DSP-CIS

Part-III : Optimal & Adaptive Filters

Chapter-9 : Square Root and Fast RLS Algorithms

Marc Moonen

Dept. E.E./ESAT-STADIUS, KU Leuven marc.moonen@kuleuven.be www.esat.kuleuven.be/stadius/









Recap 4/5	
2.1 Standard RLS	
It is observed that $\aleph_{uu}[k] = \aleph_{uu}[k-1] + \mathbf{u}_k \cdot \mathbf{u}_k^T$ (and $\aleph_{du}[k] = \aleph_{du}[k-1] + \mathbf{u}_k \cdot d_k$) The matrix inversion lemma states that $\aleph_{uu}[k]^{-1} = \aleph_{uu}[k-1]^{-1} - (\frac{1}{1 + \mathbf{u}_k^T \aleph_{uu}[k-1]^{-1}\mathbf{u}_k}) \cdot \mathbf{k}_k \mathbf{k}_k^T$ with $\mathbf{k}_k = \aleph_{uu}[k-1]^{-1}\mathbf{u}_k$	
$\mathbf{w}_{LS}[k] = \mathbf{w}_{LS}[k-1] + \underbrace{\frac{Kalman gain vector'}{K_{uu}[k]^{-1}\mathbf{u}_{k}}}_{=(\frac{1}{1+u_{k}^{T}N_{uu}[k-1]^{-1}\mathbf{u}_{k}}),\mathbf{k}_{k}}^{'a priori residual'} \cdot \underbrace{(d_{k} - \mathbf{u}_{k}^{T}\mathbf{w}_{LS}[k-1])}_{(d_{k} - \mathbf{u}_{k}^{T}\mathbf{w}_{LS}[k-1])}$	
= standard recursive least squares (RLS) algorithm	
Remark : $O(L^2)$ instead of $O(L^2)$ operations per time update	
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Square Root RLS Algorithms

Standard RLS algorithm has been shown to have unstable quantization error propagation (in lowprecision implementation)

i.e. when an infinite precision version is run next to a finite precision version (both fed with the same input signals), then after xx iterations the finite precision version produces results (far) away from the infinite precision results

Better ('square root') algorithms are based on orthogonal transformations and 'QR decomposition'

Starting point is again least squares (LS) estimation...

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Square Root RLS Algorithms

A graphical representation (i.e. a 'realization') of the QRD updating process is presented in the next slide

This is also referred to as a 'signal flow graph' (SFG)

The SFG in the next slide will be further developed in later slides, and also used explicitly for the (graphical) derivation of a 'fast' RLS algorithm (p.27)

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Square Root RLS Algorithms

Residual extraction

$$\begin{bmatrix} R[k] & \mathbf{z}[k] \\ 0 \cdots 0 & \varepsilon \end{bmatrix} = Q[k]^T \cdot \begin{bmatrix} R[k-1] & \mathbf{z}[k-1] \\ \mathbf{u}_k^T & d_k \end{bmatrix}$$

From this it is proved that the 'a posteriori residual' is

From this it is proved that the 'a posteriori residual' is

$$_{k} - \mathbf{u}_{k}^{T}\mathbf{w}_{LS}[k] = \varepsilon \cdot \prod_{i=1}^{L+1} \cos(\theta_{i})$$

and the 'a priori residual' is

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d

$$d_{k} - \mathbf{u}_{k}^{T} \mathbf{w}_{LS}[k-1] = \frac{\varepsilon}{\prod_{i=1}^{L+1} \cos(\theta_{i})}$$
 Hence ε is geometric mean of a posteriori & a priori residual

Residual extraction (v1.0): u[k] u[k-1] u[k-2] u[k-3] d[k] rv cell (delay) 0 $\Pi \,\cos\,\theta$ $d_k - \mathbf{u}_k^T \mathbf{w}_{LS}[k] = \varepsilon \cdot$ $\cos(\theta_i)$ output DSP-CIS 2019-2020 / Chapter-9: Square Root and Fast RLS Algorithms









Fast RLS Algorithms

Preliminaries

- vast literature available on fast least squares algorithms
- the derivation of fast algorithms is *highly* mathematical (see page 32)
- we show how fast (QRD-based) algorithms can be derived using signal flow graph (SFG) manipulation
- In doing so we provide additional insight to the algorithmic structure

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Fast RLS Algorithms 9.2.1. QRD-based Least Squares Lattice algo START INITIALISE (all variables) := 0: FOR n FROM 1 DO LET $\alpha_{f,0}(n) := x(n); \ \alpha_{b,0}(n-1) := x(n-1); \ \alpha_0(n-1) := y(n-1); \ \gamma_0(n-1) := 1;$ Example (headache?) FOR q FROM 1 TO p DO LET $\epsilon_{b,q-1}(n-1) := \sqrt{(\beta \epsilon_{b,q-1}(n-2))^2 + |\alpha_{b,q-1}(n-1)|^2};$
$$\begin{split} & \text{B}_{q-1}(n-1) = 0 \text{ THEN LET } c_{f,q} := 1; \ s_{f,q} := 0 \\ & \text{ELSE LET } c_{f,q} := \beta \epsilon_{b,q-1}(n-2) / \epsilon_{b,q-1}(n-1); \ s_{f,q} := \alpha_{b,q-1}(n-1) / \epsilon_{b,q-1}(n-1) \end{split}$$
See p.39 for a signal flow graph of this END_IF; $\text{LET} \; \mu_{f,q \cdot 1}(n) \coloneqq c_{f,q} \; \beta \mu_{f,q \cdot 1}(n \cdot 1) + \; s_{f,q}^* \; \alpha_{f,q \cdot 1}(n); \\$ $\alpha_{f,q}(n) := c_{f,q} \alpha_{f,q-1}(n) - s_{f,q} \beta \mu_{f,q-1}(n-1);$ $\mu_{q-1}(n-1):=c_{f,q}\,\beta\mu_{q-1}(n-2)+s_{f,q}^*\,\alpha_{q-1}(n-1);$ $\alpha_q(n-1) := c_{f,q} \alpha_{q-1}(n-1) - s_{f,q} \beta \mu_{q-1}(n-2);$ $\gamma_q(n-1):=c_{\mathbf{f},q}\,\gamma_{q-1}(n-1);$ al $e_{f,p}(n,n) = \gamma_q(n-1) \alpha_{f,q}(n)$ COMMENT COMMENT p $e_p(n-1,n-1) = \gamma_q(n-1) \alpha_q(n-1)$ COMMENT q-th order filtered residual COMMENT $\text{LET} \, \boldsymbol{\epsilon}_{f,q-1}(n) \coloneqq \sqrt{ \left(\beta \boldsymbol{\epsilon}_{f,\,q-1}(n-1) \right)^2 + \left| \boldsymbol{\alpha}_{f,\,q-1}(n) \right|^2 } \, ; \label{eq:left_left}$ IF $\varepsilon_{f,q-1}(n) = 0$ THEN LET $c_{b,q} := 1$; $s_{b,q} := 1$ $\text{ELSE LET } c_{b,q} \coloneqq \beta \epsilon_{f,q-1}(n-1) \ / \ \epsilon_{f,q-1}(n) \ ; \ s_{b,q} \coloneqq \alpha_{f,q-1}(n) \ / \ \epsilon_{f,q-1}(n)$ END_IF; $\text{LET} \ \mu_{b,q-1}(n\text{-}1) := c_{b,q} \ \beta \mu_{b,q-1}(n\text{-}2) + \ s_{b,q}^* \ \alpha_{b,q-1}(n\text{-}1);$ $\alpha_{b,q}(n):=c_{b,q}\,\alpha_{b,q-1}(n{-}1)-s_{b,q}\,\beta\mu_{b,q-1}(n{-}2);$ COMMENT $\gamma_q(n) := c_{b,q} \gamma_{q-1}(n-1);$ backward prediction re- $(n) := \gamma_0(n) \alpha_{h,0}(n) COMMENT$ END_DO END_DO FINISH DSP-CIS 2019-2020 / Chapter-9: Square Root and Fast RLS Algorithms 30 / 40

Fast RLS Algorithms

Preliminaries

- LS residuals are not changed after a permutation of the input signals (see page 32)
- This allows for a compact notation (ε-notation) for all intermediate signals in the SFG :

Every intermediate signal corresponds to the ' ε -signal' of an embedded LS problem.

Then in the ε -notation, the superscript refers to the (timeindex of the) 'right-hand side signal' of this LS problem, the subscript (Matlab-like notation) refers to the (time indices of the) *set* of 'left-hand side signals' of this LS problem (see page 33)

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