

THE GEOMETRIC APPROACH TO CONTROL SYSTEM DESIGN:
A TUTORIAL INTRODUCTION TO A FEW OF THE MAIN IDEAS.

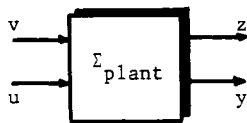
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In this talk we will discuss in a tutorial fashion some of the main applications of the geometric approach to control system design. We will briefly review the notions of controlled invariant, conditionally invariant, and controllability subspaces. Further, their 'almost' counterparts and the notion of (A,B,C)-pairs. We will show how these concepts yield the solution of the disturbance decoupling and of the non-interacting control problem.

1. Control System Design

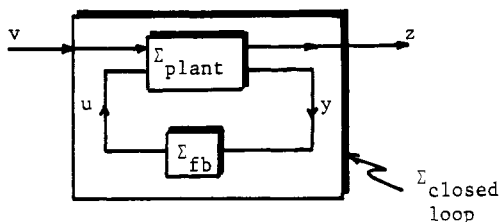
A typical control system Σ_{plant} has four types of external variables: exogenous inputs v , control inputs u , to-be-controlled outputs z , and observed outputs y . This structure is shown in the block diagram



A controller is a system Σ_{feedback} having the observed output space as its input space and the control input space as its output space:



Feedback connection of Σ_{plant} with Σ_{feedback} yields the closed loop system $\Sigma_{\text{closed loop}}$



having v as input and z as output.

A control system design problem consist in the question to design, for a given plant Σ_{plant} and a given design purpose, a feedback controller Σ_{feedback} such that the closed loop system $\Sigma_{\text{closed loop}}$ meets the specified design requirement. For example, in the disturbance decoupling problem we assume that r is a

disturbance and that in $\Sigma_{\text{closed loop}}$ z should be independent of r ; in non-interacting control, we assume that

$$v = \begin{bmatrix} v_1 \\ v_2 \\ \vdots \\ v_k \end{bmatrix} \text{ and } z = \begin{bmatrix} z_1 \\ z_2 \\ \vdots \\ z_k \end{bmatrix}$$

and we want that in $\Sigma_{\text{closed loop}}$, for each $j = 1, 2, \dots, k$, z_j is independent of v_i for $i \neq j$. We can also add stability or pole placement conditions to these design requirements.

2. Non-Interacting Control for Linear Systems

Consider the linear system

$$\dot{x} = Ax + Bu + \sum_{i=1}^k G_i v_i$$

Σ_{plant} :

$$y = Cx; z_i = H_i x$$

Σ_{feedback} :

$$\dot{w} = Kw + Ly; u = Mw + Fy$$

yielding, in the obvious notation,

$$\Sigma_{\text{closed loop}}: \dot{x}_e = A^e x_e + \sum_{i=1}^k G_i^e v_i; z_i = H_i^e x_e$$

The purpose is to design, for a given Σ_{plant} , Σ_{feedback} , such that in $\Sigma_{\text{closed loop}}$ the transfer function is block diagonal.

We will show that the solution of this problem (with $y = x$) is a nice application of the notions of controlled invariant and of controllability subspaces.

3. Disturbance Decoupling

The disturbance decoupling problems for the linear case is specified by

$$\Sigma_{\text{plant}}: \dot{x} = Ax + Bu + Gv; y = Cx; z = Hx$$

$$\Sigma_{\text{feedback}}: \dot{w} = Kw + Ly; u = Mw + Fy$$

$$\Sigma_{\text{closed loop}}: \dot{x}^e = A^e x^e + G^e v; z = H^e x^e$$

In disturbance decoupling we want that the transfer function of Σ_{closed} is zero.

loop

In almost disturbance decoupling we want that its transfer function has a norm which is less than any presigned $\epsilon > 0$ (of course, Σ_{fb} is allowed to depend

on ϵ in this case). We shall review the solution to these problems. As we will see, solution of the disturbance decoupling problem is an illustration of Schumacher's (A,B,C)-pairs and the solution of the almost disturbance decoupling problem is an illustration of the notion of almost invariant subspaces.

We will also mention a few open problems in this area.