DSP-CIS

Part-III: Optimal & Adaptive Filters

Chapter-10: Kalman Filters

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Part-III: Optimal & Adaptive Filters **Chapter-7 Optimal Filters - Wiener Filters** • Introduction : General Set-Up & Applications Wiener Filters **Adaptive Filters - LMS & RLS Chapter-8** • Least Means Squares (LMS) Algorithm · Recursive Least Squares (RLS) Algorithm **Square Root & Fast RLS Algorithms Chapter-9** Square Root Algorithms Fast Algorithms **Chapter-10** Kalman Filters • Introduction – Least Squares Parameter Estimation Kalman Filter Basics Kalman Filter Algorithms

Introduction: Least Squares Parameter Estimation

In Chapter-9, have introduced 'Least Squares' estimation as an alternative (=based on observed data/signal samples) to optimal filter design (=based on statistical information)...

> $\mathrm{filter\ input\ sequence}:\ \boldsymbol{u}_1,\boldsymbol{u}_2,\boldsymbol{u}_3,\ldots\ \boldsymbol{u}_k$ corresponding desired response sequence is : $d_1, d_2, d_3, \ldots, d_k$

$$\begin{bmatrix} e_1 \\ e_2 \\ \vdots \\ e_k \end{bmatrix} = \begin{bmatrix} d_1 \\ d_2 \\ \vdots \\ d_k \end{bmatrix} - \begin{bmatrix} \mathbf{u}_1^T \\ \mathbf{u}_2^T \\ \vdots \\ \mathbf{u}_k^T \end{bmatrix} \cdot \begin{bmatrix} w_0 \\ w_1 \\ \vdots \\ w_L \end{bmatrix}$$
error signal \mathbf{e}

cost function
$$J_{LS}(\mathbf{w}) = \sum_{l=1}^{k} e_l^2 = \|\mathbf{e}\|_2^2 = \|\mathbf{d} - U\mathbf{w}\|_2^2$$

 $\rightarrow linear\ least\ squares\ problem: \min_{\mathbf{w}} \|\mathbf{d} - U\mathbf{w}\|_2^2$

$$\mathbf{w}_{LS} = \aleph_{uu}^{-1} \cdot \aleph_{du} = \left[U^T U \right]^{-1} \cdot U^T \mathbf{d}$$

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Introduction: Least Squares Parameter Estimation

'Least Squares' approach is also used in parameter estimation in a linear regression model, where the problem statement is as follows...

Given...

k vectors of input variables (='regressors')

 $\mathbf{u}_{1}, \mathbf{u}_{2}, \mathbf{u}_{3}, ... \mathbf{u}_{k}$

k corresponding observations of a dependent variable $d_1, d_2, d_3, ...d_k$ and assume a

 $\mathbf{d} = \mathbf{U}.\mathbf{w}^0 + \mathbf{e}$ linear regression/observation model =

where \mathbf{w}^0 is an unknown parameter vector (='regression coefficients') and e is unknown additive observation/measurement noise

(see p.3 for definition of **U** and **d**)

Then the aim is to estimate wo

Q: Can Least Squares (LS) estimate be reused here?

$$\mathbf{w}_{LS} = \aleph_{uu}^{-1} \cdot \aleph_{du} = \left[U^T U \right]^{-1} \cdot U^T \mathbf{d}$$

Introduction: Least Squares Parameter Estimation

$$\mathbf{w}_{LS} = \aleph_{uu}^{-1} \cdot \aleph_{du} = \left[U^T U \right]^{-1} \cdot U^T \mathbf{d}$$

If the input variables u_i are given/fixed (*) and the additive noise e is a random vector with zero-mean E{e} = 0 then the LS estimate is 'unbiased' i.e.

$$E\{\mathbf{w}_{LS}\} = E\{(U^T U)^{-1} U^T \mathbf{d}\} = E\{(U^T U)^{-1} U^T (U \cdot \mathbf{w}^0 + \mathbf{e})\} = \mathbf{w}^0 + E\{(U^T U)^{-1} U^T \mathbf{e}\}$$
$$= \mathbf{w}^0$$

• If in addition the noise **e** has <u>unit covariance matrix</u> $E\{e.e^T\} = I$ then the (estimation) <u>error covariance matrix</u> is

$$E\{(\mathbf{w}_{LS} - \mathbf{w}^{\circ}).(\mathbf{w}_{LS} - \mathbf{w}^{\circ})^{T}\} = E\{(U^{T}U)^{-1}U^{T}\mathbf{e}.\mathbf{e}^{T}U(U^{T}U)^{-1}\} = (U^{T}U)^{-1}U^{T}E\{\mathbf{e}.\mathbf{e}^{T}\}U(U^{T}U)^{-1}$$
$$= (U^{T}U)^{-1}$$

(*) Input variables can also be random variables, possibly correlated with the additive noise, etc... Also regression coefficients can be random variables, etc...etc... All this not considered here.

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Introduction: Least Squares Parameter Estimation

• The Mean Squared Error (MSE) of the estimation is defined as $E\{\|\mathbf{w}_{LS} - \mathbf{w}^{\circ}\|^{2}\} = E\{\text{trace}[(\mathbf{w}_{LS} - \mathbf{w}^{\circ}).(\mathbf{w}_{LS} - \mathbf{w}^{\circ})^{T}]\} = \text{trace}[(U^{T}U)^{-1}]$

 $\label{eq:PS:This MSE} \textbf{ Is different from the one in Chapter-7, check formulas}$

 Under the given assumptions, it is shown that amongst all linear estimators, i.e. estimators of the form

 $\hat{\mathbf{w}} = Z.\mathbf{d} + \mathbf{z}$ (=linear function of **d**)

the LS estimator (with $Z=(U^TU)^{-1}U^T$ and z=0) mimimizes the MSE i.e. it is the Linear Minimum MSE (MMSE) estimator

Under the given assumptions, if furthermore e is a <u>Gaussian</u> <u>distributed</u> random vector, it is shown that the LS estimator is also the ('general', i.e. not restricted to 'linear') <u>MMSE estimator</u>

Optional reading: https://en.wikipedia.org/wiki/Minimum_mean_square_error

Introduction: Least Squares Parameter Estimation

• PS: If noise **e** is zero-mean with <u>non-unit covariance matrix</u>

$$E\{\mathbf{e}.\mathbf{e}^T\} = V = V^{1/2}.V^{T/2}$$

where $V^{1/2}$ is the upper triangular Cholesky factor ('square root')

the Linear MMSE estimator & error covariance matrix are

$$\hat{\mathbf{w}} = (U^T V^{-1} U)^{-1} U^T V^{-1} \mathbf{d}$$
 $E\{(\hat{\mathbf{w}} - \mathbf{w}^0).(\hat{\mathbf{w}} - \mathbf{w}^0)^T\} = (U^T V^{-1} U)^{-1}$

which corresponds to the <u>LS estimator</u> for the so-called pre-whitened observation model

 $\underbrace{V^{-1/2}\mathbf{d}}_{\widetilde{\mathbf{d}}} = \underbrace{V^{-1/2}U}_{\widetilde{U}}.\mathbf{w}^{0} + \underbrace{V^{-1/2}\mathbf{e}}_{\widetilde{\mathbf{e}}}$

where the additive noise is indeed white.. $E\{\tilde{\mathbf{e}}.\tilde{\mathbf{e}}^T\} = V^{-1/2}E\{\mathbf{e}.\mathbf{e}^T\}V^{-T/2} = I$

Example: If $V=\sigma^2.I$ then $\mathbf{w}^{\mathbf{A}} = (\mathbf{U}^{\mathsf{T}}\mathbf{U})^{-1}\mathbf{U}^{\mathsf{T}}\mathbf{d}$ with error covariance matrix $\sigma^2.(\mathbf{U}^{\mathsf{T}}\mathbf{U})^{-1}$

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Introduction: Least Squares Parameter Estimation

• <u>PS</u>: If an initial estimate $\hat{\mathbf{w}}^{0}$ is available (e.g. from previous observations) with error covariance matrix

$$E\{(\underbrace{\hat{\mathbf{w}}^0 - \mathbf{w}^0}_{\mathbf{e}^0}).(\hat{\mathbf{w}}^0 - \mathbf{w}^0)^T\} = P = P^{1/2}.P^{T/2}$$

where $P^{1/2}$ is the upper triangular Cholesky factor ('square root'),

the **Linear MMSE estimator** & error covariance matrix are

$$\hat{\mathbf{w}} = (\underbrace{P^{-1} + U^{T}V^{-1}U}_{\text{EXT}})^{-1} \cdot (\underbrace{P^{-1}\hat{\mathbf{w}}^{0} + U^{T}V^{-1}\mathbf{d}}_{\text{EXT}})$$

$$E\{(\hat{\mathbf{w}} - \mathbf{w}^{0}) \cdot (\hat{\mathbf{w}} - \mathbf{w}^{0})^{T}\} = (P^{-1} + U^{T}V^{-1}U)^{-1}$$

which corresponds to the **LS estimator** for the model

$$\begin{bmatrix}
\mathbf{d}_{\text{EXT}} & U_{\text{EXT}} \\
P^{-1/2}\hat{\mathbf{w}}^{0} \\
V^{-1/2}\mathbf{d}
\end{bmatrix} = \begin{bmatrix}
P^{-1/2}I \\
V^{-1/2}U
\end{bmatrix} \cdot \mathbf{w}^{0} + \begin{bmatrix}
P^{-1/2}\mathbf{e}^{0} \\
V^{-1/2}\mathbf{e}
\end{bmatrix}$$

Example: P-1=0 corresponds to ∞ variance of the initial estimate, i.e. back to p.7

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Kalman Filter Basics

A **Kalman Filter** also solves a parameter estimation problem, but now the parameter vector is **dynamic** instead of static, i.e. changes over time

The time-evolution of the parameter vector is described by the 'state equation' in a <u>state-space model</u>, and the linear regression model of p.4 then corresponds to the 'output equation' of the state-space model (details in next slides...)

- In the next slides, the general
 Kalman Filter problem statement is given
- In p.14 it is seen how this relates to previous fixed parameter estimation problem

Kalman Filters are used everywhere! (aerospace, economics manufacturing, instrumentation, weather forecasting, navigation, ...)

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Rudolf Emil Kálmán (1930 -20

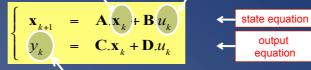
Kalman Filter Basics

State space model

of a time-invariant discrete-time system

state vector

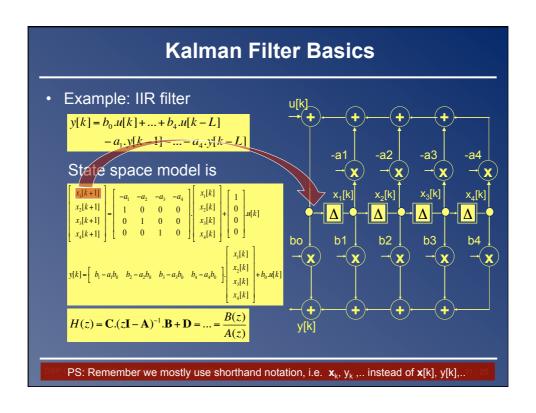
input signal

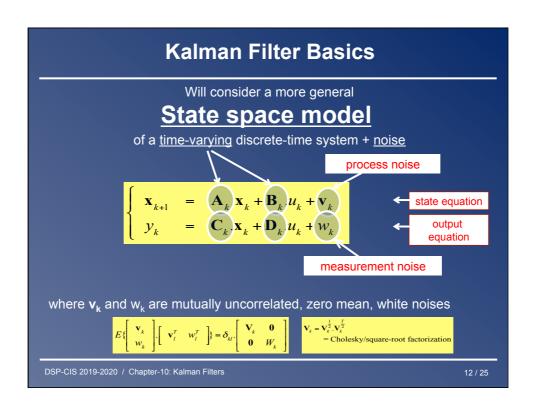


output signal

- This is single-input/single-output ('SISO'), can also have multiple inputs and multiple outputs ('MIMO')
- For L-th order system, **x**[k] is L-vector (then **A**=LxL, **B**=Lx1, **C**=1xL, **D**=1x1)
- State-space model describes input-output behavior, and is equivalent to transfer function: $H(z) = C.(zI A)^{-1}.B + D$

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State estimation problem

state vector

$$\begin{cases} \mathbf{x}_{k+1} &= \mathbf{A}_{k} \cdot \mathbf{x}_{k} + \mathbf{B}_{k} \cdot u_{k} + \mathbf{v}_{k} \\ y_{k} &= \mathbf{C}_{k} \cdot \mathbf{x}_{k} + \mathbf{D}_{k} \cdot u_{k} + w_{k} \end{cases}$$

Given... A_k , B_k , C_k , D_k , V_k , W_k , k=0,1,2,...

and input/output observations $u_k, y_k, k=0,1,2,...$

Then... estimate the internal state vectors \mathbf{x}_k , k=0,1,2,...

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Kalman Filter Basics

Fixed parameter estimation (see p.4) is seen to be a special case, with

State space model

$$\begin{cases} \mathbf{w}_{k+1}^o = I \cdot \mathbf{w}_k^o + 0 + 0 \\ \mathbf{y}_k = \mathbf{u}_k^T \cdot \mathbf{v}_k^o + 0 + e_k \end{cases}$$
 the place of the state verbut is assumed to be time-invariant
$$\mathbf{w}_{k+1}^o = \mathbf{w}_k^o$$

Parameter vector wo takes the place of the state vector,

 $\mathbf{u}^{\mathsf{T}_{\mathsf{k}}}$ takes the place of \mathbf{C}_{k} !

With the above substitutions, Kalman filter algorithms will be turned into Recursive Least Squares algorithms (standard (p. 25) & square-root (p. 22)) · · ·

Kalman Filter Basics

- **Definition**:
- = Linear MMSE-estimate of xk using all available data up until time I
- `FILTERING' = estimate $\hat{\mathbf{X}}_{klk}$
- $\hat{\mathbf{x}}_{k|k-n}, n > 0$
- `SMOOTHING' = estimate $\hat{\mathbf{x}}_{kk+n}, n > 0$
- Kalman filter will compute $\hat{\mathbf{x}}_{k|k}$ and $\hat{\mathbf{x}}_{k+1|k}$ @ time k

For every estimate, a corresponding error covariance matrix will be defined/computed, i.e.

$$\begin{split} P_{k|k} &= P_{k|k}^{1/2}. P_{k|k}^{T/2} = E\{(\hat{\mathbf{x}}_{k|k} - \mathbf{x}_k).(\hat{\mathbf{x}}_{k|k} - \mathbf{x}_k)^T\} \\ P_{k+1|k} &= P_{k+1|k}^{1/2} P_{k+1|k}^{T/2} = E\{(\hat{\mathbf{x}}_{k+1|k} - \mathbf{x}_{k+1}).(\hat{\mathbf{x}}_{k+1|k} - \mathbf{x}_{k+1})^T\} \end{split}$$

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Kalman Filter Algorithms

 First, the state estimation @ time k corresponds to a (large) parameter estimation problem in a linear regression model (see p.4), where the parameter vector contains all state vectors $\mathbf{x}_0 \cdots \mathbf{x}_{k+1}$, i.e.

$$\begin{bmatrix} \hat{\mathbf{x}}_{0|-1} \\ -B_0 u_0 \\ y_0 - D_0 u_0 \\ -B_1 u_1 \\ y_1 - D_1 u_1 \\ \vdots \\ -B_k u_k \\ y_k - D_k u_k \end{bmatrix} = \begin{bmatrix} \boxed{\begin{bmatrix} \boxed{I} & 0 & 0 & \dots & 0 \\ A_0 & -I & 0 & \dots & 0 \\ C_0 & 0 & 0 & \dots & 0 \\ 0 & A_1 & -I & \dots & 0 \\ 0 & C_1 & 0 & \dots & 0 \\ \vdots & \vdots & \vdots & & \vdots \\ \hline 0 & 0 & 0 & A_k & -I \\ 0 & 0 & 0 & C_k & 0 \end{bmatrix} \cdot \begin{bmatrix} \mathbf{x}_0 \\ \mathbf{x}_1 \\ \mathbf{x}_2 \\ \vdots \\ \mathbf{x}_{k+1} \end{bmatrix} + \begin{bmatrix} \mathbf{e}_0 \\ \mathbf{v}_0 \\ w_0 \\ \mathbf{v}_1 \\ \vdots \\ \mathbf{v}_k \\ w_k \end{bmatrix}.$$

PS: $\mathbf{x}^{\wedge}_{0l-1}$ is initial estimate in the sense of p.8

$$\begin{array}{ll} E\{\mathbf{x}_0\} &= \hat{\mathbf{x}}_{0|-1} \\ E\{\underbrace{(\hat{\mathbf{x}}_{0|-1} - \mathbf{x}_0)}_{\mathbf{e}_0} (\hat{\mathbf{x}}_{0|-1} - \mathbf{x}_0)^T\} &= P_{0|-1} = P_{0|-1}^{\frac{1}{2}} P_{0|-1}^{\frac{7}{2}} \end{array}$$

Kalman Filter Algorithms

If the covariances for \mathbf{e}_0 , \mathbf{v}_i and w_i differ from the identity, i.e.

See p.7 !

$$E\{\mathbf{e}\cdot\mathbf{e}^T\}\neq I$$

it is necessary to perform a **pre-whitening**:

$$\begin{bmatrix} P_{0|-1}^{-\frac{1}{2}} \cdot \hat{\mathbf{x}}_{0|-1} \\ -\tilde{B}_{0}u_{0} \\ \tilde{y}_{0} - \tilde{D}_{0}u_{0} \\ -\tilde{B}_{1}u_{1} \\ \tilde{y}_{1} - \tilde{D}_{1}u_{1} \\ \vdots \\ -\tilde{B}_{k}u_{k} \\ \tilde{y}_{k} - \tilde{D}_{k}u_{k} \end{bmatrix} = \begin{bmatrix} \begin{bmatrix} P_{0|-1}^{\frac{1}{2}} & 0 & 0 & \dots & 0 \\ \tilde{A}_{0} & -V_{0}^{-\frac{1}{2}} & 0 & 0 & \dots & 0 \\ \tilde{A}_{0} & -V_{0}^{-\frac{1}{2}} & 0 & \dots & 0 \\ 0 & \tilde{A}_{1} & -V_{1}^{-\frac{1}{2}} & \dots & 0 \\ 0 & \tilde{C}_{1} & 0 & \dots & \vdots \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & 0 & \tilde{A}_{k} & -V_{k}^{-\frac{1}{2}} \\ 0 & 0 & 0 & \tilde{C}_{k} & 0 \end{bmatrix} \cdot \begin{bmatrix} \mathbf{x}_{0} \\ \tilde{\mathbf{x}}_{1} \\ \mathbf{x}_{2} \\ \vdots \\ \mathbf{x}_{k+1} \end{bmatrix} + \underbrace{\begin{bmatrix} \tilde{e}_{0} \\ \tilde{v}_{0} \\ \tilde{w}_{0} \\ \tilde{v}_{1} \\ \tilde{w}_{1} \\ \vdots \\ \tilde{v}_{k} \\ \tilde{w}_{k} \end{bmatrix}}_{\tilde{e}}$$

where

$$\mathbf{\tilde{e}}_0 = \mathbf{\textit{P}}_{0|-1}^{-\frac{1}{2}} \cdot \mathbf{e}_0$$

$$\tilde{A}_i = V_i^{-\frac{1}{2}} \cdot A_i$$

on that

$$E\{\mathbf{\tilde{e}}\cdot\mathbf{\tilde{e}}^T\}=I.$$

Similar derivation, but not considered here (except p.22) for clarity... (i.e. stick to previous page)

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Kalman Filter Algorithms

- Linear regression model (p.16 or 17) has L+(k+1).(L+1) equations in (k+1).L unknowns, i.e. corresponds to an overdetermined set of linear equations
- Linear MMSE state estimation problem now comes down to computing the least squares solution to this overdetermined set of linear equations, which may be done by applying the QRD method.

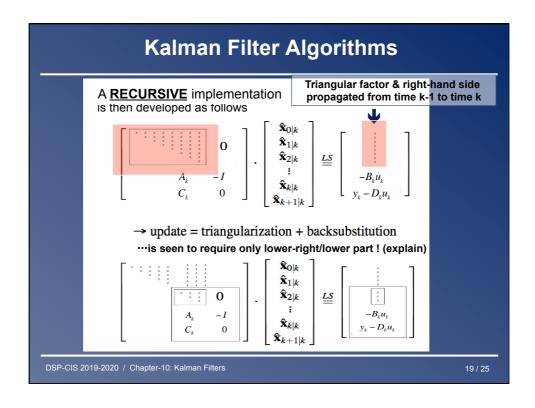
The least squares solution is obtained by first performing a *QR-factorization* and then a *backsubstitution*.

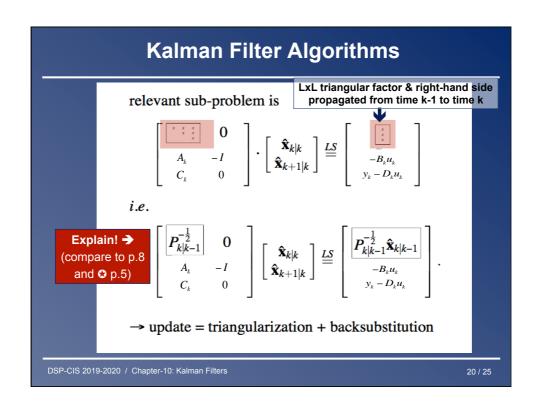
The end result is

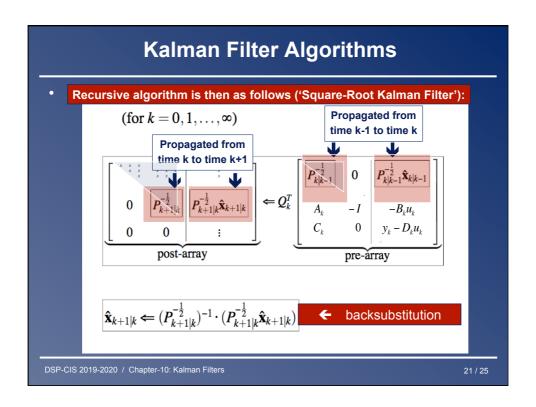
$$\begin{bmatrix} \hat{\mathbf{x}}_{0|k}^T & \hat{\mathbf{x}}_{1|k}^T & \hat{\mathbf{x}}_{2|k}^T & \dots & \hat{\mathbf{x}}_{k|k}^T & \hat{\mathbf{x}}_{k+1|k}^T \end{bmatrix}^T$$

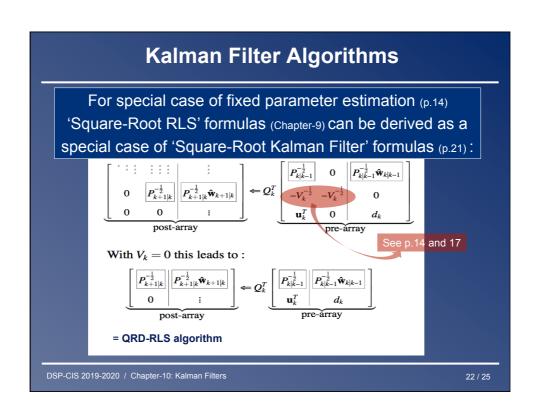
 Note that # equations as well as # unknowns grows with time, hence need (cheaper) recursive algorithm (with QRD updating as in Chapter-9)!

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Kalman Filter Algorithms

 In textbooks, mostly 'Standard (a.k.a. conventional) Kalman Filter' formulas are given, i.e.

Initialization:

$$E\{\mathbf{x}_0\} = \hat{\mathbf{x}}_{0|-1}$$

$$E\{\underbrace{(\hat{\mathbf{x}}_{0|-1} - \mathbf{x}_0)}_{\mathbf{e}_0} (\hat{\mathbf{x}}_{0|-1} - \mathbf{x}_0)^T\} = P_{0|-1}^{\frac{1}{2}} = P_{0|-1}^{\frac{T}{2}} = P_{0|-1}^{\frac{T}{2}}$$

For k=0,1,2,...

Step-1: Measurement Update (corresponding to output equation)

$$\begin{split} & \left[P_{k|k} = P_{k|k-1} - P_{k|k-1} C_k^T (W_k + C_k P_{k|k-1} C_k^T)^{-1} C_k P_{k|k-1} \right] \\ & \left[\mathbf{\hat{x}}_{k|k} = \mathbf{\hat{x}}_{k|k-1} + P_{k|k} C_k^T W_k^{-1} \cdot (y_k - C_k \mathbf{\hat{x}}_{k|k-1} - D_k u_k) \right] \end{split}$$

Step-2: Time Update

(corresponding to state equation)

$$\begin{aligned} & \boxed{P_{k+1|k} = A_k P_{k|k} A_k^T + V_k} \\ & \boxed{\mathbf{\hat{x}}_{k+1|k} = A_k \cdot \mathbf{\hat{x}}_{k|k} + B_k \cdot u_k} \end{aligned}$$

read on →

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Kalman Filter Algorithms

'Standard Kalman Filter' formulas (p.23) are straightforwardly derived from 'Square-Root Kalman Filter' formulas (p.21):

Core problem is

$$\begin{bmatrix} \boxed{P_{k|k-1}^{-\frac{1}{2}}} & 0 \\ A_k & -I \\ C_k & 0 \end{bmatrix} \begin{bmatrix} \mathbf{\hat{x}}_{k|k} \\ \mathbf{\hat{x}}_{k+1|k} \end{bmatrix} \stackrel{LS}{=} \begin{bmatrix} \boxed{P_{k|k-1}^{-\frac{1}{2}} \mathbf{\hat{x}}_{k|k-1}} \\ -B_k u_k \\ y_k - D_k u_k \end{bmatrix}.$$

L+1 equations in $\hat{\mathbf{x}}_{k|i}$ can be worked into measurement update eq. L equations in $\hat{\mathbf{x}}_{k+1|k}$ can be worked into state update eq.

[details omitted] can be worked into state update e

In infinite precision, algorithms are equivalent In finite precision, square-root algorithm is preferred

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Kalman Filter Algorithms

For special case of fixed parameter estimation (p.14) 'Standard RLS' formulas (Chapter-8) are straightforwardly derived as a special case of 'Standard Kalman Filter' formulas (p.23):

$$\begin{cases} \mathbf{w}_{k+1}^{o} = I \cdot \mathbf{w}_{k}^{o} + 0 + 0 \\ y_{k} = \mathbf{u}_{k}^{T} \cdot \mathbf{w}_{k}^{o} + 0 + e_{k} \quad E\{e_{k}^{2}\} = 1 \\ = \mathbf{C}_{k} \end{cases}$$

Same substitutions in the conventional KF:

$$P_{k|k} = P_{k|k-1} - rac{P_{k|k-1} \mathbf{u}_k \mathbf{u}_k^T P_{k|k-1}}{1 + \mathbf{u}_k^T P_{k|k-1} \mathbf{u}_k}$$

$$\mathbf{\hat{w}}_{k|k} = \mathbf{\hat{w}}_{k|k-1} + P_{k|k}\mathbf{u}_k \cdot (d_k - \mathbf{u}_k^T \mathbf{\hat{w}}_{k|k-1})$$

$$egin{aligned} egin{aligned} egin{aligned} egin{aligned} P_{k+1|k} &= P_{k|k} \ \hline \hat{\mathbf{w}}_{k+1|k} &= \hat{\mathbf{w}}_{k|k} \end{aligned} \end{aligned}$$
 'void'

= standard RLS algorithm

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