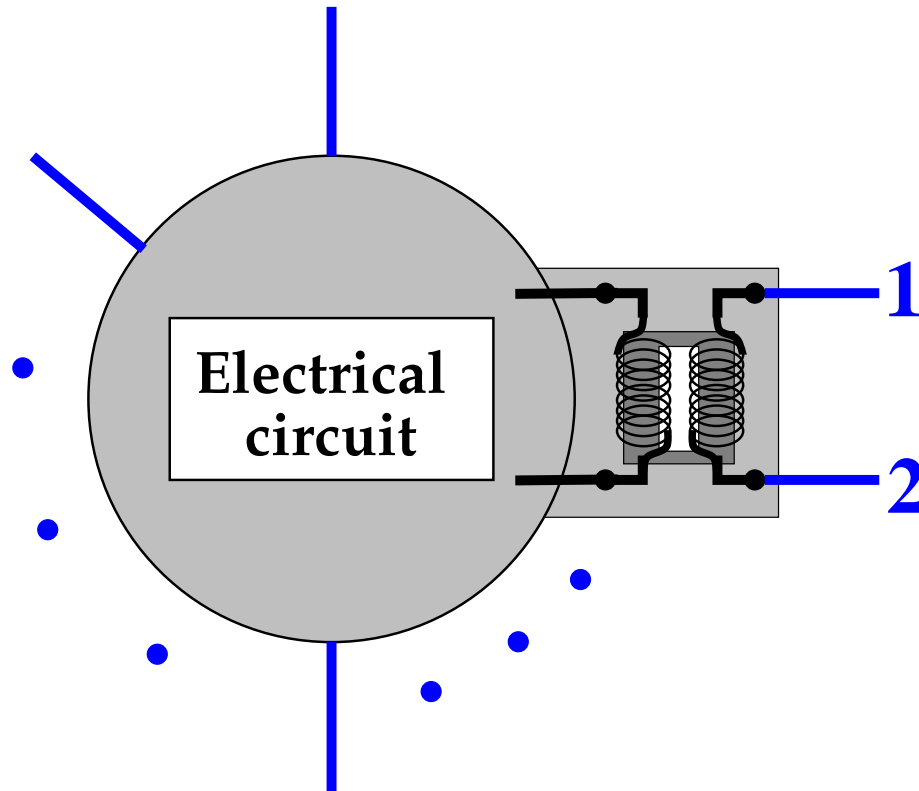
The port behavior



Port behavior:

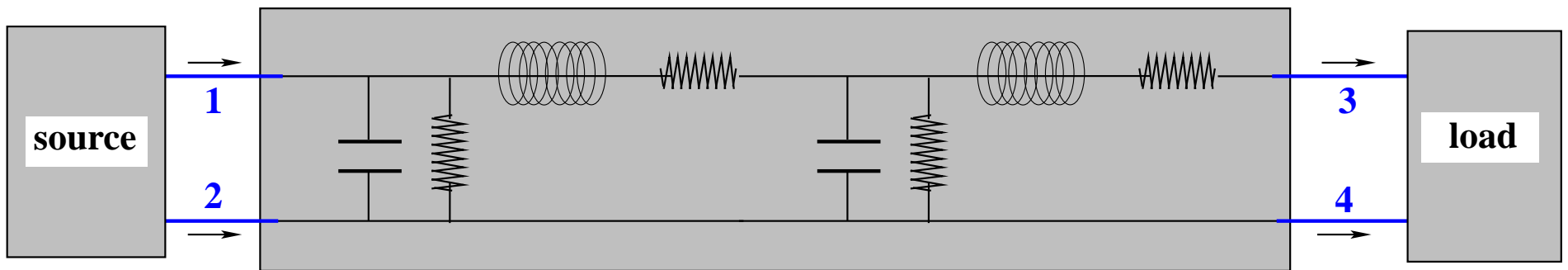
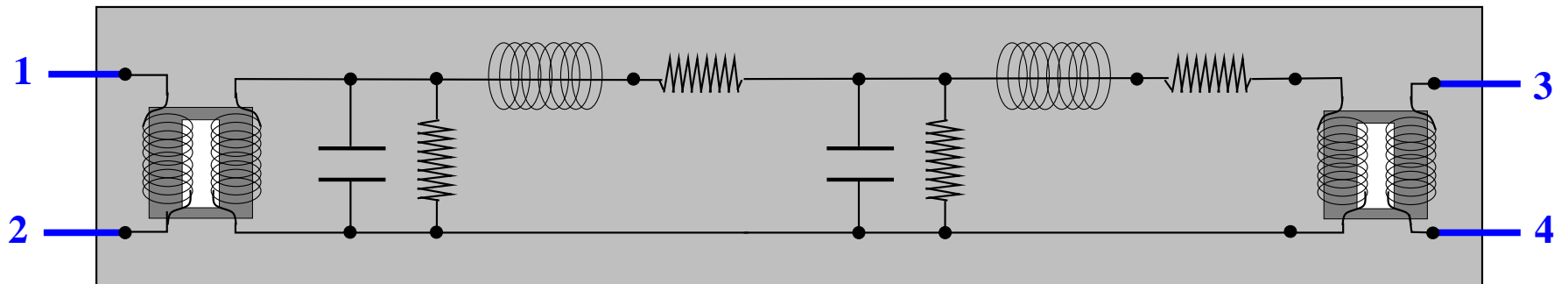
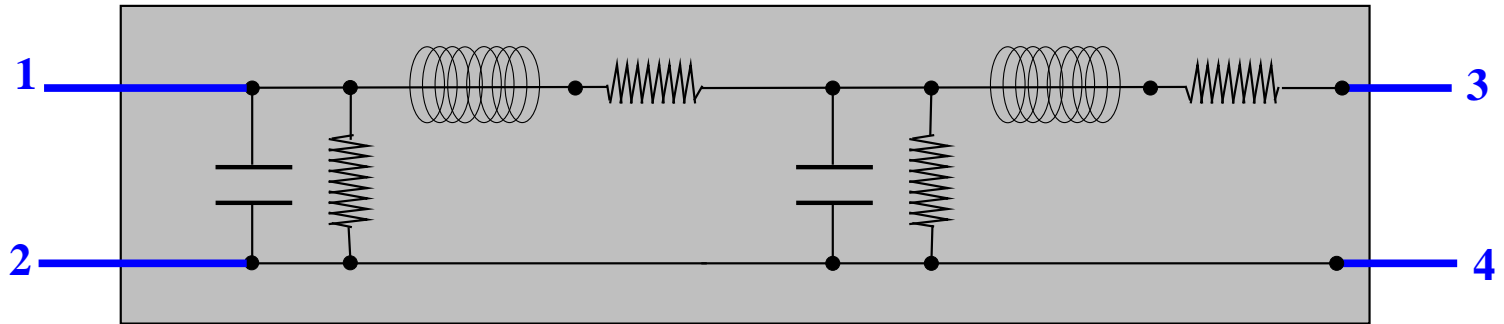
$$\mathcal{B}_{IV} = \{(I_1, V_1, I_2, V_2, \dots, V_N, I_N) \mid \exists P_1, P_3, \dots, P_{2N-1} : \\ (I_1, P_1, -I_1, P_1 - V_1, \dots, I_N, P_{2N-1}, -I_N, P_{2N-1} - V_N) \in \mathcal{B}_{IP}\}$$

Unit transformer termination

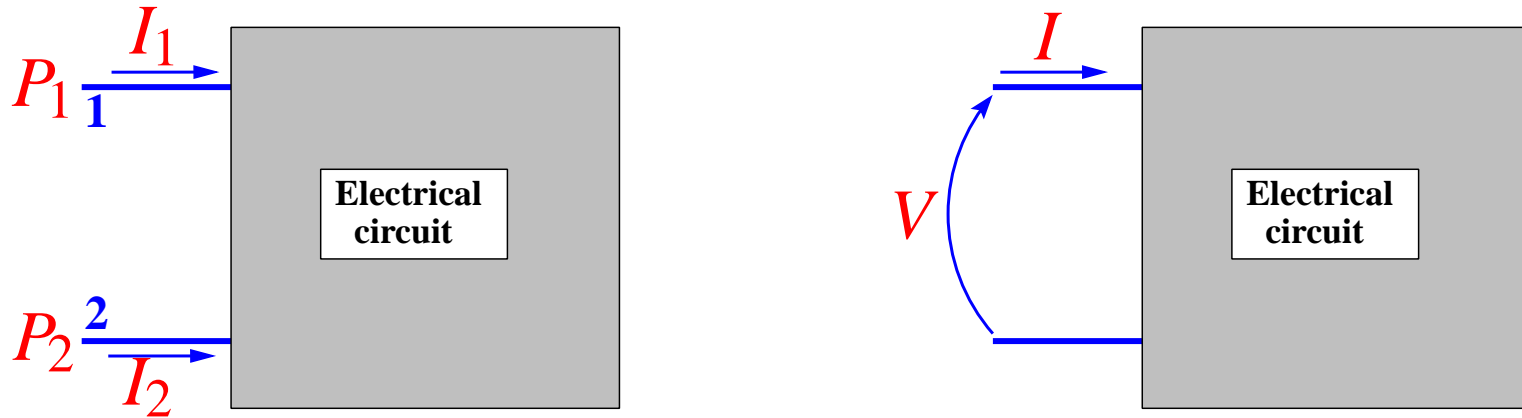


Port behavior \Leftrightarrow unit transformer termination.

4-terminal circuit or 2-port circuit?



2-terminal case



**A 2-terminal circuit is a 1-port,
(assuming KCL and KVL).**

Port = WLOG in 2-terminal case.

Port synthesis

Determine the port behaviors \mathcal{B}_{IV} that are achievable by interconnecting a finite number of

positive resistors,
positive capacitors,
positive inductors,
transformers,
and gyrators.

Synthesis conditions

Properties of the port behavior

The port behavior $B_{IV} \subseteq (\mathbb{R}^N \times \mathbb{R}^N)^{\mathbb{R}}$ of an interconnection of RLCTG's is

- ▶ **linear,**
- ▶ **time-invariant,**
- ▶ **finite-dimensional,**
- ▶ **hybrid,**
- ▶ **passive.**

Properties of the port behavior

‘Hybrid’ $:\Leftrightarrow$ Each port is either **current driven** or **voltage driven**, meaning \exists an input/output partition

$$\begin{array}{ccc} \text{input} & \begin{bmatrix} I_1 \\ \vdots \\ I_{N'} \\ V_{N'+1} \\ \vdots \\ V_N \end{bmatrix} & \mapsto & \begin{bmatrix} V_1 \\ \vdots \\ V_{N'} \\ I_{N'+1} \\ \vdots \\ I_N \end{bmatrix} & \text{output} \end{array}$$

LTIFD \rightsquigarrow rational transfer f'n $G \in \mathbb{R}(\xi)^{N \times N}$.

E.g. ‘impedance’ $V = ZI$, ‘admittance’ $I = YV$.

Properties of the port behavior

‘Passive’ $:\Leftrightarrow$

$$\llbracket (I, V) \in \mathcal{B}_{IV} \rrbracket \Rightarrow \llbracket \int_{-\infty}^0 I^\top(t) V(t) dt \geq 0 \rrbracket.$$

\Leftrightarrow **G is ‘positive real’**

i.e. $\llbracket s \in \mathbb{C}, \text{real part}(s) > 0 \rrbracket \Rightarrow$

$$\llbracket G(s) + G(\bar{s})^{\text{Hermitian conjugate}} \succeq 0 \rrbracket.$$

Basic synthesis result

A hybrid transfer function

$$G \in \mathbb{R}(\xi)^{N \times N}$$

**is synthesizable as the
 N -port behavior of an RLCTG circuit
if and only if G is
positive real.**

V. Belevitch, *Classical Network Theory*, Holden-Day, 1968.

**B.D.O. Anderson, S. Vongpanitlerd, *Network Analysis and Synthesis.
A Modern Systems Approach*, Prentice Hall, 1972.**

Other synthesis questions

Of interest:

- ▶ **RLCTG,**
- ▶ **RLCT,**
- ▶ **RLC,**
- ▶ **LC, RL, RC,**
- ▶ **RTG, RT, GT, T,**
- ▶ **R,**
- ▶ **etc.**

Memoryless synthesis

$$G \in \mathbb{R}^{N \times N}$$

Transformer circuit

$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 0 & N \\ -N^\top & 0 \end{bmatrix} \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$$

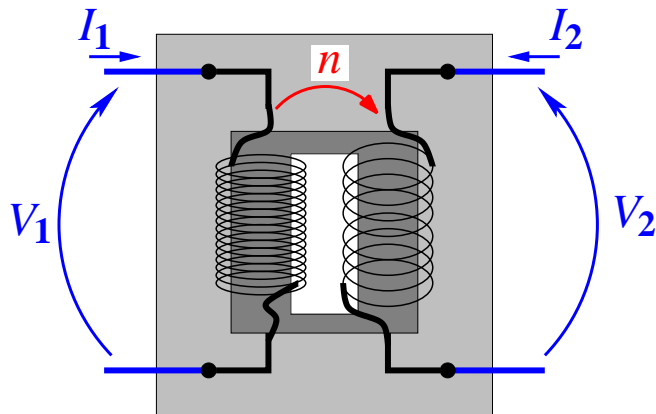
is realizable using transformers.

Transformer circuit

$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 0 & N \\ -N^\top & 0 \end{bmatrix} \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$$

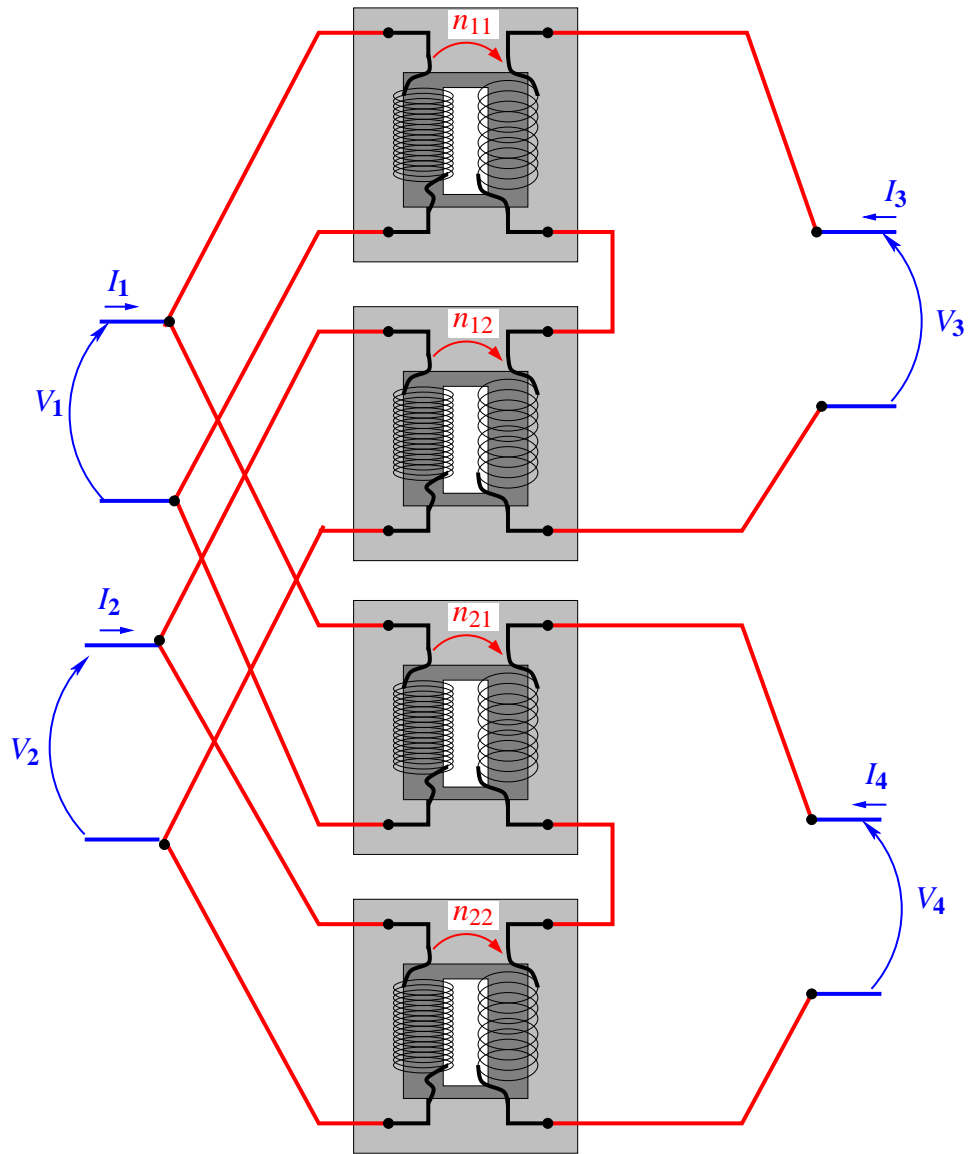
is realizable using transformers.

Multivariable generalization of

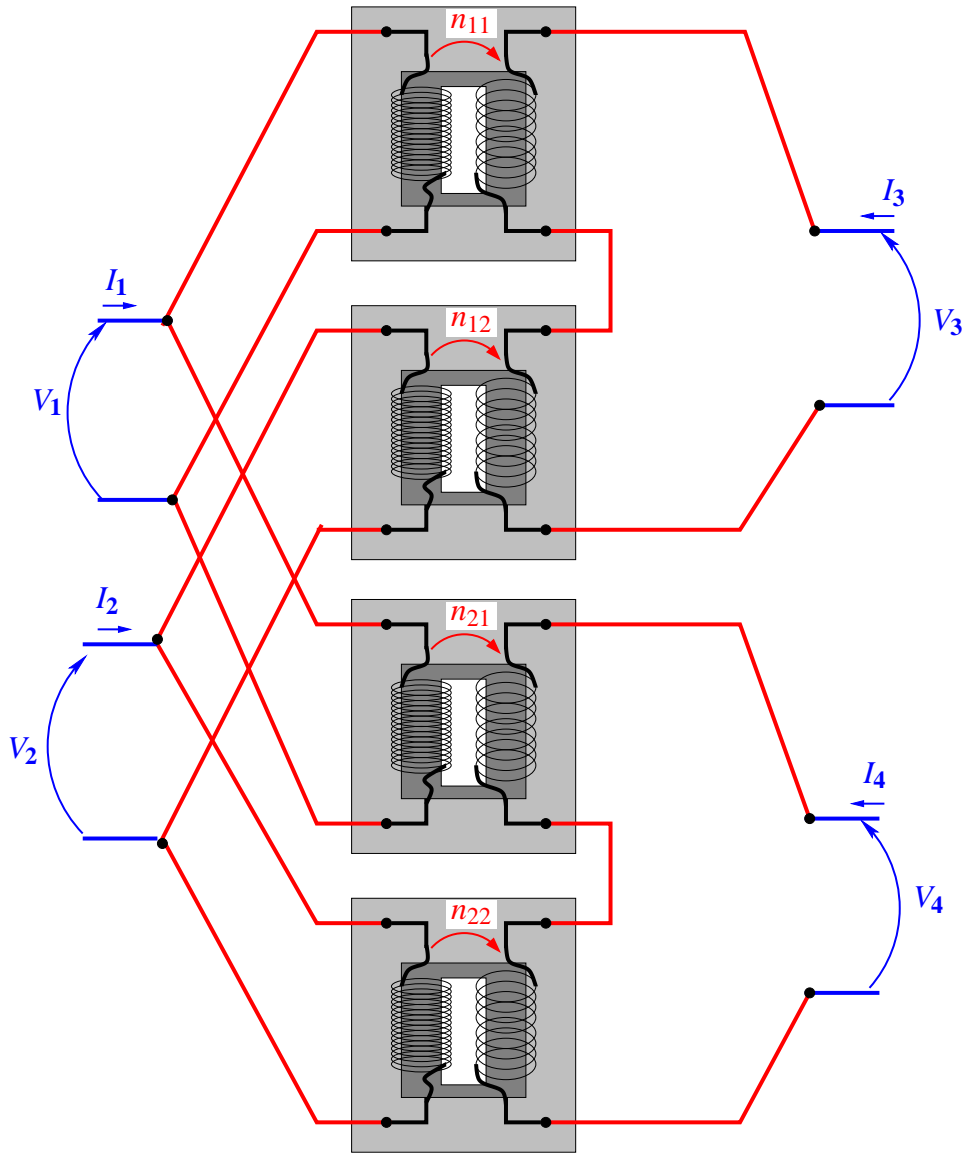


$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 0 & n \\ -n & 0 \end{bmatrix} \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$$

Transformer synthesis



Transformer synthesis



$$\begin{bmatrix} V_1 \\ V_2 \\ I_3 \\ I_4 \end{bmatrix} = \begin{bmatrix} 0 & 0 & n_{11} & n_{12} \\ 0 & 0 & n_{21} & n_{22} \\ -n_{11} & -n_{21} & 0 & 0 \\ -n_{12} & -n_{22} & 0 & 0 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ V_3 \\ V_4 \end{bmatrix}$$

Transformers & Gytrators

$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = G \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$$

with

$$G + G^T = 0$$

is realizable using gyrators and transformers.

Transformers & Gytrators

$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = G \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$$

with

$$G + G^\top = 0$$

is realizable using gyrators and transformers.

Assume (WLOG) $V = GI$ (impedance)

Factor $G = NJN^\top$ with $J = \begin{bmatrix} 0 & -I \\ I & 0 \end{bmatrix}$

Gyrator extraction

Synthesize using transformers

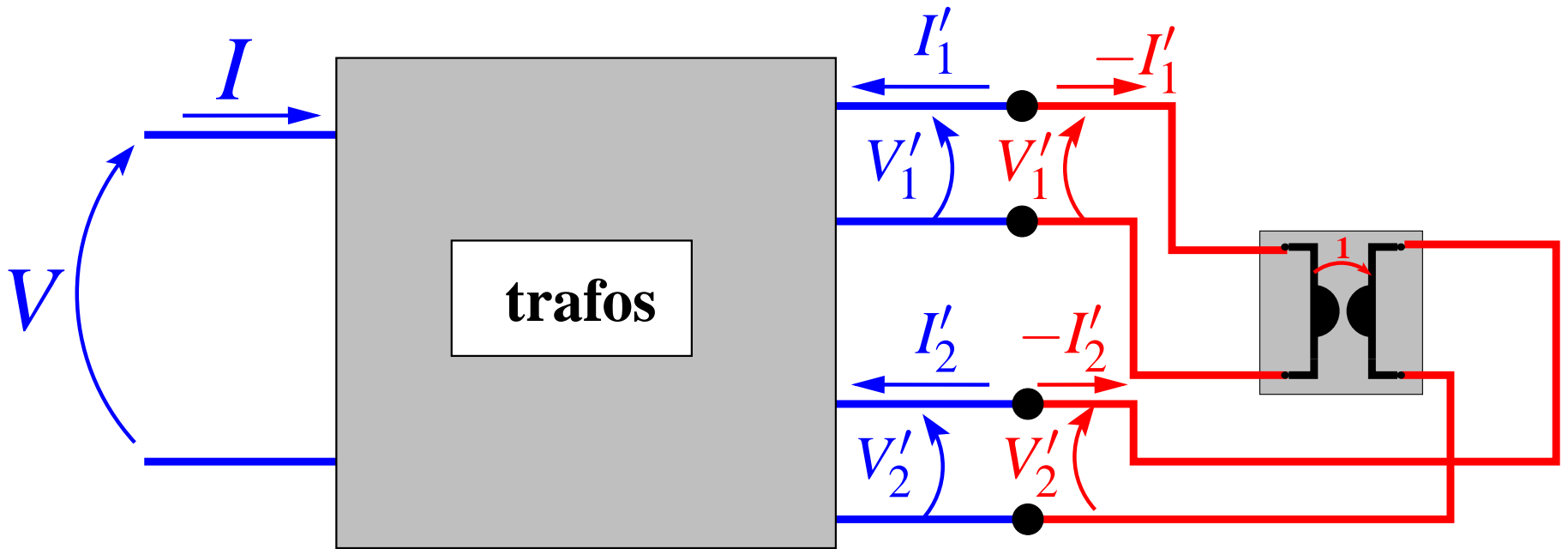
$$\begin{bmatrix} V \\ I' \end{bmatrix} = \begin{bmatrix} 0 & N \\ -N^\top & 0 \end{bmatrix} \begin{bmatrix} I \\ V' \end{bmatrix}.$$

Synthesize using gyrators

$$V' = -JI'.$$

$$V = NV' = -NJI' = NJN^\top I = GI.$$

Gyrator extraction



Transformers, gyrators, & resistors

$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = G \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$$

with

$$G + G^T \succeq 0$$

is realizable using resistors, gyrators, and trafos.

Resistor extraction

Assume (WLOG) $V = GI$ (impedance)

Write $\frac{1}{2}(G + G^\top) = S^\top S$.

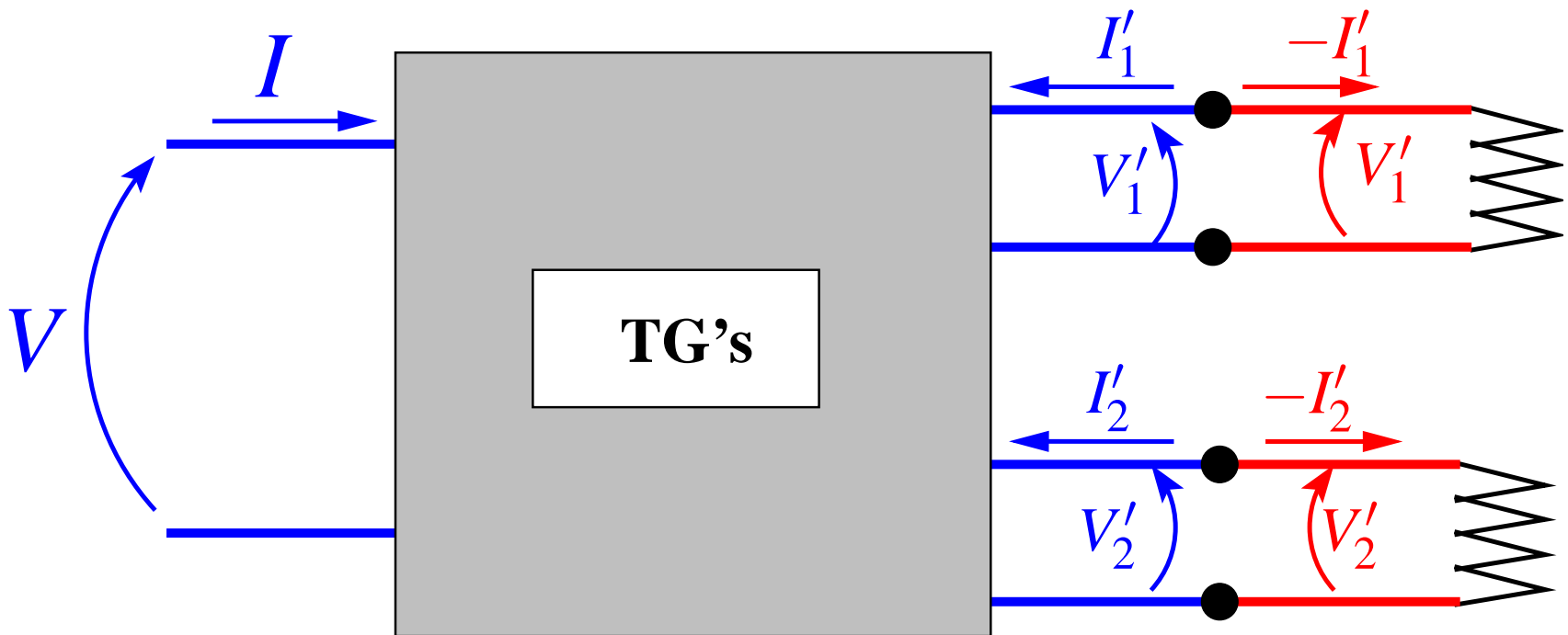
Synthesize using gyrators and transformers:

$$\begin{bmatrix} V \\ V' \end{bmatrix} = \begin{bmatrix} G - S^\top S & S^\top \\ -S & 0 \end{bmatrix} \begin{bmatrix} I \\ I' \end{bmatrix}.$$

Synthesize using resistors: $V' = -I'$.

$$V = (G - S^\top S)I + S^\top I' = (G - S^\top S)I - S^\top V' = (G - S^\top S)I + S^\top SI = GI.$$

Resistor extraction



Memoryless synthesis

We have hence proven that

$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = G \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$$

with

$$G + G^T \succeq 0$$

realizable using resistors, gyrators, transformers.

Dynamic synthesis

$$G \in \mathbb{R}(\xi)^{N \times N}$$

Realization theory

$G \in \mathbb{R}(\xi)^{N \times N}$. (WLOG) proper, impedance.

to be synthesized: $V = GI$

Realization theory

$G \in \mathbb{R}(\xi)^{N \times N}$. (WLOG) proper, impedance.

to be synthesized: $V = GI$

Well-known theorem: $\exists \left[\begin{array}{c|c} A & B \\ \hline C & D \end{array} \right]$ such that

$$\frac{d}{dt}x = Ax + BI, V = Cx + DI \rightsquigarrow G(s) = D + C(Is - A)^{-1}B.$$

$\exists \left[\begin{array}{c|c} A & B \\ \hline C & D \end{array} \right]$ such that the system is state-controllable and state-observable ('minimality').

KYP lemma

$G(s) = D + C(Is - A)^{-1}B$ is positive real

if and only if there exists $P = P^\top \succ 0$ such that

$\frac{d}{dt}x = Ax + BI, V = Cx + DI$ implies

$$\frac{d}{dt} \frac{1}{2} x^\top P x \leq I^\top V.$$

‘dissipation inequality’.

KYP lemma

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$\frac{d}{dt}x = Ax + BI, V = Cx + DI$ implies

$$\frac{d}{dt} \frac{1}{2} x^\top P x \leq I^\top V.$$

‘dissipation inequality’.

$$\Leftrightarrow \begin{bmatrix} A^\top P + PA & PB - C^\top \\ B^\top P - C & -D - D^\top \end{bmatrix} \preceq 0.$$

KYP lemma

$$P = S^\top S, \quad \left[\begin{array}{c|c} A & B \\ \hline C & D \end{array} \right] \mapsto \left[\begin{array}{c|c} S^{-1}AS & S^{-1}B \\ \hline CS & D \end{array} \right]$$

G positive real **if and only if** $\exists \left[\begin{array}{c|c} A & B \\ \hline C & D \end{array} \right]$ such that

$$G(s) = D + C(Is - A)^{-1}B,$$

$$\begin{bmatrix} -A & -B \\ C & D \end{bmatrix} + \begin{bmatrix} -A & -B \\ C & D \end{bmatrix}^\top \preceq 0.$$

i.e., $\frac{d}{dt} \frac{1}{2} \|x\|^2 \leq I^\top V$. **‘Passive realization’.**

Reactance extraction

Synthesize using trafos, gyrators, and resistors:

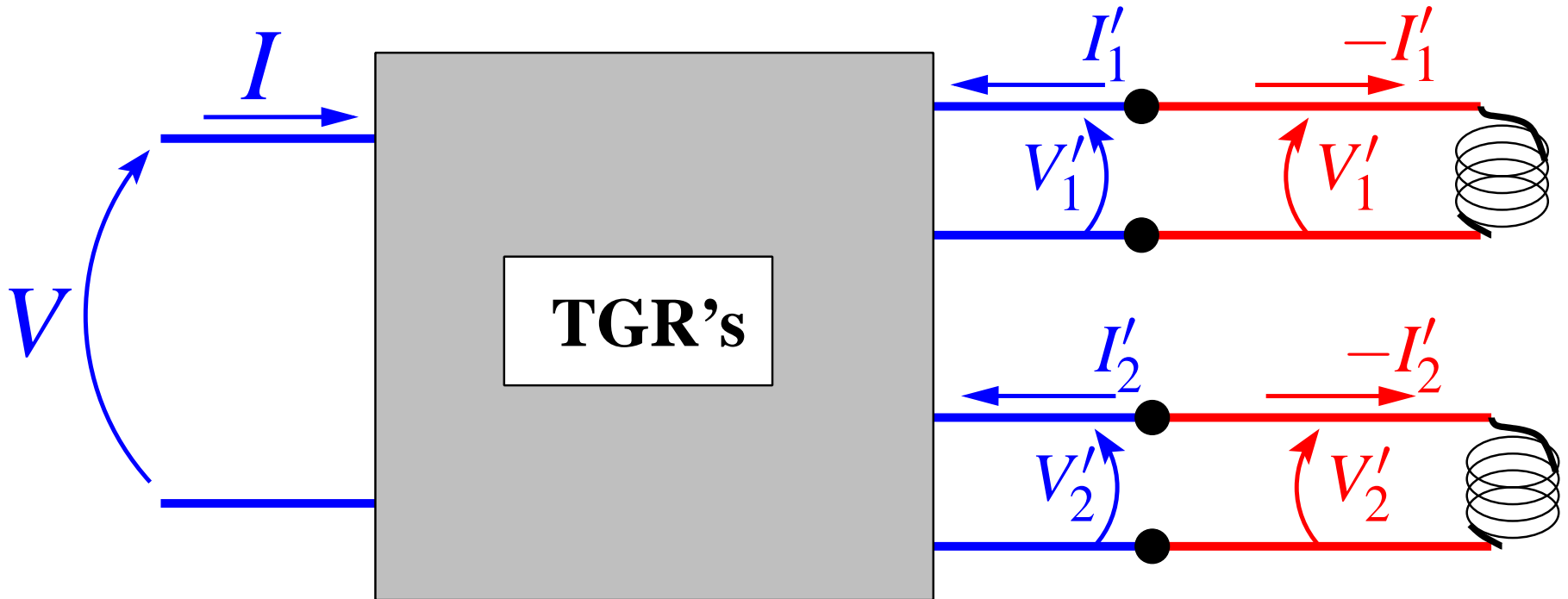
$$\begin{bmatrix} V' \\ V \end{bmatrix} = \begin{bmatrix} -A & -B \\ C & D \end{bmatrix} \begin{bmatrix} I' \\ I \end{bmatrix} .$$

Synthesize using inductors:

$$-\frac{d}{dt}I' = V' .$$

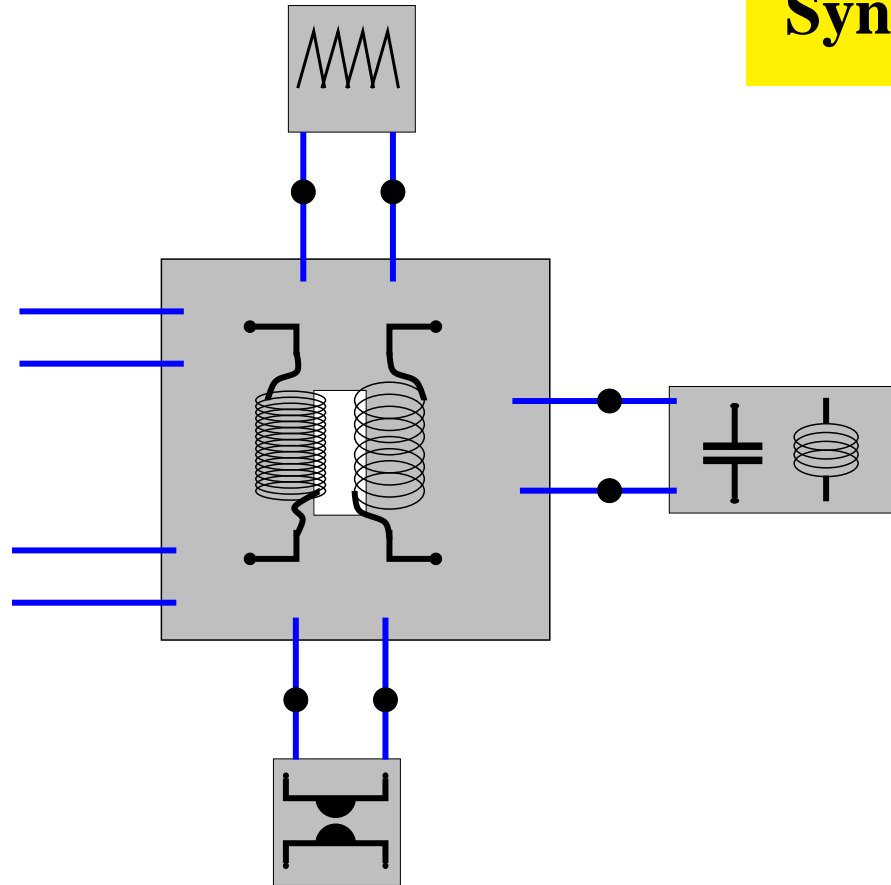
$$\begin{bmatrix} V' = -AI' - BI \\ I = CI' + DI \end{bmatrix} \Leftrightarrow \begin{bmatrix} \frac{d}{dt}I' = AI' + BI \\ I = CI' + DI \end{bmatrix} \Leftrightarrow \text{tf fn equals } G .$$

Reactance extraction



Summary

Synthesis



Positive realness \Rightarrow KYP lemma $\Rightarrow \exists$ passive realization.

Passive realization + reactance extraction \Rightarrow memoryless case.

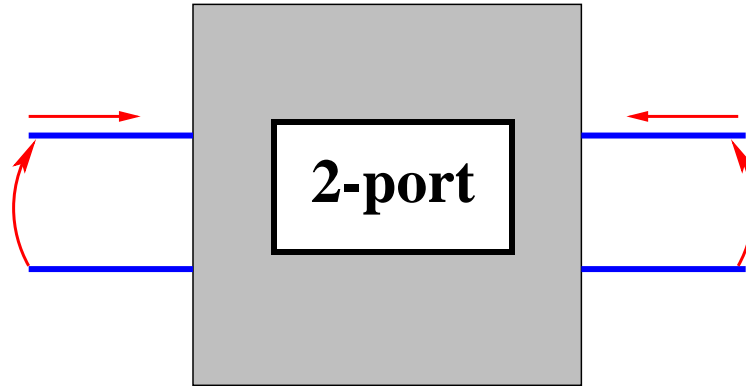
$G + G^\top \succeq 0$ + resistor extraction $\Rightarrow G + G^\top = 0$.

$G + G^\top = 0$ + gyrator extraction \Rightarrow symmetric $G + G^\top = 0$.

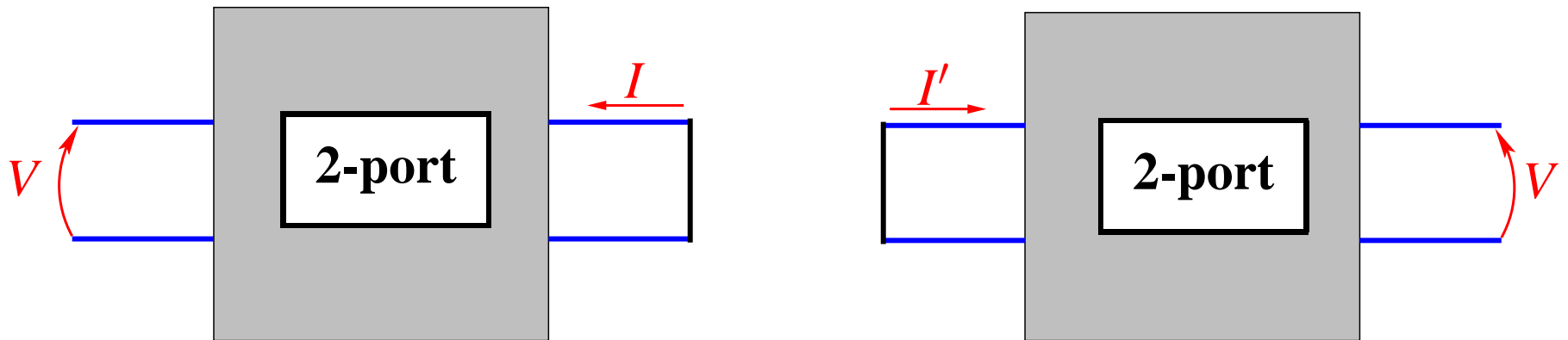
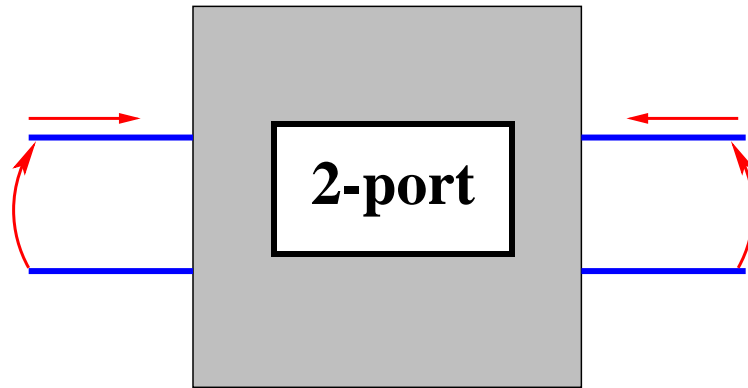
symmetric $G + G^\top = 0 \Rightarrow$ bank of transformers.

Reciprocity

Symmetry



Symmetry



[[reciprocity]] $:\Leftrightarrow$ $[[I = I']$.

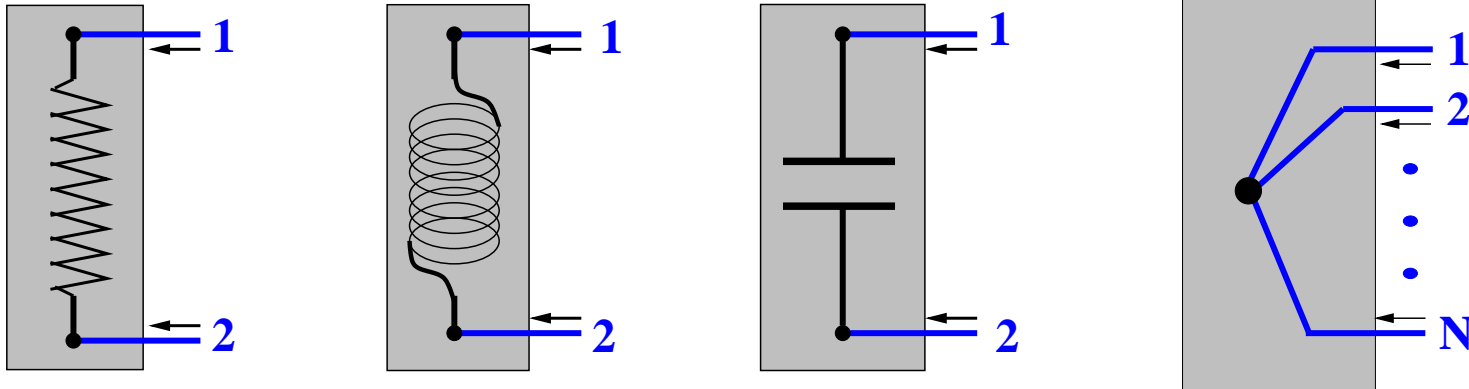
Reciprocity

$V = ZI$ is reciprocal $\Leftrightarrow Z = Z^T$.

$$\begin{bmatrix} V_1 \\ \vdots \\ V_{N'} \\ \text{---} \\ I_{N'+1} \\ \vdots \\ I_N \end{bmatrix} = G \begin{bmatrix} I_1 \\ \vdots \\ I_{N'} \\ \text{---} \\ V_{N'+1} \\ \vdots \\ V_N \end{bmatrix} \text{ is reciprocal}$$

$\Leftrightarrow \begin{bmatrix} G_{11} & G_{12} \\ G_{21} & G_{22} \end{bmatrix}$ satisfies $G_{11} = G_{11}^T, G_{22} = G_{22}^T, G_{12} = -G_{21}^T$.

Reciprocal building blocks

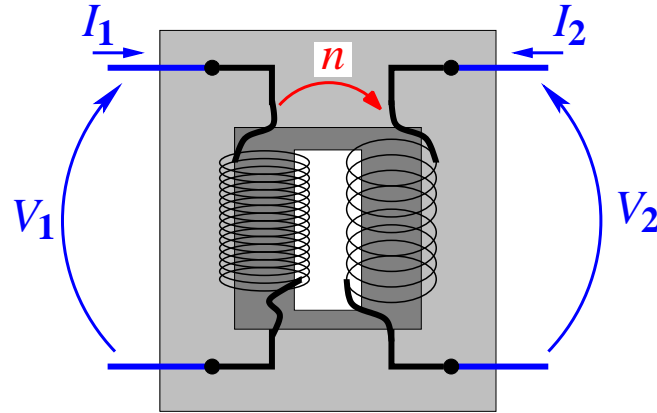


$V = RI, \frac{d}{dt}I = LV, I = C \frac{d}{dt}V$, trivially reciprocal.

Connector:

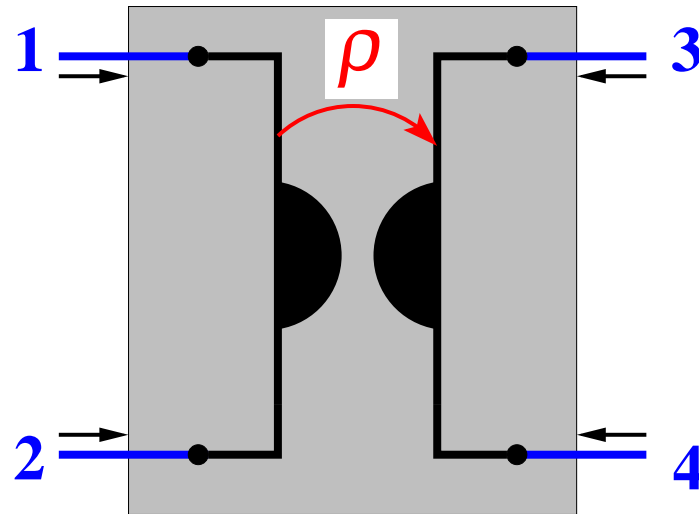
$$\begin{bmatrix} P_1 \\ P_2 \\ \vdots \\ P_{n-1} \\ \text{---} \\ I_n \end{bmatrix} = \begin{bmatrix} 0 & 0 & \cdots & 0 & 1 \\ 0 & 0 & \cdots & 0 & 1 \\ & & \vdots & & \\ 0 & 0 & \cdots & 0 & 1 \\ -1 & -1 & \cdots & -1 & 0 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ \vdots \\ I_{n-1} \\ \text{---} \\ P_n \end{bmatrix} \Rightarrow \text{reciprocal.}$$

Reciprocal building blocks



$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 0 & n \\ -n & 0 \end{bmatrix} \begin{bmatrix} I_1 \\ V_2 \end{bmatrix} \Rightarrow \text{trafo is reciprocal.}$$

Non-reciprocal building block



$$V_1 = \rho I_2, \quad V_2 = -\rho I_1;$$

$$\begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} 0 & \rho \\ -\rho & 0 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} \Rightarrow \text{gyrator is NOT reciprocal.}$$

Interconnection

The interconnection of reciprocal is reciprocal.

RLCT circuits are therefore

- ▶ **linear, time-invariant, finite-dimensional,**
- ▶ **hybrid,**
- ▶ **passive,**
- ▶ **reciprocal.**

Reciprocal synthesis

Reciprocal synthesis conditions

$$\begin{bmatrix} V_1 \\ \vdots \\ V_{N'} \\ \text{---} \\ I_{N'+1} \\ \vdots \\ I_N \end{bmatrix} = G \begin{bmatrix} I_1 \\ \vdots \\ I_{N'} \\ \text{---} \\ V_{N'+1} \\ \vdots \\ V_N \end{bmatrix}$$

is synthesizable as a RLCT circuit if and only if

$G \in \mathbb{R}(\xi)^{N \times N}$ is **positive real** &

$$G = \begin{bmatrix} G_{11} & G_{12} \\ G_{21} & G_{22} \end{bmatrix} \text{ satisfies } G_{11} = G_{11}^\top, G_{22} = G_{22}^\top, G_{12} = -G_{21}^\top.$$

Memoryless reciprocal synthesis procedure

Memoryless case. First impedance:

$$V = ZI, \quad Z \in \mathbb{R}^{N \times N}, \quad Z = Z^T \succeq 0.$$

Memoryless reciprocal synthesis procedure

Memoryless case. First impedance:

$$V = ZI, \quad Z \in \mathbb{R}^{N \times N}, \quad Z = Z^\top \succeq 0.$$

Write $Z = NN^\top$. Synthesize using transformers:

$$\begin{bmatrix} V \\ V' \end{bmatrix} = \begin{bmatrix} 0 & N \\ -N^\top & 0 \end{bmatrix} \begin{bmatrix} I \\ I' \end{bmatrix}.$$

Synthesize using resistors: $V' = -I'$.

$$V = NI' = -NV' = NN^\top I = ZI.$$

Memoryless reciprocal synthesis procedure

Memoryless case. General case:

$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} G_{11} & G_{12} \\ -G_{12}^\top & G_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$$

Maximally current driven

$$\Rightarrow G_{22} = 0, G_{11} = G_{11}^\top \succeq 0.$$

Memoryless reciprocal synthesis procedure

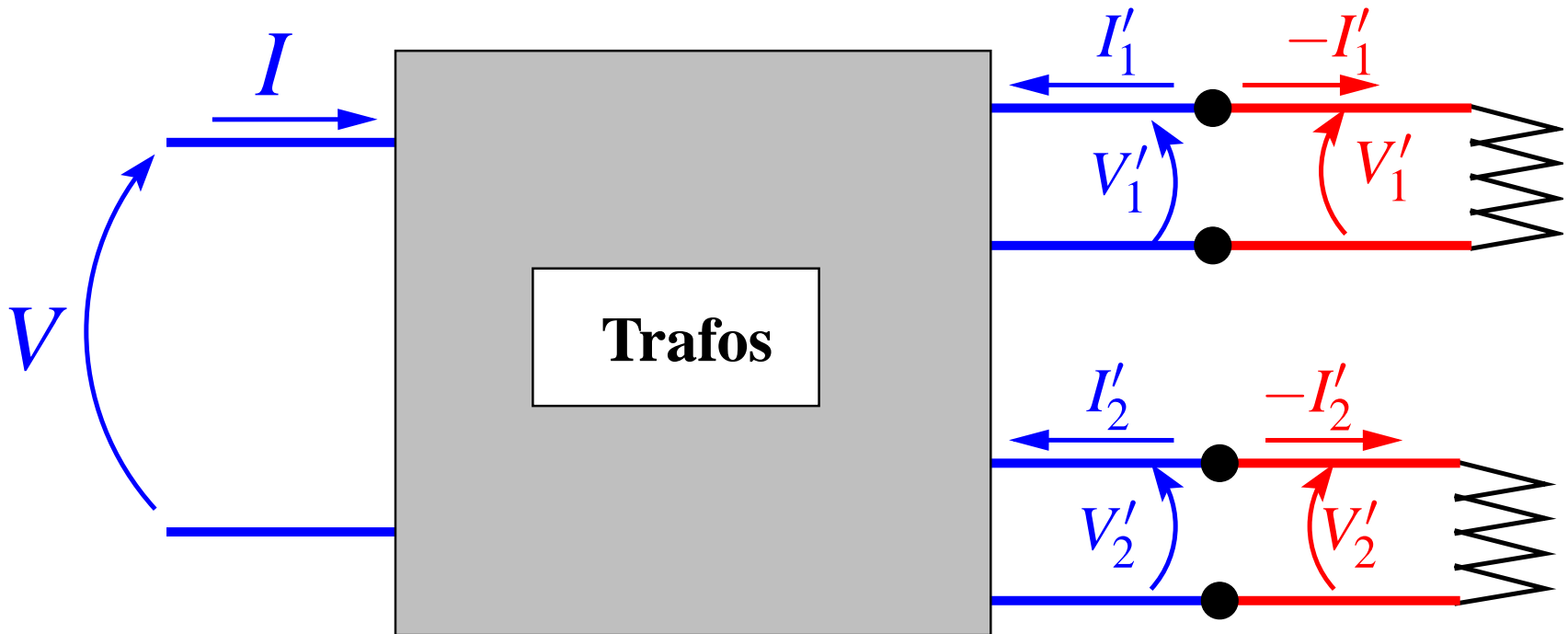
Write $G_{11} = NN^\top$. **Synthesize using transformers:**

$$\begin{bmatrix} V_1 \\ V' \\ I_2 \end{bmatrix} = \begin{bmatrix} 0 & N & G_{12} \\ -N^\top & 0 & 0 \\ -G_{12}^\top & 0 & 0 \end{bmatrix} \begin{bmatrix} I_1 \\ I' \\ V_2 \end{bmatrix}.$$

Synthesize using resistors: $V' = -I'$.

$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} NI' + G_{12}V_2 \\ -G_{12}^\top I_1 \end{bmatrix} = \begin{bmatrix} -NV' + G_{12}V_2 \\ -G_{12}^\top I_1 \end{bmatrix} = \begin{bmatrix} NN^\top I_1 + G_{12}V_2 \\ -G_{12}^\top I_1 \end{bmatrix} = G \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}.$$

Resistor extraction



Dynamic reciprocal synthesis procedure

Reciprocity and passivity:

$G \in \mathbb{R}(\xi)^{N \times N}$ is **positive real** &

$$G = \begin{bmatrix} G_{11} & G_{12} \\ G_{21} & G_{22} \end{bmatrix} \text{ satisfies } G_{11} = G_{11}^\top, G_{22} = G_{22}^\top, G_{12} = -G_{21}^\top.$$

i.e.

$$G \text{ p.r. and } G \Sigma_e = \Sigma_e G^\top, \Sigma_e = \begin{bmatrix} I_{n_1} & 0 \\ 0 & -I_{n_2} \end{bmatrix}.$$

Dynamic reciprocal synthesis procedure

\exists realization $\left[\begin{array}{c|c} A & B \\ \hline C & D \end{array} \right]$ such that simultaneously

$$\begin{bmatrix} -A & -B \\ C & D \end{bmatrix} + \begin{bmatrix} -A & -B \\ C & D \end{bmatrix}^\top \succcurlyeq 0.$$

$$\begin{bmatrix} -A & -B \\ C & D \end{bmatrix} \begin{bmatrix} \Sigma_i & 0 \\ 0 & \Sigma_e \end{bmatrix} = \begin{bmatrix} -A & -B \\ C & D \end{bmatrix}^\top \begin{bmatrix} \Sigma_i & 0 \\ 0 & \Sigma_e \end{bmatrix}.$$

$\Sigma_e = \begin{bmatrix} I_{n_3} & 0 \\ 0 & -I_{n_4} \end{bmatrix}$: ‘internal passivity’ and ‘internal reciprocity’.

Various proofs

- ▶ **Several constructions in**
B.D.O. Anderson, S. Vongpanitlerd, *Network Analysis and Synthesis. A Modern Systems Approach*, Prentice Hall, 1972.
- ▶ **JCW, Dissipative dynamical systems, Part II,**
***Archive for Rational Mechanics and Analysis*, 45, pp. 352-393,**
1972.
Sol. set to KYP is convex compact, and Brouwer's fixpoint thm.
- ▶ **Reciprocal synthesis is also in the classical literature.**
V. Belevitch, *Classical Network Theory*, Holden-Day, 1968.
- ▶ **T. Reis and JCW, A balancing approach to the realization of**
systems with internal passivity and reciprocity, *SCLetters*, 60,
pp. 69-74, 2011.
Positive real balancing \Rightarrow internal passive and reciprocal.

Reactance extraction

Trafos and resistors:

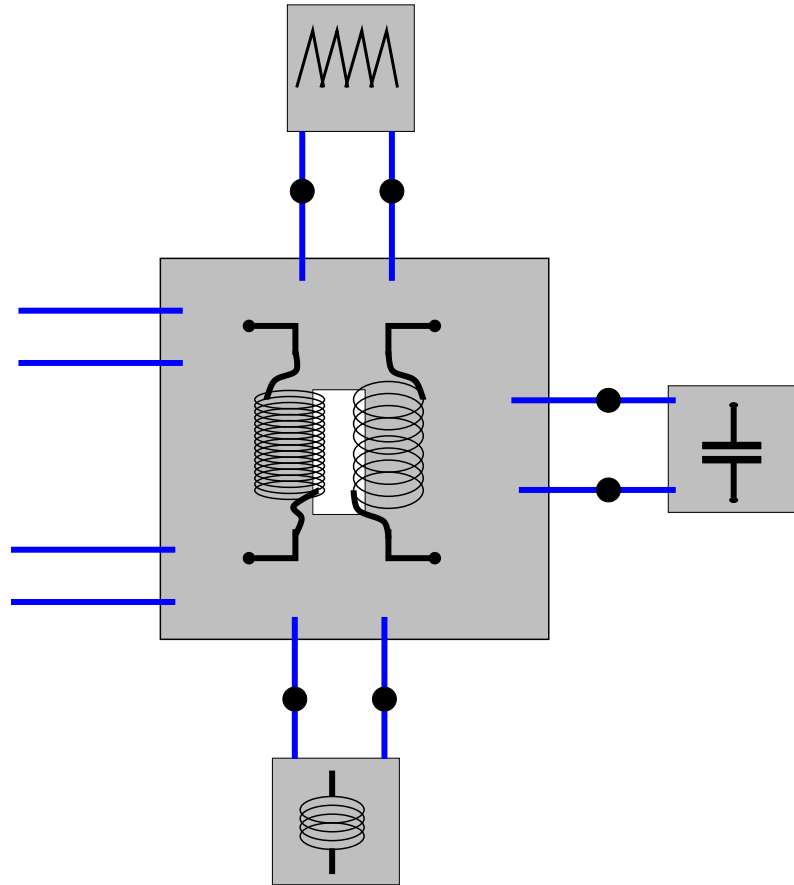
$$\begin{bmatrix} V_1' \\ I_2' \\ V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} -A & -B \\ C & D \end{bmatrix} \begin{bmatrix} I_1' \\ V_2' \\ I_1 \\ V_2 \end{bmatrix} .$$

Inductors: $-\frac{d}{dt}I_1' = V_1'.$

Capacitors: $-\frac{d}{dt}V_2' = I_2'.$

$$\begin{bmatrix} \frac{d}{dt}I_1' \\ \frac{d}{dt}V_2' \\ V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} -V_1' \\ -I_2' \\ V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} I_1' \\ V_2' \\ I_1 \\ V_2 \end{bmatrix} \Leftrightarrow \text{tf fn equals } G.$$

Summary



Conclusions

Conclusions

- ▶ **Circuit synthesis shows the power and effectiveness of state space thinking.**
- ▶ **Why ports instead of terminals in case $N > 1$, surely $N > 2$?**
- ▶ **Open problem: Do Bott-Duffin from state point of view. Requires non-minimal, non-controllable realizations.**
- ▶ **Open problem: Resistive N -ports.**
Transformerless N -terminal resistive case solved.

Copies of the lecture frames available from/at

<http://www.esat.kuleuven.be/~jwillems>

Thank you

Thank you

Thank you

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