



# RATIONAL SYMBOLS

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**Seminar, Kyoto University**

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## Joint research with



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# Outline

**I. Behaviors defined by rational symbols**

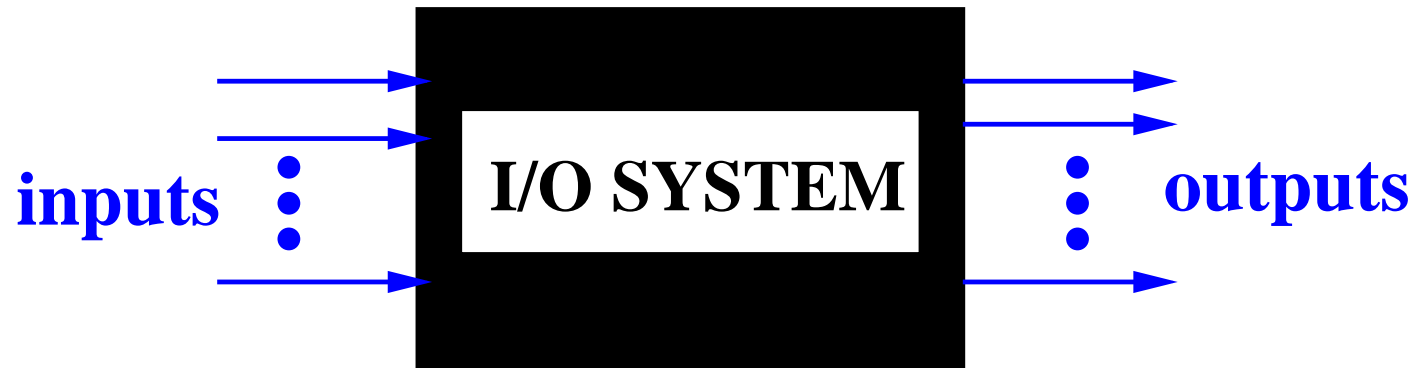
**II. Model reduction**

**(III. Parametrization of the stabilizing controllers)**

# Introduction

## Motivation

In system theory, it is customary to think of dynamical models in terms of inputs and outputs.



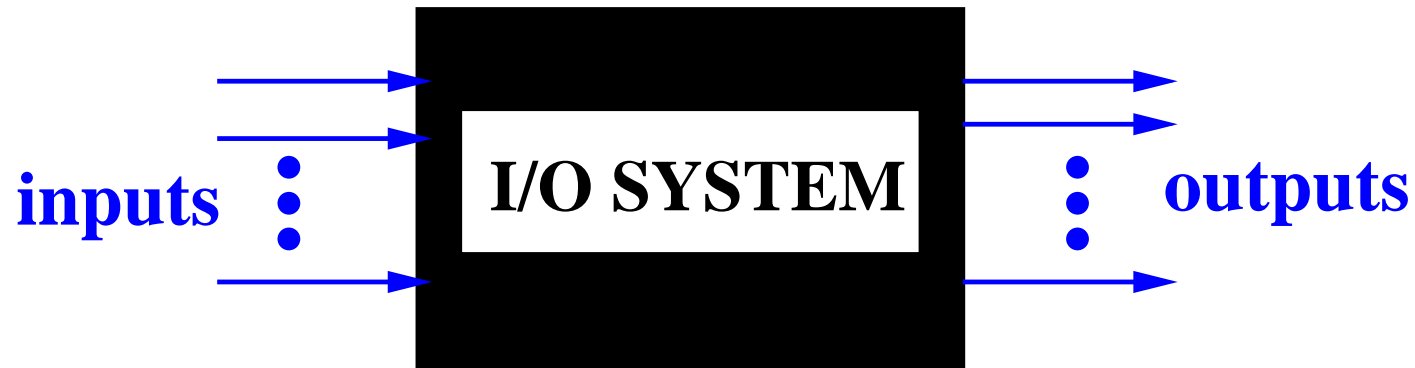
~> say,

$$p_0 y + p_1 \frac{d}{dt} y + \cdots + p_n \frac{d^n}{dt^n} y = q_0 u + q_1 \frac{d}{dt} u + \cdots + q_n \frac{d^n}{dt^n} u$$

**i.e.,**  $p\left(\frac{d}{dt}\right)y = q\left(\frac{d}{dt}\right)u,$

## Motivation

In system theory, it is customary to think of dynamical models in terms of inputs and outputs.



$$p\left(\frac{d}{dt}\right)y = q\left(\frac{d}{dt}\right)u, \quad \text{or} \quad y = F(s)u$$

with  $p, q$  polynomials, or  $F$  a rational transfer function.

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In the present talk, we will

- (for good reasons) make no distinction between  $u$  and  $y$

$\rightsquigarrow$  system variables

$$w = \begin{bmatrix} u \\ y \end{bmatrix}$$

- interpret  $F$ , not in terms of Laplace transforms, but in terms of differential equations.  
Important for, among other things, pedagogical reasons.

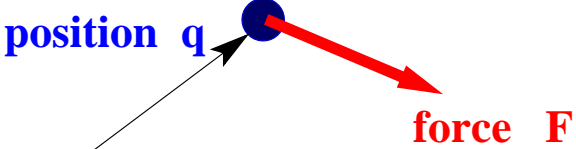
## Example

unit mass

position  $q$

force  $F$

$\dagger$   $F = \frac{d^2}{dt^2} q, \quad w = \begin{bmatrix} F \\ q \end{bmatrix}, \quad F, q \in \mathbb{R}^3, w \in \mathbb{R}^6$

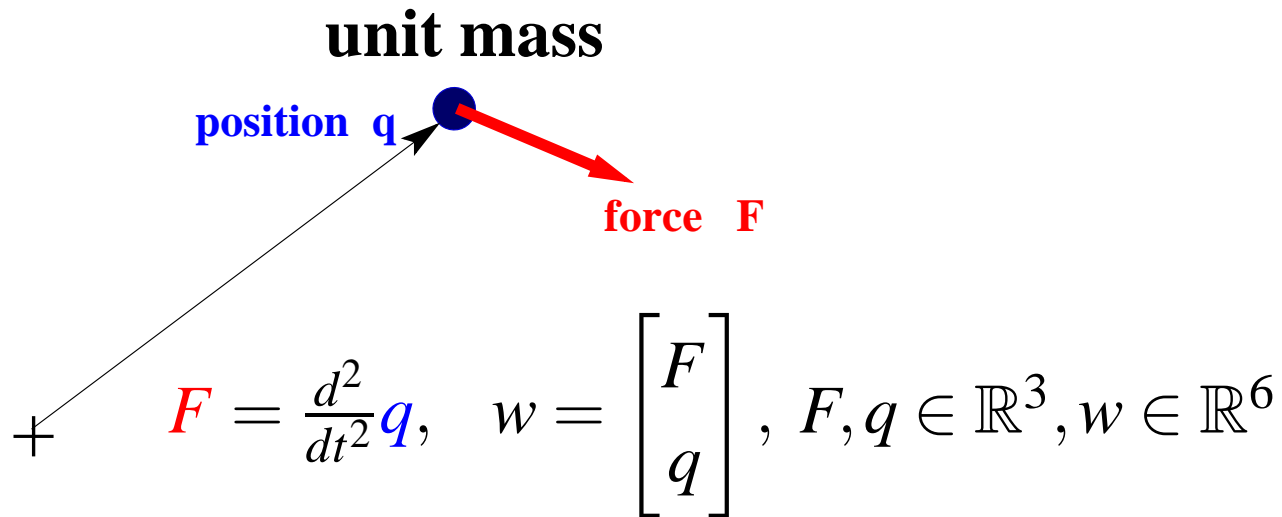
A diagram showing a blue dot representing a unit mass. A black arrow labeled 'position q' points from a '+' sign to the dot. A red arrow labeled 'force F' points away from the dot.



**Isaac Newton**  
by **William Blake**

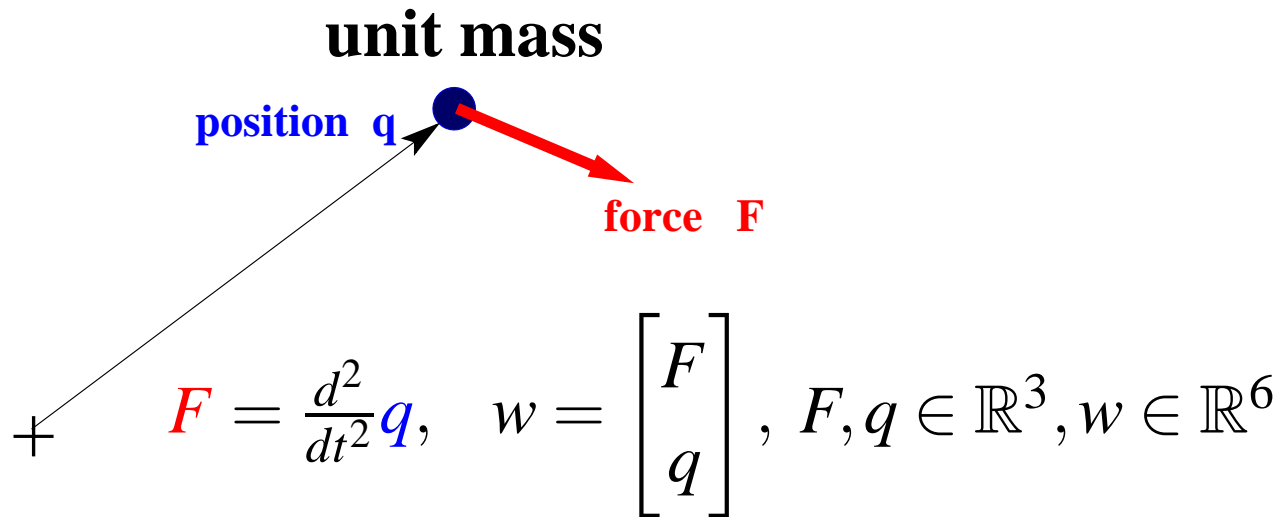


## Example



$$\begin{bmatrix} I_{3 \times 3} \vdots -\left(\frac{d}{dt}\right)^2 I_{3 \times 3} \end{bmatrix} w = 0 \rightsquigarrow q = \frac{1}{\left(\frac{d}{dt}\right)^2} F \rightsquigarrow \begin{bmatrix} -\frac{1}{\left(\frac{d}{dt}\right)^2} I_{3 \times 3} \vdots I_{3 \times 3} \end{bmatrix} w = 0$$

## Example



$$\left[ I_{3 \times 3} \quad \vdots \quad -\left(\frac{d}{dt}\right)^2 I_{3 \times 3} \right] w = 0 \quad \rightsquigarrow \quad q = \frac{1}{\left(\frac{d}{dt}\right)^2} F \quad \rightsquigarrow \quad \left[ -\frac{1}{\left(\frac{d}{dt}\right)^2} I_{3 \times 3} \quad \vdots \quad I_{3 \times 3} \right] w = 0$$

In the scalar case with simple polynomials, it is easy to see how to proceed, but with general multivariable rational functions, less obvious. Today's pbm: **What do we mean by**

$$y = \frac{q\left(\frac{d}{dt}\right)}{p\left(\frac{d}{dt}\right)} u, \quad \text{or} \quad G\left(\frac{d}{dt}\right) w = 0 \quad \text{with } G \text{ rational?}$$

# PART I

**Linear time-invariant differential systems**

**LTIDSs**

**defined by rational symbols**

# LTIDSs

A system  $\rightarrow (\mathbb{T}, \mathbb{W}, \mathcal{B})$  where

- $\mathbb{T}$  = set of independent variables
  - $\mathbb{T}$  = time  $\leadsto$  dynamical systems
  - $\mathbb{T}$  = time & space  $\leadsto$  distributed systems
- $\mathbb{W}$  = set of dependent variables; ‘signal space’
- $\mathcal{B}$  the *behavior*  $\rightarrow \mathcal{B} \subseteq \mathbb{W}^{\mathbb{T}}$ ,  
set of trajectories  $w : \mathbb{T} \rightarrow \mathbb{W}$

$w : \mathbb{T} \rightarrow \mathbb{W}$  belongs to  $\mathcal{B}$  means:

the model ‘accepts’ the trajectory  $w$

## LTIDSs

A **dynamical** system  $\rightarrow (\mathbb{R}, \mathbb{R}^w, \mathcal{B})$  where

- $\mathbb{T}$  = set of independent variables  $\rightsquigarrow \mathbb{T} = \mathbb{R}$  ‘time’
- $\mathbb{W}$  = set of dependent variables;  $\rightsquigarrow \mathbb{W} = \mathbb{R}^w$
- $\mathcal{B}$  the *behavior*  $\rightarrow \mathcal{B} \subseteq \mathbb{W}^{\mathbb{T}}$ ,  
time-trajectories  $w : \mathbb{T} \rightarrow \mathbb{W}$

$\mathcal{B}$  = the solutions of a set of

**linear constant coefficient ODEs**

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$\mathcal{B}$  = the solutions of

$$R_0 w + R_1 \frac{d}{dt} w + \cdots + R_L \frac{d^L}{dt^L} w = 0, \quad R_0, R_1, \dots \text{ matrices}$$

**Polynomial matrix notation**  $\rightsquigarrow R \left( \frac{d}{dt} \right) w = 0 \quad R \in \mathbb{R}[\xi]^{\bullet \times w}$

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- $\mathcal{B}$  the *behavior*  $\rightarrow \mathcal{B} \subseteq \mathbb{W}^{\mathbb{T}}$ ,

$\mathcal{B}$  = the  $\mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w)$ -solutions of

$$R_0 w + R_1 \frac{d}{dt} w + \cdots + R_L \frac{d^L}{dt^L} w = 0$$

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# Representations

**Behaviors of LTIDSs allow many useful representations**

- **As the set of solutions of  $R \left( \frac{d}{dt} \right) w = 0$   $R \in \mathbb{R} [\xi]^{\bullet \times w}$**



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- As the set of solutions of  $R \left( \frac{d}{dt} \right) w = 0$   $R \in \mathbb{R} [\xi]^{\bullet \times w}$
- With input/output partition

$$P \left( \frac{d}{dt} \right) y = Q \left( \frac{d}{dt} \right) u \quad w \cong \begin{bmatrix} u \\ y \end{bmatrix} \quad \det(P) \neq 0, P^{-1}Q \text{ proper}$$

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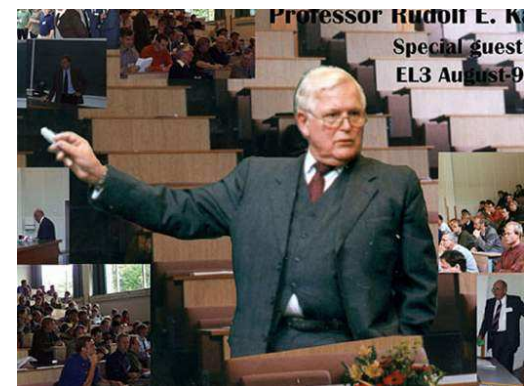
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$$P \left( \frac{d}{dt} \right) y = Q \left( \frac{d}{dt} \right) u \quad w \cong \begin{bmatrix} u \\ y \end{bmatrix}$$

- Input/state/output representation  
 $\exists$  matrices  $A, B, C, D$  such that  
 $\mathcal{B}$  consists of all  $w$ 's generated by

$$\frac{d}{dt} x = Ax + Bu, \quad y = Cx + Du \quad w \cong \begin{bmatrix} u \\ y \end{bmatrix}$$

- ...



Rudolf E. Kalman

# Rational Symbols

## Rational representations

In signal processing, control, etc., we often meet models that involve rational functions, instead of ODEs. Cfr. transfer functions,

$$y = F('s')u$$

etc.  $\rightsquigarrow$

Let  $G \in \mathbb{R}(\xi)^{\bullet \times w}$ , and consider the ‘differential equation’

$$G\left(\frac{d}{dt}\right)w = 0 \quad G \text{ is called the ‘symbol’}$$

**What do we mean by its solutions, i.e. by the behavior?**

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What do we mean by its solutions, i.e. by the behavior?

$$\begin{aligned} \llbracket M \text{ left prime} \rrbracket &: \Leftrightarrow \llbracket \llbracket M = FM' \rrbracket \Rightarrow \llbracket F \text{ unimodular} \rrbracket \rrbracket \\ &\Leftrightarrow \exists H \text{ such that } MH = I. \end{aligned}$$

In scalar case,  $M = \begin{bmatrix} m_1 & m_2 & \cdots & m_n \end{bmatrix}$ , this means:

$m_1, m_2, \cdots, m_n$  have no common root.

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**What do we mean by its solutions, i.e. by the behavior?**

Let  $(P, Q)$  be a **left coprime** polynomial factorization of  $G$

**i.e.  $P, Q \in \mathbb{R}[\xi]^{\bullet \times \bullet}$ ,  $\det(P) \neq 0$ ,  $G = P^{-1}Q$ ,  $[P : Q]$  left-prime.**

**E.g., in scalar case, means  $P$  and  $Q$  have no common roots.**

## Rational representations

Let  $(P, Q)$  be a **left coprime** polynomial factorization of  $G$

$$\llbracket G\left(\frac{d}{dt}\right)w = 0 \rrbracket \Leftrightarrow \llbracket P^{-1}Q\left(\frac{d}{dt}\right)w = 0 \rrbracket \Leftrightarrow \llbracket Q\left(\frac{d}{dt}\right)w = 0 \rrbracket$$

**By definition**, therefore, the behavior of  $G\left(\frac{d}{dt}\right)w = 0$  is equal to the behavior of  $Q\left(\frac{d}{dt}\right)w = 0$ .

## Rational representations

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### Justification:

**1.  $G$  proper.  $G(\xi) = C(I\xi - A)^{-1}B + D$  controllable realization. Consider output nulling inputs:**

$$\frac{d}{dt}x = Ax + Bw, \quad 0 = Cx + Dw$$

**This set of  $w$ 's are exactly those that satisfy  $G\left(\frac{d}{dt}\right)w = 0$ .**

**Analogous for  $\frac{d}{dt}x = Ax + Bw, 0 = Cx + D\left(\frac{d}{dt}\right)w, D \in \mathbb{R}[\xi]^{\bullet \times \bullet}$ .**



## Rational representations

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### Justification:

**2. Consider  $y = G(s)w$ . View  $G(s)$  as a transfer f'n.  
Take your favorite definition of input/output pairs.**

**Output nulling inputs exactly those that satisfy  $G\left(\frac{d}{dt}\right)w = 0$ .**

**3. ...**

## Rational representations

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**Note!** With this def., we can deal with transfer functions,

$$y = F\left(\frac{d}{dt}\right)u, \quad \text{i.e.} \quad \begin{bmatrix} F\left(\frac{d}{dt}\right) & \vdots & -I \end{bmatrix} \begin{bmatrix} u \\ y \end{bmatrix} = 0$$

with  $F$  a matrix of rational functions, while completely avoiding Laplace transforms, domains of convergence, and such mathematical traps.



# Caveats

**$F\left(\frac{d}{dt}\right)$  is not a map!**

**Consider**

$$y = F\left(\frac{d}{dt}\right)u$$

**We now know what it means that  $(u, y) \in \mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^\bullet)$  satisfies this ‘ODE’.**

**Is there a unique  $y$  for a given  $u$ ?**

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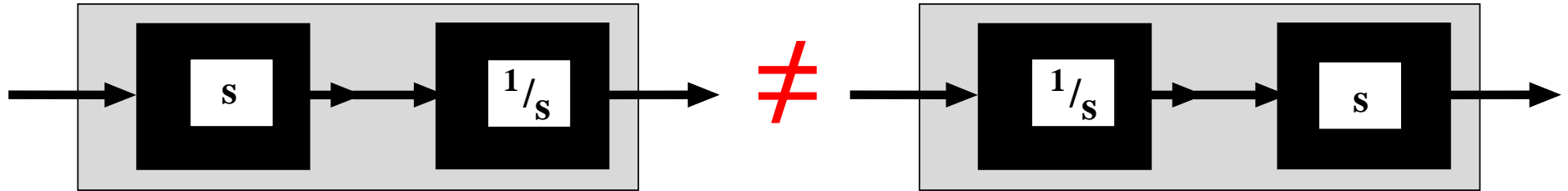
**$F = P^{-1}Q$  coprime fact.  $\Leftrightarrow P^{-1} \begin{bmatrix} P & -Q \end{bmatrix}$  coprime fact.**

$$F = P^{-1}Q \rightsquigarrow y = F\left(\frac{d}{dt}\right)u \Leftrightarrow P\left(\frac{d}{dt}\right)y = Q\left(\frac{d}{dt}\right)u$$

**If  $P \neq I$  (better, not unimodular), there are many sol’ns  $y$  of this ODE for a given  $u$ .**

$$y = y_{\text{particular}} + y_{\text{homogeneous}} \quad P\left(\frac{d}{dt}\right)y_{\text{homogeneous}} = 0$$

# $G_1\left(\frac{d}{dt}\right)$ and $G_2\left(\frac{d}{dt}\right)$ do not commute



$$G_1(s) = \frac{1}{s} \quad \text{and} \quad G_2(s) = s$$

$$y = \frac{1}{\frac{d}{dt}}v, \quad v = \frac{d}{dt}u \quad \Rightarrow \quad y(t) = u(t) + \text{constant}$$

$$y = \frac{d}{dt}v, \quad v = \frac{1}{\frac{d}{dt}}u \quad \Rightarrow \quad y(t) = u(t)$$

## Raison d'être

**LTIDSs** are **defined** in terms of **polynomial** symbols

$$R \left( \frac{d}{dt} \right) w = 0 \quad R \in \mathbb{R} [\xi]^{\bullet \times w}$$

(behavior  $\mathcal{B} :=$  the  $\mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w)$  solutions)

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**Behavior** := the set of solutions of

$$Q \left( \frac{d}{dt} \right) w = 0 \quad Q \in \mathbb{R} [\xi]^{\bullet \times w}$$

where  $G = P^{-1}Q$ ,  $P, Q \in \mathbb{R} [\xi]^{\bullet \times \bullet}$ ,  $P$  and  $Q$  left coprime

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$$G \left( \frac{d}{dt} \right) w = 0 \quad G \in \mathbb{R}(\xi)^{\bullet \times w}$$

**This added flexibility**  $\rightsquigarrow$  **better adapted to certain applications,**

e.g. distance between systems

e.g. behavioral model reduction

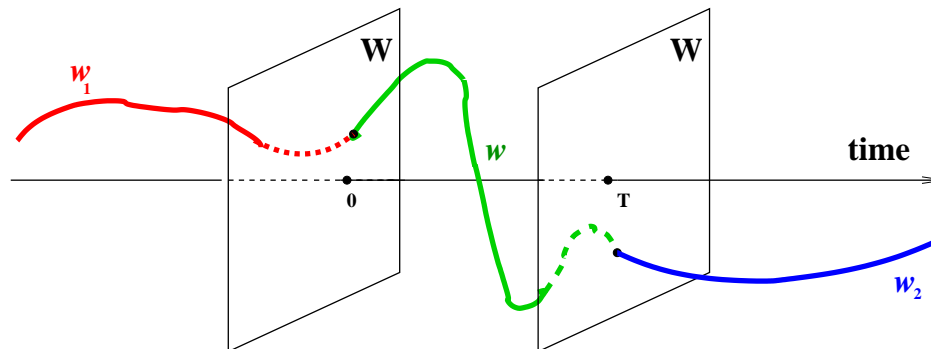
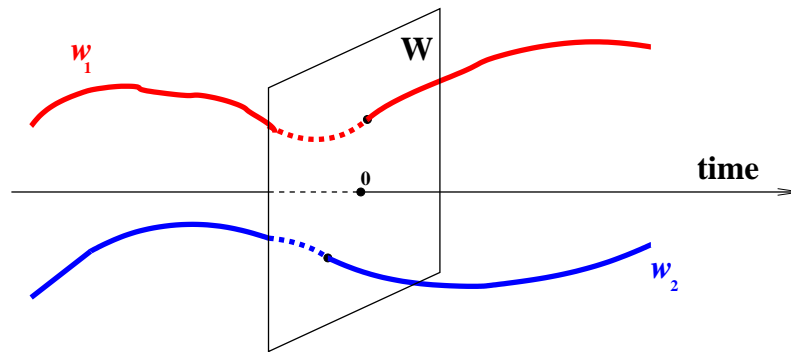
e.g. parametrization of the set of stabilizing controllers

# **Controllability c.s.**

# Controllability and stabilizability

$\mathcal{B}$  is said to be **controllable**  $:\Leftrightarrow$

$\forall w_1, w_2 \in \mathcal{B}, \exists T \geq 0$  and  $w \in \mathcal{B}$  such that ...

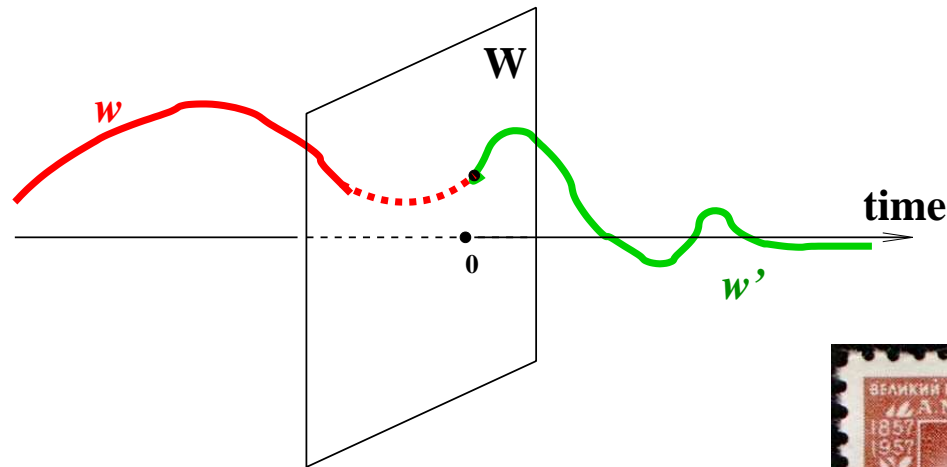


# Controllability and stabilizability

$\mathcal{B}$  is said to be **controllable**  $:\Leftrightarrow$

$\mathcal{B}$  is said to be **stabilizable**  $:\Leftrightarrow$

$\forall w \in \mathcal{B}, \exists w' \in \mathcal{B}$  such that ...



Stability in the sense of **Lyapunov**



# Representations

**What properties on  $G$  imply that the system with rational representation**

$$G \left( \frac{d}{dt} \right) w = 0$$

$$G \in \mathbb{R}(\xi)^{\bullet \times w}$$

**has any of these properties?**

# Representations

**What properties on  $G$  imply that the system with rational representation**

$$G \left( \frac{d}{dt} \right) w = 0 \quad G \in \mathbb{R}(\xi)^{\bullet \times w}$$

**has any of these properties?**

**Under what conditions on  $G$  does  $G \left( \frac{d}{dt} \right) w = 0$  define a controllable or a stabilizable system?**

**Can a rational representation be used to put one of these properties in evidence?**

## Tests

### Theorem: The LTIDS

$$G\left(\frac{d}{dt}\right)w = 0 \quad G \in \mathbb{R}(\xi)^{\bullet \times w}$$

is **controllable** if and only if

$$G(\lambda) \text{ has the same rank } \forall \lambda \in \mathbb{C}$$

Interpret carefully in cases like

$$G(s) = \begin{bmatrix} s & 0 \\ 0 & \frac{1}{s} \end{bmatrix}, G(s) = \begin{bmatrix} s \\ \frac{1}{s} \end{bmatrix}, G(s) = \begin{bmatrix} s & 1 \\ & s \end{bmatrix}$$



## Tests

### Theorem: The LTIDS

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### Theorem: The LTIDS

$$G \left( \frac{d}{dt} \right) w = 0 \quad G \in \mathbb{R}(\xi)^{\bullet \times w}$$

is **stabilizable** if and only if

$$G(\lambda) \text{ has the same rank } \forall \lambda \in \mathbb{C} \text{ with } \operatorname{realpart}(\lambda) \geq 0$$

## Image representation

For example,

**Theorem:** A LTIDS is **controllable** if and only if its behavior allows an image representation

$$w = M\left(\frac{d}{dt}\right)\ell$$

$$M \in \mathbb{R}(\xi)^{w \times \bullet}$$

## Module & vector spaces

Take a LTIDS  $\mathcal{B}$ .

$n \in \mathbb{R}(\xi)^{1 \times w}$  is an annihilator  $:\Leftrightarrow n\left(\frac{d}{dt}\right)\mathcal{B} = 0$ , i.e.,

$$n\left(\frac{d}{dt}\right)w = 0 \quad \forall w \in \mathcal{B}$$

What structure does the set of annihilators of a given  $\mathcal{B}$  have?

## Module & vector spaces

Take a LTID behavior  $\mathcal{B}$ .

$n \in \mathbb{R}[\xi]^{1 \times w}$  is a **polynomial annihilator**  $:\Leftrightarrow n\left(\frac{d}{dt}\right)\mathcal{B} = 0$

**The polynomial annihilators form a  $\mathbb{R}[\xi]$ -module:**

$n_1, n_2$  polynomial annihilators,  $p \in \mathbb{R}[\xi]$

$\Rightarrow n_1 + pn_2$  polynomial annihilator.

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$n \in \mathbb{R}(\xi)^{1 \times w}$  is a **rational** annihilator  $:\Leftrightarrow n\left(\frac{d}{dt}\right)\mathcal{B} = 0$

The rational annihilators of a **controllable**  $\mathcal{B}$  form a  $\mathbb{R}(\xi)$ -vector space:

$n_1, n_2$  rational annihilators,  $p \in \mathbb{R}(\xi)$

$\Rightarrow n_1 + pn_2$  rational annihilator.

## Module & vector spaces

By identifying a system with its polynomial annihilators, we obtain the one-to-one relation between LTIDSs with  $w$  variables and the

$\mathbb{R}[\xi]$ -submodules of  $\mathbb{R}[\xi]^w$

By identifying a system with its rational annihilators, we obtain the one-to-one relation between the **controllable** LTIDSs with  $w$  variables and the

$\mathbb{R}(\xi)$ -subspaces of  $\mathbb{R}(\xi)^w$

**LTIDS  $\cong$  finite dimensional  $\mathbb{R}[\xi]$ -modules**

**Controllable LTIDS  $\cong$  finite dimensional  $\mathbb{R}(\xi)$ -subspaces.**

# PART II

## Model reduction

## Reducing the state dimension

What is a good, computable, definition for the **distance** between two LTIDS?

Basic issue underlying model reduction, robustness, etc.

- Approximate a system by a simpler one.
- If a system has a particular property (e.g., stabilized by a controller), will this also hold for close by systems?

What is meant by ‘approximate’, by ‘close by’?



## Reducing the state dimension

There is an elegant theory for reducing the state space dimension of **stable** LTI **input/output** systems.

Let  $\mathcal{B}$  be described by

$$\frac{d}{dt}x = Ax + Bu, \quad y = Cx + Du \quad w \cong \begin{bmatrix} u \\ y \end{bmatrix}$$

with  $A$  Hurwitz ( $:\Leftrightarrow$  eigenvalues in left half plane).

There are effective methods (balancing, AAK) with good error bounds (in terms of the  $\mathcal{H}_\infty$  norm) for approximating  $\mathcal{B}$  by a (stable) system with a lower dimensional state space.



Keith Glover

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with  $A$  Hurwitz. **T'f f'n**  $F(s) = C(Is - A)^{-1}B + D$

**proper stable rational. Reduced system**

$$\frac{d}{dt}x_{\text{reduced}} = A_{\text{reduced}}x_{\text{reduced}} + B_{\text{reduced}}u, \quad y = C_{\text{reduced}}x_{\text{reduced}} + Du$$

**T'f f'n**  $F_{\text{reduced}}(s) = C_{\text{reduced}}(Is - A_{\text{reduced}})^{-1}B_{\text{reduced}} + D$

**proper stable rational. Balanced model reduction  $\Rightarrow$**

$$\|F(i\omega) - F_{\text{reduced}}(i\omega)\| \leq 2 \left( \sum_{\text{neglected Hankel SVs}} \sigma_k \right) \quad \forall \omega \in \mathbb{R}$$

## Reducing the state dimension

There is an elegant theory for reducing the state space dimension of **stable** LTI **input/output** systems.

Let  $\mathcal{B}$  be described by

$$\frac{d}{dt}\mathbf{x} = A\mathbf{x} + B\mathbf{u}, \quad \mathbf{y} = C\mathbf{x} + D\mathbf{u}$$

with  $A$  Hurwitz.

$F(s)$  proper **stable** rational  $\Rightarrow$  reducible !

Extend this to situations where we do not make a distinction between inputs and outputs, and to unstable systems!

# Distance between systems

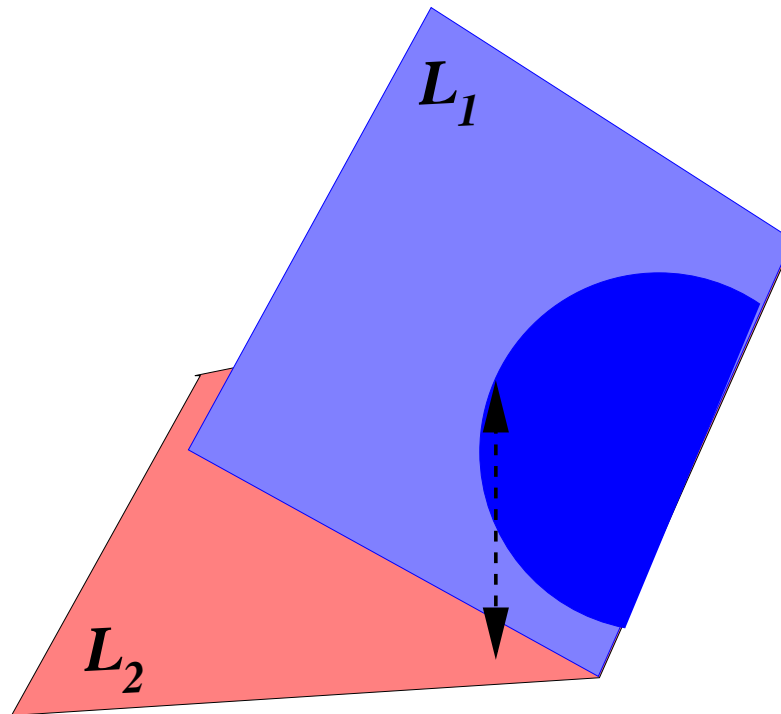
## Distance between linear subspaces

**In the behavioral theory, we identify a dynamical system with its behavior, a subspace  $\mathcal{B} \subseteq \mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w)$ . We are hence led to study the distance between linear subspaces of a vector space.**

# Linear subspaces of $\mathbb{R}^n$

$\mathcal{L}_1, \mathcal{L}_2 \subseteq \mathbb{R}^n$ , linear subspaces

$$\vec{d}(\mathcal{L}_1, \mathcal{L}_2) \cong \max_{x_1 \in \mathcal{L}_1, \|x_1\|=1} \min_{x_2 \in \mathcal{L}_2} \|x_1 - x_2\|$$



## Linear subspaces of $\mathbb{R}^n$

$\mathcal{L}_1, \mathcal{L}_2 \subseteq \mathbb{R}^n$ , **linear subspaces**

$$d(\mathcal{L}_1, \mathcal{L}_2) :=$$

$$\max \left\{ \max_{x_1 \in \mathcal{L}_1, \|x_1\|=1} \min_{x_2 \in \mathcal{L}_2} \|x_1 - x_2\|, \max_{x_2 \in \mathcal{L}_1, \|x_2\|=1} \min_{x_1 \in \mathcal{L}_1} \|x_1 - x_2\| \right\}$$

$$0 \leq d(\mathcal{L}_1, \mathcal{L}_2) \leq 1$$

$$= 1 \text{ if } \text{dimension}(\mathcal{L}_1) \neq \text{dimension}(\mathcal{L}_2)$$

## Linear subspaces of $\mathbb{R}^n$

$\mathcal{L}_1, \mathcal{L}_2 \subseteq \mathbb{R}^n$ , linear subspaces

$P_{\mathcal{L}} \perp$  projection onto  $\mathcal{L}$

$S_1, S_2$  matrices, columns orthonormal basis for  $\mathcal{L}_1, \mathcal{L}_2$

$S_1 S_1^\top, S_2 S_2^\top$  orthogonal projectors

$$\begin{aligned} d(\mathcal{L}_1, \mathcal{L}_2) &= \|P_{\mathcal{L}_1} - P_{\mathcal{L}_2}\| && \text{‘gap’, ‘aperture’} \\ &= \|S_1 S_1^\top - S_2 S_2^\top\| \\ &= \min_{\text{matrices } U} \|S_1 - S_2 U\| \\ &= \min_{U \text{ such that } U \mathcal{L}_1 = \mathcal{L}_2} \|I - U\| \end{aligned}$$



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**Note**

$$d(\mathcal{L}_1, \mathcal{L}_2) = \|S_1 S_1^\top - S_2 S_2^\top\| \leq \|S_1 - S_2\|$$

## Distance between controllable behaviors

$\min \rightarrow \inf, \max \rightarrow \sup$ , etc., readily generalized to closed subspaces of Hilbert space.

For LTIDS, behaviors  $\mathcal{B} \mapsto \mathcal{B} \cap \mathcal{L}_2(\mathbb{R}, \mathbb{R}^w)$ . Keep notation. So, we consider only  $\mathcal{L}_2$ -behavior for measuring distance.

$$d(\mathcal{B}_1, \mathcal{B}_2) := \text{gap}(\mathcal{B}_1, \mathcal{B}_2)$$

$$\forall w_1 \in \mathcal{B}_1, \exists w_2 \in \mathcal{B}_2 \text{ such that } \|w_1 - w_2\| \leq \text{gap}(\mathcal{B}_1, \mathcal{B}_2) \|w_1\|$$

and vice-versa. Small gap  $\Rightarrow$  the models are ‘close’.

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- How to compute the gap?
- Model reduce according to the gap!

## Norm-preserving representations

Let  $\mathcal{B}$  be the behavior of a controllable LTIDS. Then it allows a rational symbol based image representation

$$w = M\left(\frac{d}{dt}\right)\ell \quad \text{with } M \in \mathbb{R}(\xi)^{w \times \bullet} \quad \& \quad M(-\xi)^\top M(\xi) = I$$

**i.e.,**  $\|\ell\|_{\mathcal{L}_2(\mathbb{R}, \mathbb{R}^\bullet)}^2 = \|w\|_{\mathcal{L}_2(\mathbb{R}, \mathbb{R}^w)}^2$  ‘**norm preserving image repr.**’

$$\int_{-\infty}^{+\infty} \|w(t)\|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{+\infty} \|\hat{w}(i\omega)\|^2 d\omega =$$
$$\frac{1}{2\pi} \int_{-\infty}^{+\infty} \|M(i\omega)\hat{\ell}(i\omega)\|^2 d\omega = \frac{1}{2\pi} \int_{-\infty}^{+\infty} \|\hat{\ell}(i\omega)\|^2 d\omega = \int_{-\infty}^{+\infty} \|\ell(t)\|^2 dt$$

**Note:**  $M$  cannot be polynomial, it must be rational.

Obviously  $M$  must be proper. Can also make it stable.

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**Note:**  $M$  cannot be polynomial, it must be rational.

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**Proof:** Start with an observable polynomial image

representation  $w = N\left(\frac{d}{dt}\right)\ell, N \in \mathbb{R}[\xi]^{w \times m(\mathcal{B})}$ . **Factor**

$$N(-\xi)^\top N(\xi) = F(-\xi)^\top F(\xi), F \in \mathbb{R}[\xi]^{m(\mathcal{B}) \times m(\mathcal{B})}$$

Can make determinant( $F$ ) Hurwitz. Take  $M = NF^{-1}$ .

## Norm-preserving representations

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**Note that**

$$f \in \mathcal{L}_2(\mathbb{R}, \mathbb{R}^w) \mapsto M(i\omega)M(-i\omega)^\top \hat{f}(i\omega)$$

**is the orthogonal projection onto**  $\mathcal{B} \cap \mathcal{L}_2(\mathbb{R}, \mathbb{R}^w)$ .

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**Note:**  $M$  cannot be polynomial, it must be rational.

Obviously  $M$  must be proper. Can also make it stable.

$\mathcal{B}_1 \mapsto M_1, \mathcal{B}_2 \mapsto M_2$  norm preserving, then

$$\begin{aligned} \text{gap}(\mathcal{B}_1, \mathcal{B}_2) &= \|M_1(i\omega)M_1(-i\omega)^\top - M_2(i\omega)M_2(-i\omega)^\top\|_{\mathcal{L}_\infty} \\ &\leq \|M_1(i\omega) - M_2(i\omega)\|_{\mathcal{H}_\infty} \end{aligned}$$

## Model reduction by balancing

Start with  $\mathcal{B}$ . Take representation

$$w = M\left(\frac{d}{dt}\right)\ell \quad \text{with } M \in \mathbb{R}(\xi)^{w \times \bullet} \quad \text{norm preserving, stable}$$

Now model reduce  $w = M\left(\frac{d}{dt}\right)\ell$  (viewed as a stable input/output system) using, for example, balancing

$$\rightsquigarrow w = M_{\text{reduced}}\left(\frac{d}{dt}\right)\ell$$

and an error bound

$$\|M - M_{\text{reduced}}\|_{\mathcal{H}_\infty} \leq 2 \left( \sum_{\text{neglected SVs of } M} \sigma_k \right)$$



## Behavioral error bound

Start with stable norm preserving representation of  $\mathcal{B}$

$$w = M\left(\frac{d}{dt}\right)\ell \quad \text{with } M \in \mathbb{R}(\xi)^{w \times \bullet}$$

Model reduce using balancing  $\rightsquigarrow w = M_{\text{reduced}}\left(\frac{d}{dt}\right)\ell$ .

Call behavior  $\mathcal{B}_{\text{reduced}}$ . Error bound

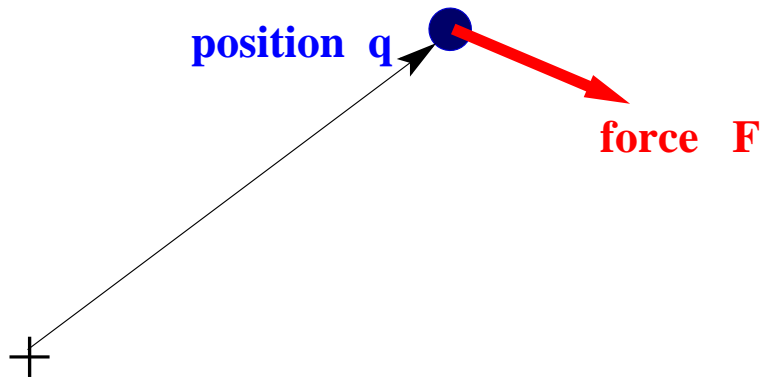
$$\begin{aligned} \text{gap}(\mathcal{B}, \mathcal{B}_{\text{reduced}}) &= \|MM^{\top} - M_{\text{reduced}}M_{\text{reduced}}^{\top}\|_{\mathcal{L}_{\infty}} \\ &\leq \|M - M_{\text{reduced}}\|_{\mathcal{H}_{\infty}} \\ &\leq 2 \left( \sum_{\text{neglected SVs of } M} \sigma_k \right) \end{aligned}$$

$\forall w \in \mathcal{B} \exists w' \in \mathcal{B}_{\text{red}}$  such that  $\|w - w'\| \leq 2(\sum_{\text{neglected SVs}} \sigma_k) \|w\|$

and vice-versa.

$\sum_{\text{neglected SVs}} \sigma_k$  small  $\Rightarrow$  good approximation in the gap.

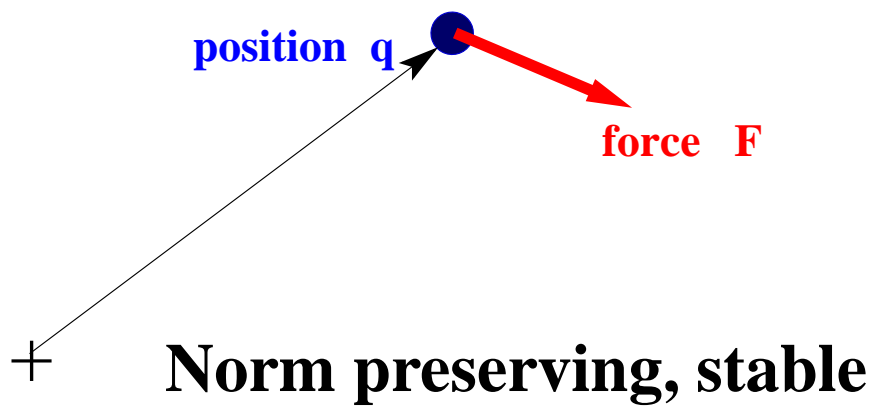
# Example



$$F = \frac{d^2}{dt^2} q,$$

$$w = \begin{bmatrix} F \\ q \end{bmatrix}$$

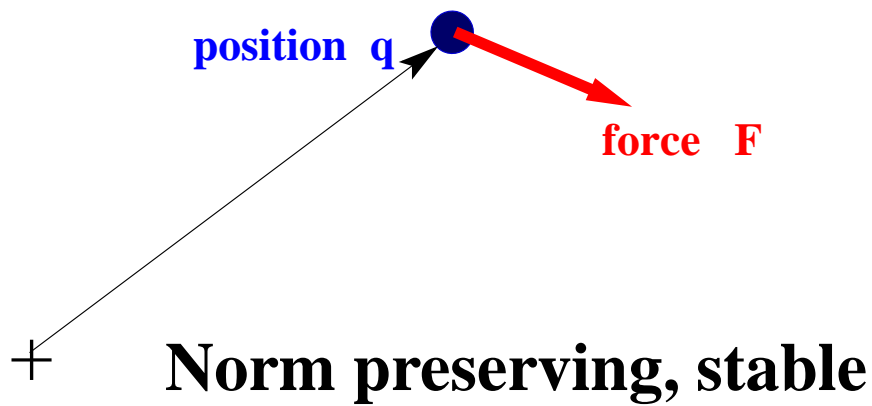
# Example



$$F = \frac{d^2}{dt^2} q, \quad w = \begin{bmatrix} F \\ q \end{bmatrix}$$

$$\begin{bmatrix} F \\ q \end{bmatrix} \cong \begin{bmatrix} \frac{\xi^2}{\xi^2 + \sqrt{2}\xi + 1} \\ \frac{1}{\xi^2 + \sqrt{2}\xi + 1} \end{bmatrix} \ell$$

# Example



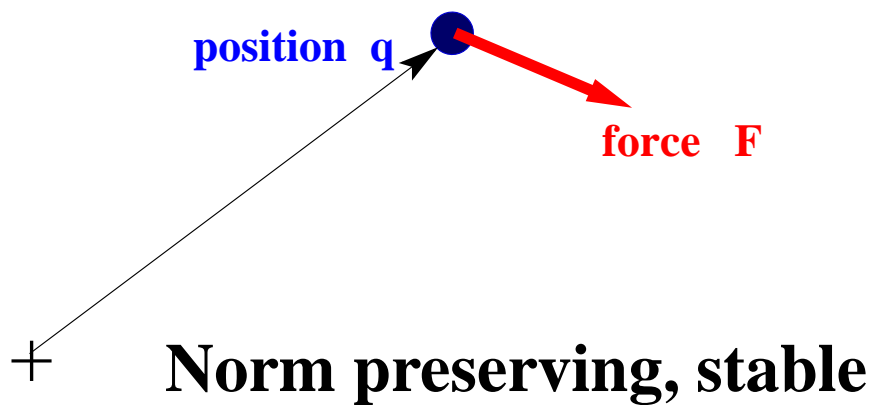
$$F = \frac{d^2}{dt^2} q, \quad w = \begin{bmatrix} F \\ q \end{bmatrix}$$

$$\begin{bmatrix} F \\ q \end{bmatrix} \approx \begin{bmatrix} \frac{\xi^2}{\xi^2 + \sqrt{2}\xi + 1} \\ \frac{1}{\xi^2 + \sqrt{2}\xi + 1} \end{bmatrix} \ell$$

reduced model

$$\begin{bmatrix} F \\ q \end{bmatrix} \approx \begin{bmatrix} \frac{\xi - \frac{1}{2}}{\xi + \frac{1}{\sqrt{2}}} \\ \frac{\frac{1}{2}}{\xi + \frac{1}{\sqrt{2}}} \end{bmatrix} \ell$$

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$$F = \frac{d^2}{dt^2} q$$

first order approximation

$$\frac{1}{2} F = \frac{d}{dt} q - \frac{1}{2} q$$

# Summary

## Conclusions

- $G(\frac{d}{dt})w = 0$  defined in terms left-coprime factorization of rational  $G$ .

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- Stable norm preserving representation  $w = M(\frac{d}{dt})\ell$  leads to model reduction of unstable systems and systems without input/output partition.

# PART III

## Parametrization of stabilizing controllers

# $\mathbb{R}(\xi)$ and some of its subrings

**Field of (real) rationals**

**Subrings of interest**

**polynomials**

**proper rationals**

**stable rationals**

**proper stable rationals**

## Relevant rings

**unimodularity**  $:\Leftrightarrow$  invertibility in the ring

Field of (real) rationals      **nonzero**

Subrings of interest

polynomials      **nonzero constant**

proper rationals      **biproper**

stable rationals      **miniphase**

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unimodularity of square matrices over rings

$\Leftrightarrow$  determinant **unimodular**

left primeness of matrices over rings

$:\Leftrightarrow \left[ \left[ M = FM' \right] \Rightarrow \left[ F \text{ unimodular} \right] \right]$

## Representability

The LTIDS  $\mathcal{B}$  admits a representation that is **left prime** over

- **rationals: always**
- **proper rationals: always**
- **stable rationals: iff  $\mathcal{B}$  is stabilizable**
- **proper stable rationals: iff  $\mathcal{B}$  is stabilizable**
- **polynomials: iff  $\mathcal{B}$  is controllable**

**Left prime representations over subrings allow to express certain system properties...**

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- **stable rationals: iff  $\mathcal{B}$  is stabilizable**
- **proper stable rationals: iff  $\mathcal{B}$  is stabilizable**

$\mathcal{B}$  **stabilizable**  $\Leftrightarrow \exists G$ , matrix of rational functions, such that

- $\mathcal{B} = \text{kernel} \left( G \left( \frac{d}{dt} \right) \right)$
- $G$  is proper (no poles at  $\infty$ )
- $G^\infty := \lim_{\lambda \rightarrow \infty} G(\lambda)$  has full row rank (no zeros at  $\infty$ )
- $G$  has no poles in  $\mathbb{C}_+ := \{\lambda \in \mathbb{C} \mid \text{real}(\lambda) \geq 0\}$
- $G(\lambda)$  has full row rank  $\forall \lambda \in \mathbb{C}_+$  (no zeros in  $\mathbb{C}_+$ )

# Unimodular completion

## Unimodular completion lemma

Let  $G$  be a matrix over one of our rings  
(polynomial, proper rat., stable rat., proper stable rat.).

¿ Does there exist a **unimodular completion**  $G'$   
i.e. a matrix  $G'$  over that same ring such that

$$\begin{bmatrix} G \\ G' \end{bmatrix}$$

is unimodular (determinant is invertible in the ring) ?

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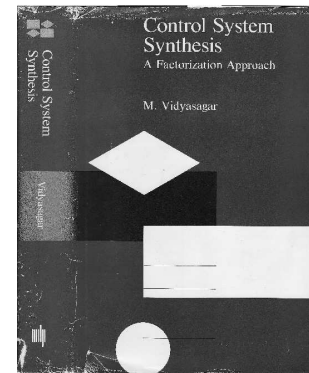
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**if and only if**

$G$  is **left prime** over the ring !



M. Vidyasagar

# Unimodular completion lemma

**$G$ : 1 row, 2 columns**

$$G = \begin{bmatrix} p & q \end{bmatrix} \quad G' = \begin{bmatrix} -y & x \end{bmatrix} \quad \begin{bmatrix} G \\ G' \end{bmatrix} = \begin{bmatrix} p & q \\ -y & x \end{bmatrix}$$

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determinant =  $px + qy$ ,      **unimodularity**  $\Leftrightarrow$   $px + qy = 1$

**solvable for  $x, y \Leftrightarrow p$  &  $q$  coprime  $\Leftrightarrow G = \begin{bmatrix} p & q \end{bmatrix}$  left prime**



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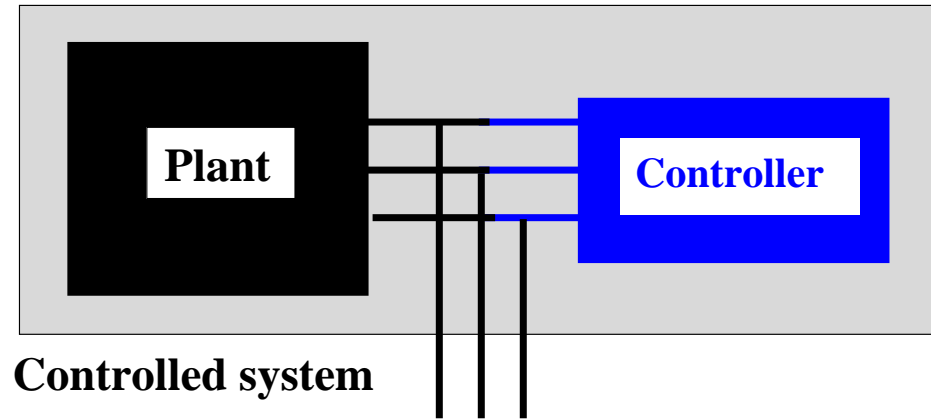
Our rings are **Hermite rings**



$G$  left prime  $\Leftrightarrow$  unimodularly completable  $\Leftrightarrow \exists H : GH = I \Leftrightarrow \dots$

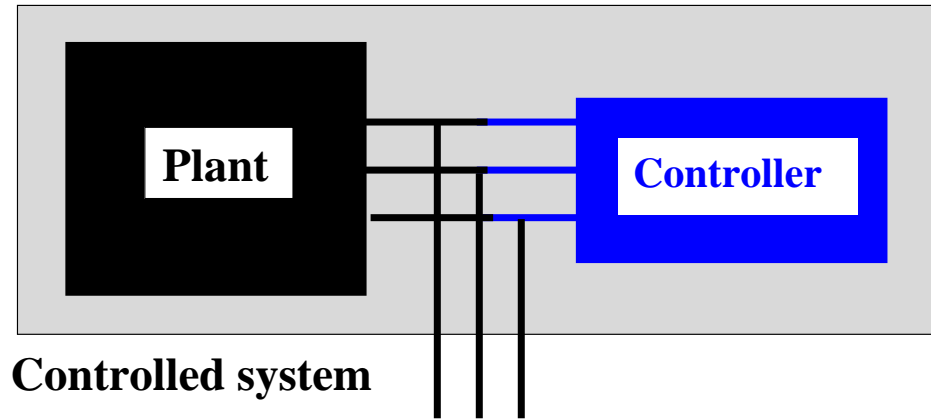
# Control

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Plant  $\mathcal{P}$ , controller  $\mathcal{C}$ , controlled system  $\mathcal{P} \cap \mathcal{C}$

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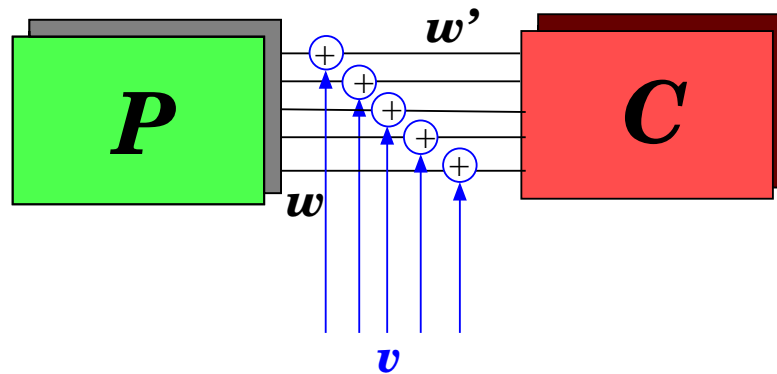
$\llbracket \mathcal{C} \text{ is stabilizing} \rrbracket \Leftrightarrow \llbracket \mathcal{P} \cap \mathcal{C} \text{ is stable} \rrbracket$

$\Leftrightarrow \llbracket \llbracket w \in \mathcal{P} \cap \mathcal{C} \rrbracket \Rightarrow \llbracket w(t) \rightarrow 0 \text{ for } t \rightarrow \infty \rrbracket \rrbracket$

# Control

[[ $\mathcal{C}$  is a **regular controller**]]  $:\Leftrightarrow$  [[ $\mathcal{P} + \mathcal{C} = \mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w)$ ]]

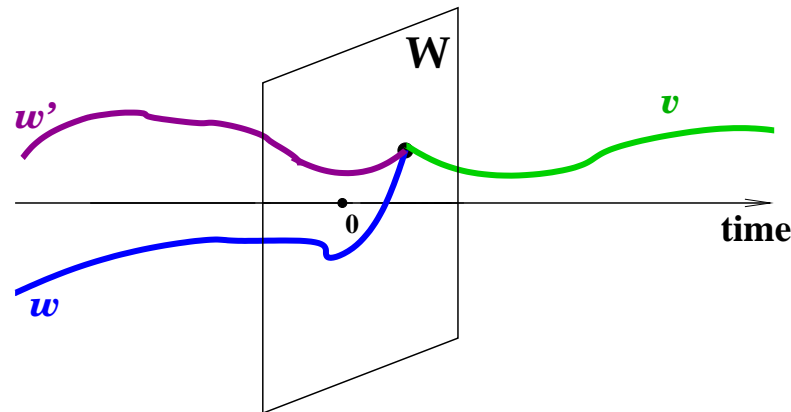
$\forall v \in \mathcal{C}^\infty(\mathbb{R}, \mathbb{R}^w) \exists w \in \mathcal{P}$  and  $w' \in \mathcal{C}$  such that  $v = w + w'$



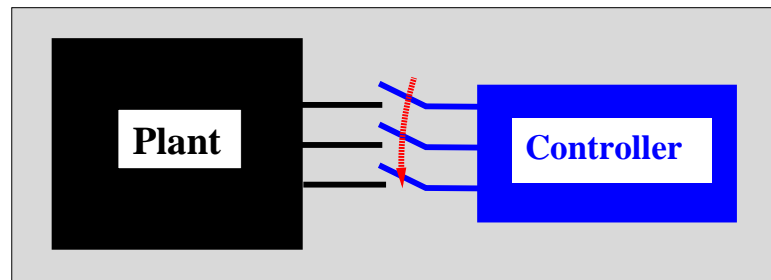
# Control

[[ $\mathcal{C}$  is a **superregular** controller]]  $:\Leftrightarrow$  in addition,

[[ $\forall w \in \mathcal{P}, \forall w' \in \mathcal{C} \exists v$  such that  $w \wedge_0 v, w' \wedge_0 v \in \mathcal{P} \cap \mathcal{C}$ ]]



**A superregular controller can be engaged at any time**



Controlled system

**superregular  $\Rightarrow$  controller can be engaged at any time**

## **(Super)regular controllers**

**Usual feedback controllers are superregular**

**PID controllers are regular, but not superregular**

**Controllers that are not superregular are relevant:**

**control is interconnection**, not just signal processing



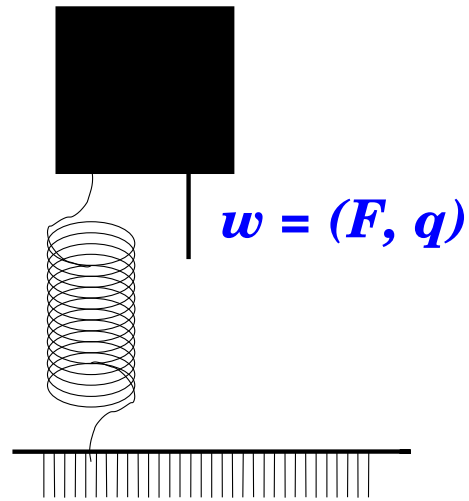
**Harry Trentelman**



**Madhu Belur**

# A regular, but not superregular, controller

**Plant:**

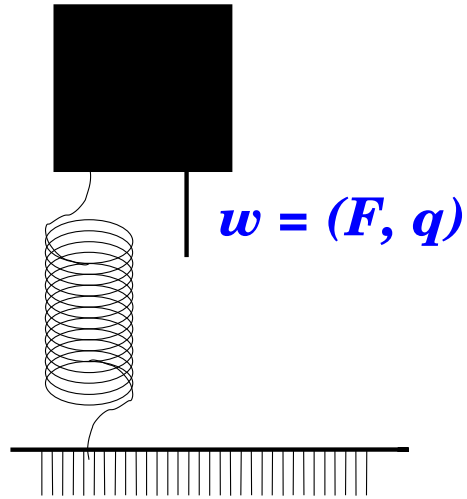


$$M \frac{d^2}{dt^2} q + Kq = F, \quad w = (F, q)$$



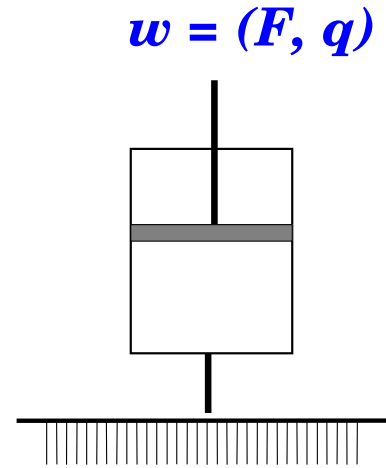
# A regular, but not superregular, controller

**Plant:**



$$M \frac{d^2}{dt^2} q + Kq = F, \quad w = (F, q)$$

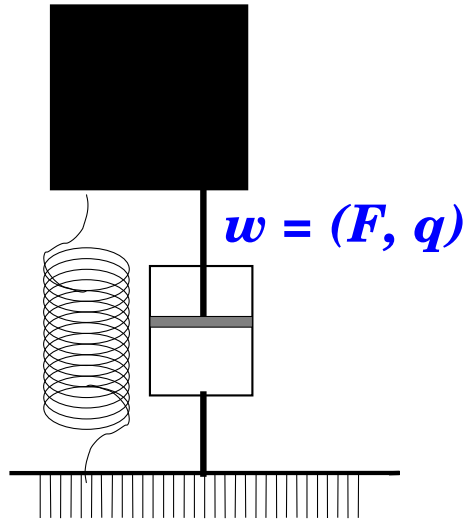
**Controller:**



$$F = -D \frac{d}{dt} q$$

# A regular, but not superregular, controller

**Controlled system:**



$$M \frac{d^2}{dt^2} q + D \frac{d}{dt} q + K q = 0, \quad F = -D \frac{d}{dt} q$$

# **Existence of stabilizing controllers**

# Existence

## Proposition

$\mathcal{P}$  is stabilizable  $\Leftrightarrow \exists$  a regular stabilizing controller

$\Leftrightarrow \exists$  a superregular stabilizing controller

# Existence

## Proposition

$\mathcal{P}$  is stabilizable  $\Leftrightarrow \exists$  a regular stabilizing controller  
 $\Leftrightarrow \exists$  a superregular stabilizing controller

$\mathcal{P}$  is controllable  $\Leftrightarrow \exists$  pole placement for  $\mathcal{P} \cap \mathcal{C}$

$\nexists$  a controller that is superregular  
&  $\mathcal{P} \cap \mathcal{C}$  has a low order characteristic polynomial.

# Parametrization of stabilizing controllers

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Start with  $G \left( \frac{d}{dt} \right) w = 0$  a (rational symbol based) representation of the plant

Assume  **$G$  left prime** over proper stable rational functions.  
Iff the plant is stabilizable, such a  $G$  exists.

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Par'ion of superregular stabilizing controllers  $C \left( \frac{d}{dt} \right) w = 0$

$$C = F_1 G + F_2 G'$$

$F_1$  **free** over ring of proper stable rational

$F_2$  **unimodular** over proper stable rational

**So**

Using **rational symbol** based representations  $G \left( \frac{d}{dt} \right) w = 0$  that are left prime over suitable rings, we obtain parametrizations of regular and superregular stabilizing controllers

$\cong$  Kučera-Youla parametrization, with proper attention for the uncontrollable part



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